

Formulation of Fitness Functions

- Aggregation (or averaging) of constraints

$$\bar{c} = \frac{1}{C} \sum_{i=1}^C c_i$$

- We can then define a fitness function as

$$f_i = \begin{cases} \varepsilon(\bar{c} - 1) & \bar{c} < 1 \\ m_i & \bar{c} = 1 \end{cases} \quad (\text{maximize metric})$$

$$f_i = \begin{cases} \varepsilon(\bar{c} - 1) & \bar{c} < 1 \\ \frac{1}{m_i} & \bar{c} = 1 \end{cases} \quad (\text{minimize metric})$$

Formulation of Fitness Functions

- An approach to minimize computational effort

calculate constraint c_i

$$C_I = C_I + 1$$

$$C_S = C_S + c_i$$

if ($C_S < C_I$)

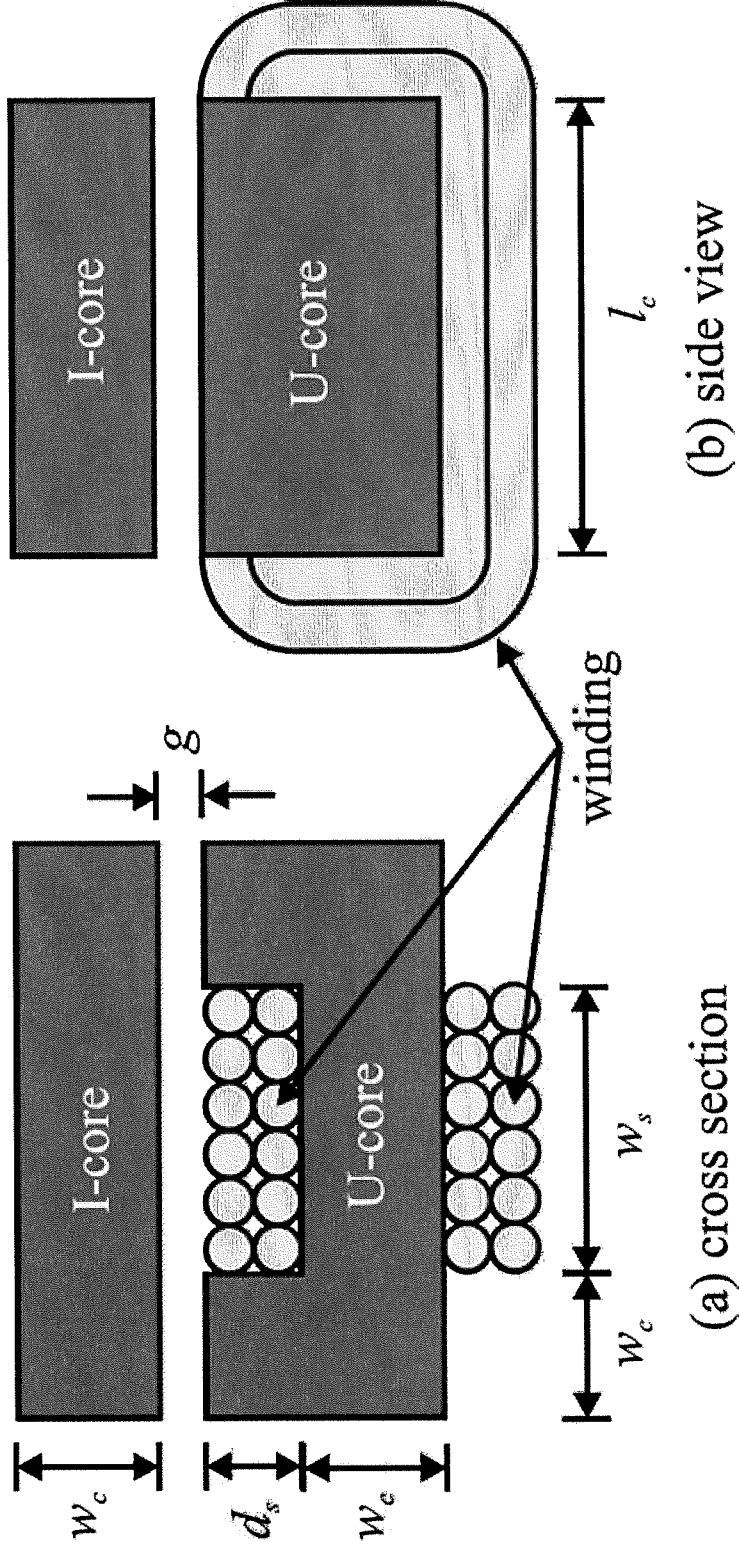
$$\mathbf{f} = \varepsilon \left(\frac{C_S - C}{C} \right) [1 \ 1 \ \dots \ 1]^T$$

return

end

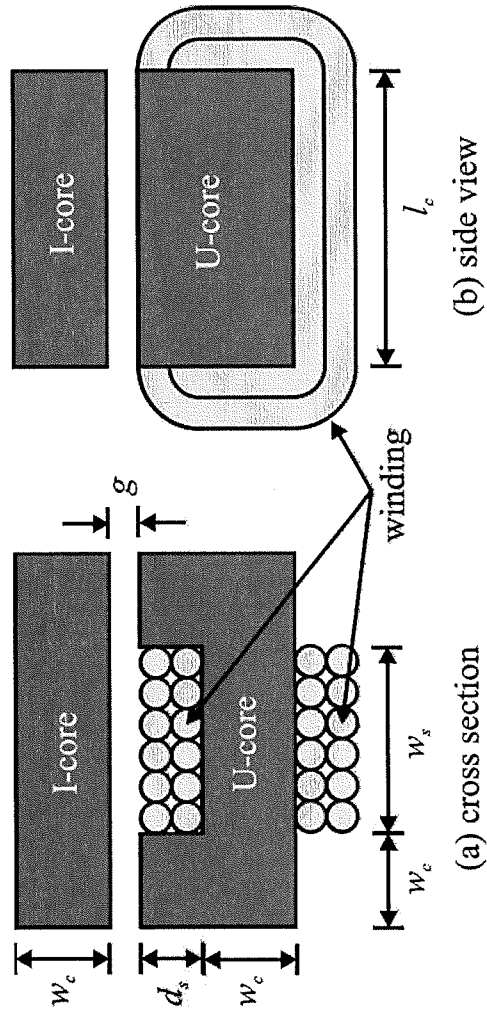
A Design Example

- Consider a UI core inductor



Free Parameters

$$\mathbf{x} = [N^* \quad d_s \quad w_s \quad w_c \quad l_c \quad g]^T$$



Analysis

$$N = \text{round}(N^*)$$

$$M = 2(2w_c + w_s + d_s)l_c w_c \rho_{mc} + (2l_c + 2w_c + \pi d_s)d_s w_s k_{pf} \rho_{wc}$$

$$P_{rt} = \frac{(2l_c + 2w_c + \pi d_s)N^2 i_{rt}^2}{d_s w_s k_{pf} \sigma_{wc}}$$

$$L = \frac{\mu_0 l_c w_c N^2}{2g}$$

$$B_{rt} = \frac{\mu_0 N i_{rt}}{2g}$$

$$J_{rt} = \frac{Ni}{w_s d_s k_{pf}}$$

Constraints

$$c_1 = \text{gte}(L, L_{mn})$$

$$c_2 = \text{lte}(B_{rt}, B_{mx})$$

$$c_3 = \text{lte}(J_{rt}, J_{mx})$$

$$c_4 = \text{lte}(P_{rt}, P_{mx})$$

$$c_5 = \text{lte}(M, M_{mx})$$

$$\bar{c} = \frac{1}{5}(c_1 + c_2 + c_3 + c_4 + c_5)$$

Fitness

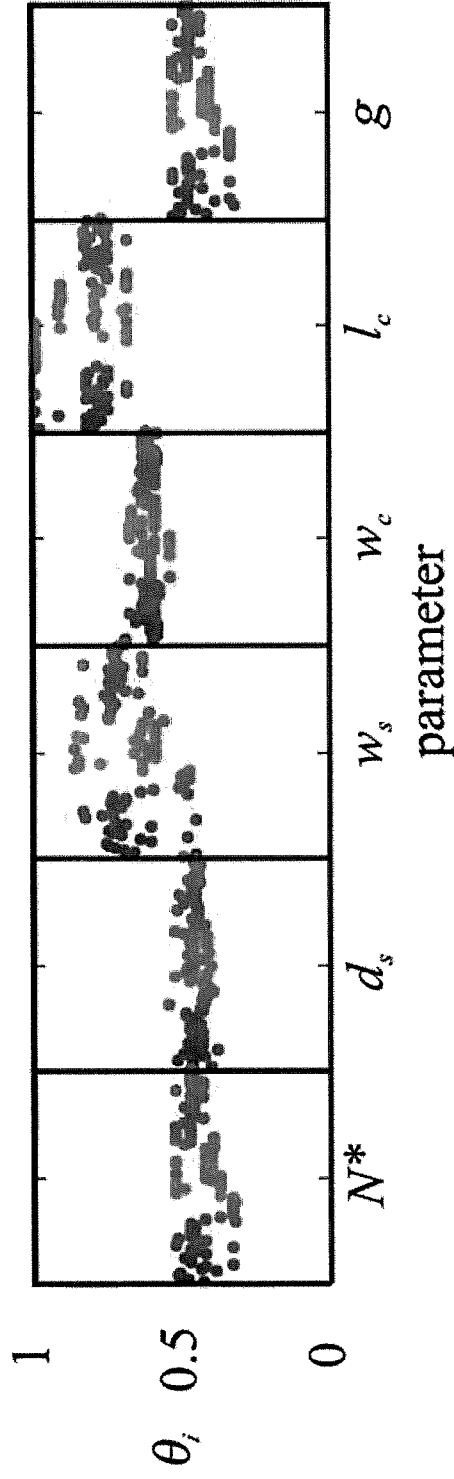
$$f = \begin{cases} \varepsilon(\bar{c}-1) & \bar{c} < 1 \\ \frac{1}{M} & \bar{c} = 1 \end{cases}$$

$$f = \begin{cases} \varepsilon(\bar{c}-1)[1 \quad 1]^T & \bar{c} < 1 \\ \begin{bmatrix} \frac{1}{M} & 1 \\ \frac{1}{M} & P_{rt} \end{bmatrix}^T & \bar{c} = 1 \end{cases}$$

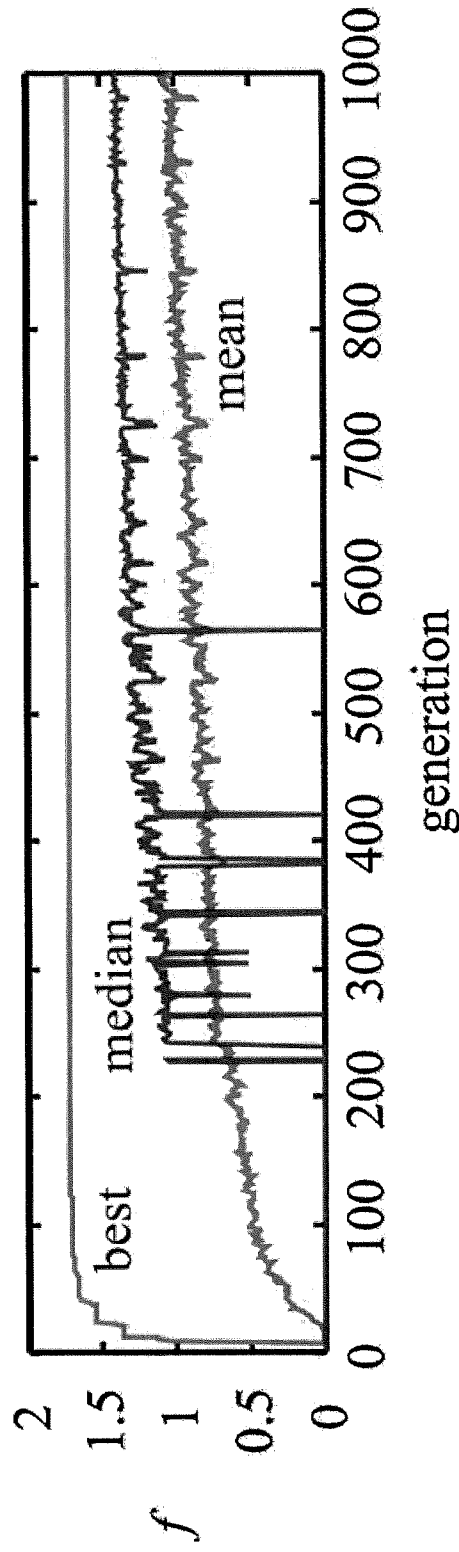
Domain of Design Parameters

parameter	N	d_s (m)	w_s (m)	w_c (m)	l_c (m)	g (m)
min. value	1	10^{-3}	10^{-3}	10^{-3}	10^{-3}	10^{-5}
max. value	10^3	10^{-1}	10^{-1}	10^{-1}	10^{-1}	10^{-2}
encoding	log	log	log	log	log	log
chromosome	1	1	1	1	1	1

Single Objective Optimization Study



(a) gene distribution



(b) fitness evolution

UI-Core Design

Turns $N = 25$ ($N^* = 25.3$)

Slot depth d_s (mm) = 8.39

Slot width w_s (mm) = 25.5

Core width w_c (mm) = 15.0

Core length l_c (mm) = 43.3

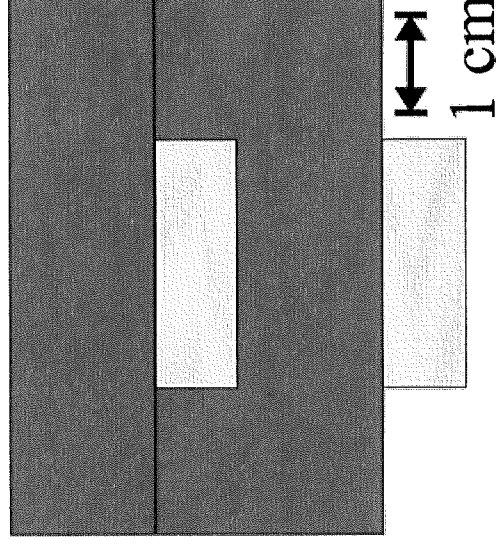
Air gap g (mm) = 0.255

Current density (A/mm^2) = 1.67

Flux density (T) = 0.617

Inductance (mH) = 1

Mass (kg) = 0.578



Design Repeatability

- Running 100 studies yields standard deviations
(normalized to means)

– N : 11%

– d_s : 4.4%

– w_s : 17%

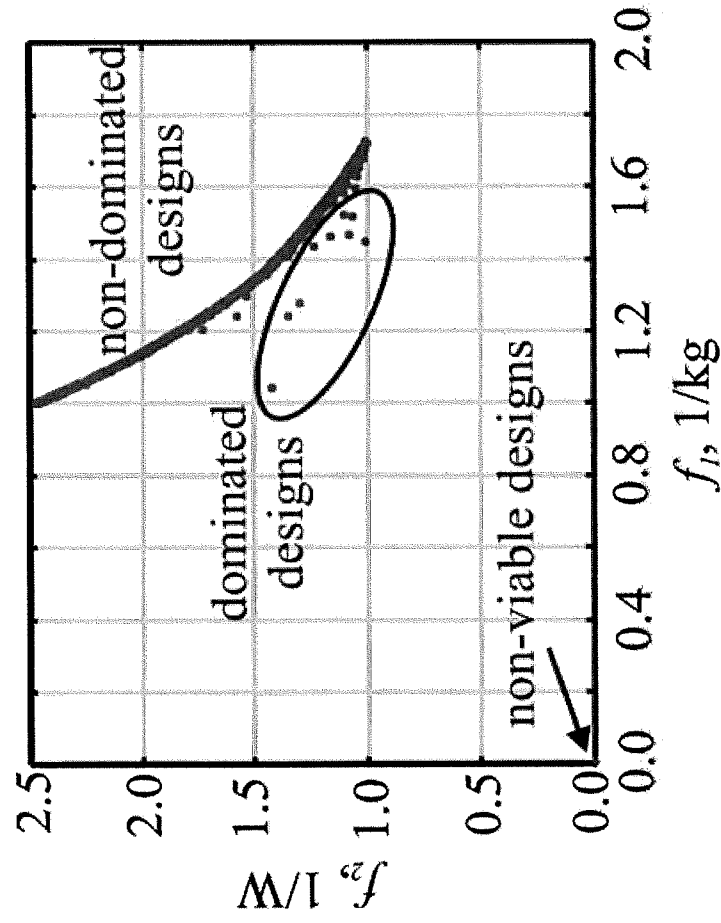
– w_c : 6.4%

– l_c : 14%

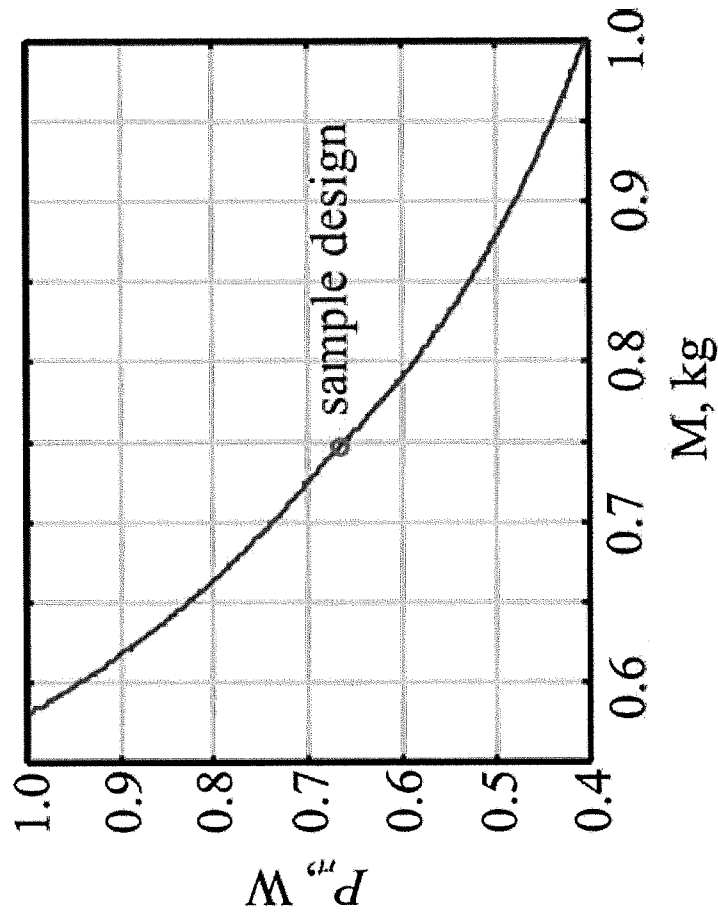
– g : 11%

– M : 1%

Multi-Objective Optimization

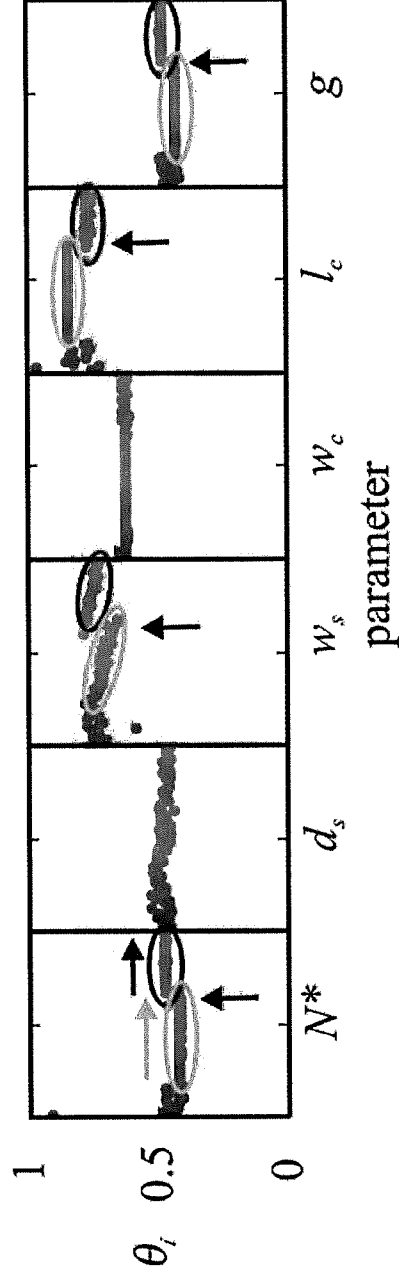


(a) objective space

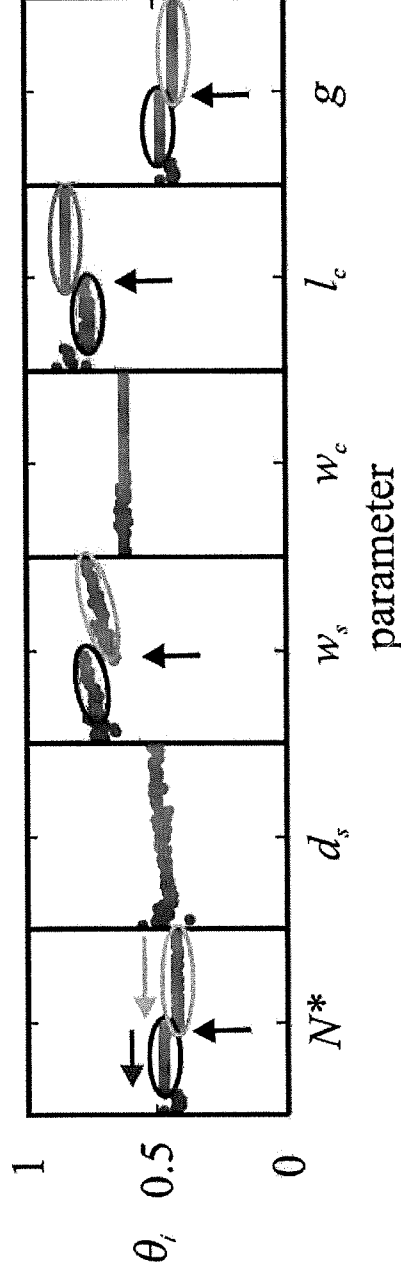


(b) Pareto-optimal front

Population Distribution



(a) gene distribution sorted by objective 1



(b) gene distribution sorted by objective 2

Optional Reading

- K. Deb, *Multi-Objective Optimization using Evolutionary Algorithms*, John Wiley & Sons, Inc., 2001
- S.D. Sudhoff, *Power Magnetic Devices: A Multi-Objective Design Approach*, Wiley/IEEE Press, 2014 (Chapter 1)

**ECE61016 Power Electronics
Converters and Systems**

Homework 2

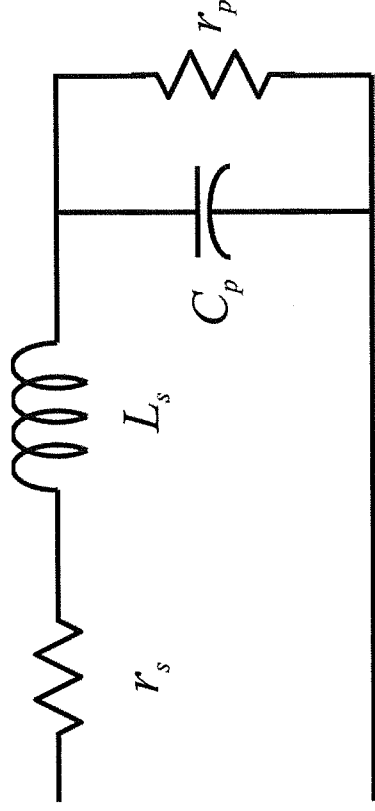
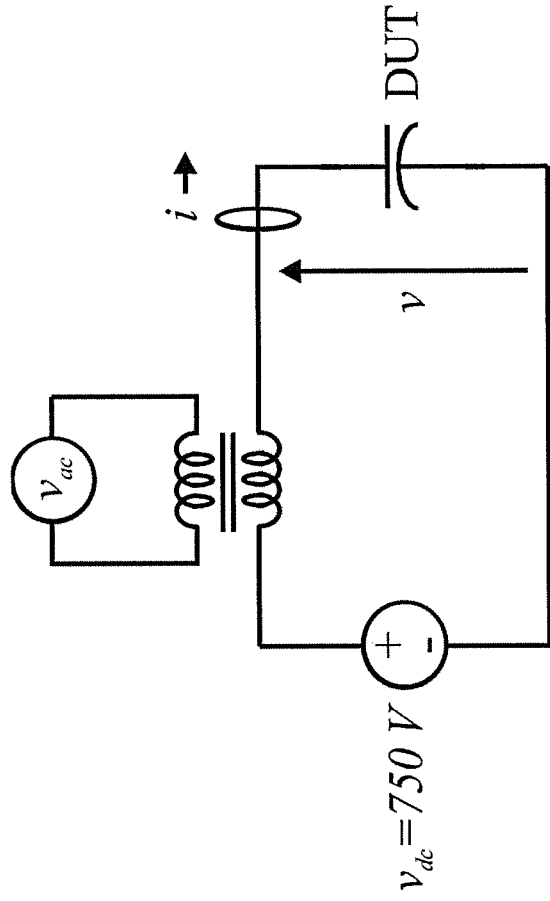
S.D. Sudhoff

Fall 2016

Aspects of Assignment

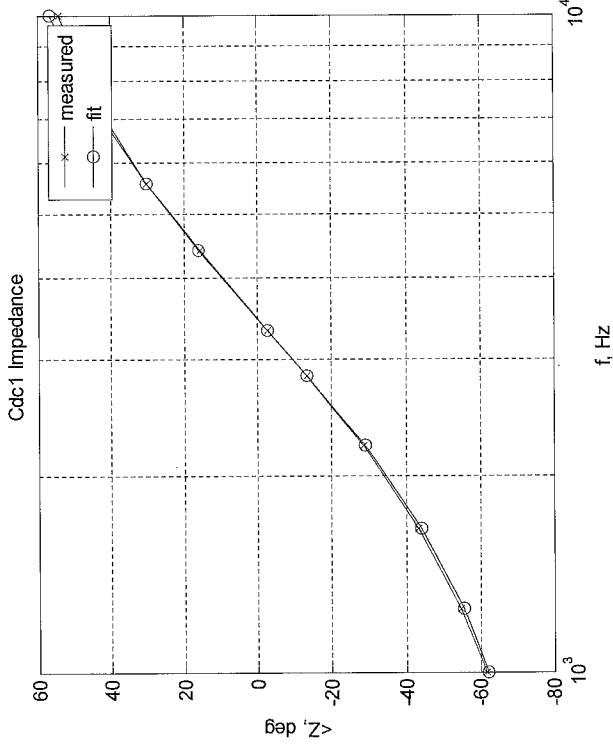
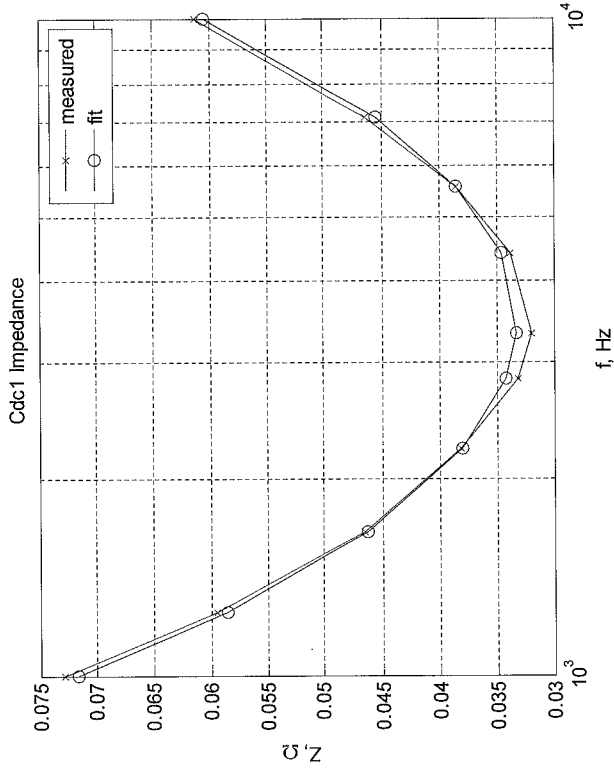
- Problem 3: Repeat Problem 1 from Assignment 1, except use Simulink. Print relevant files/scripts. Plot waveforms on same scale that we used last time.
- Problem 3B: Suppose the inductance and capacitance are all subject to +/- 20% random and uncorrelated variations. Does the control still work? Use brute force with at least 500 simulation runs to form a conclusion.
- Problem 3C: Repeat 3B for +/- 40% variations.
- Problem 4: Identify the parameters of a capacitor, by formulating a single-objective optimization problem and using GASET. Print relevant Matlab files, and parameters values.

Capacitor Frequency Response



Assume $r_p = 10^5$
Find r_s , L_s , and C_p

Capacitor Frequency Response



The file Cdc1_Impedance contains structure data
data.Zm – vector magnitude of impedance (Ohms)
data.Za – vector of angle of impedance (rad)
data.f – vector of frequency (Hz)
data.w – vector of radian frequency (rad/s)

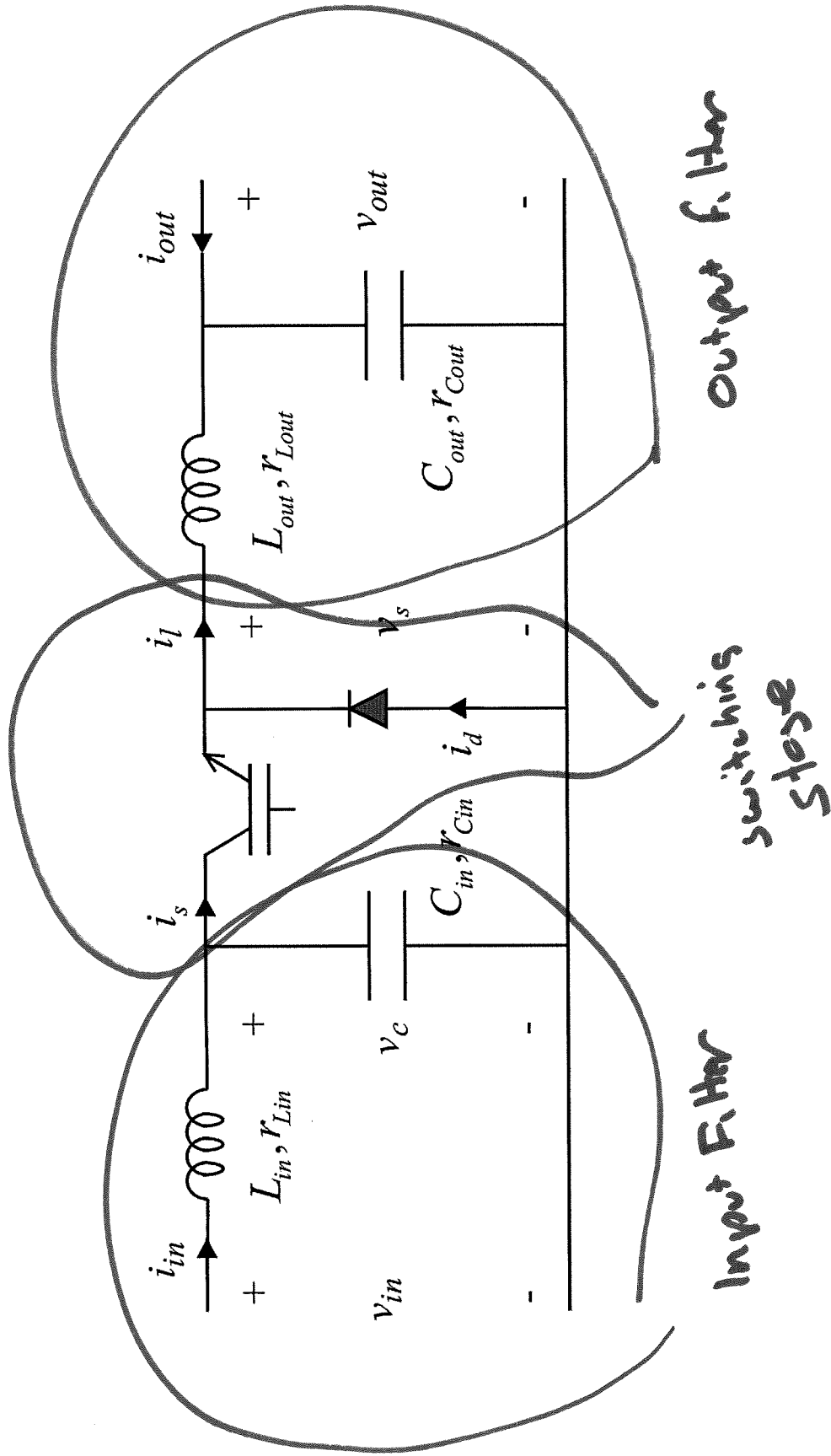
**ECE61016 Power Electronics
Converters and Systems**

**Lecture Set 3A:
Steady-State Analysis of Buck Converter**

S.D. Sudhoff

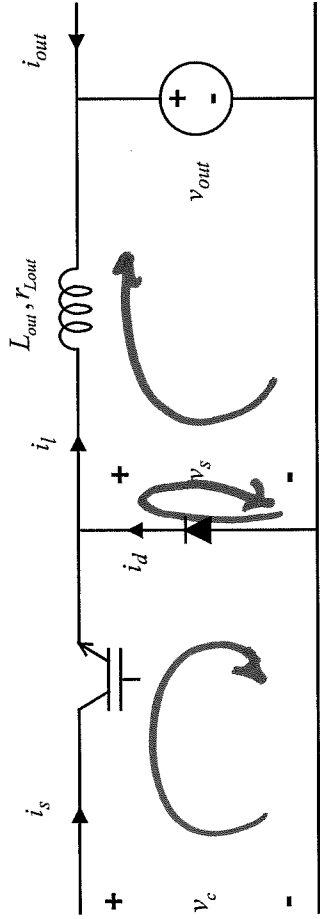
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Buck Converter



Continuous Current (CC) Buck Operation

$V_{fsw} \rightarrow$ Forward on-state drop on switch
 $V_{fd} \rightarrow$ Forward on-state diode drop

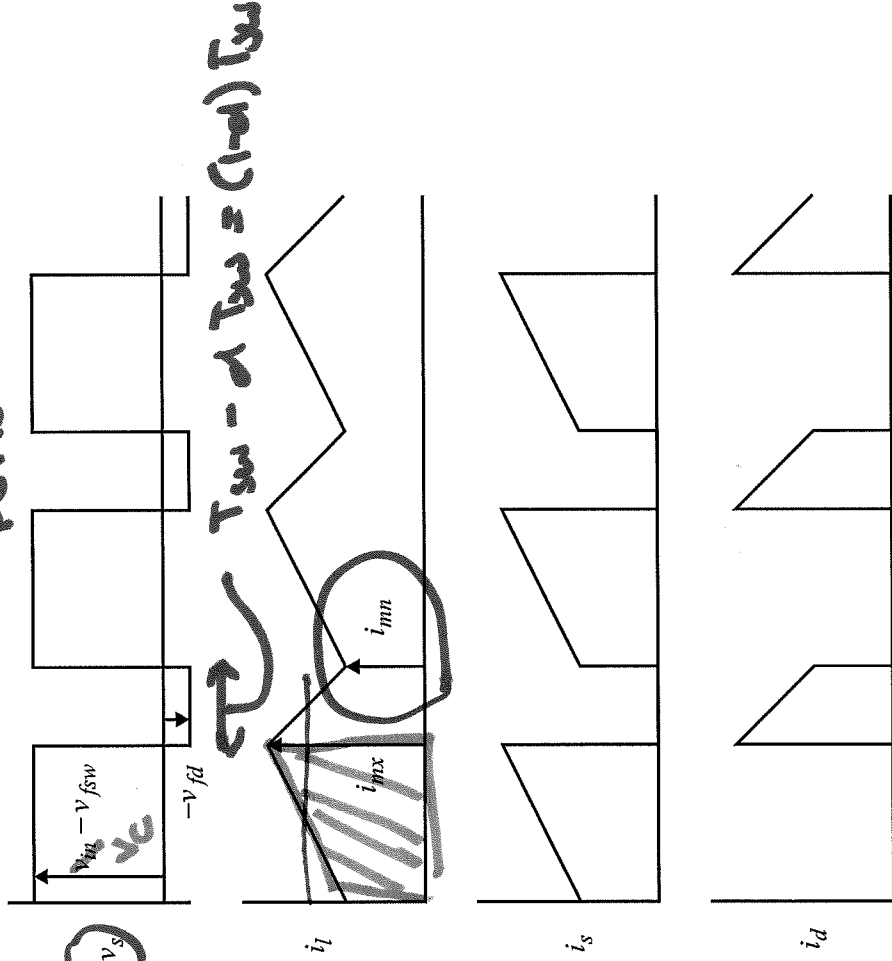
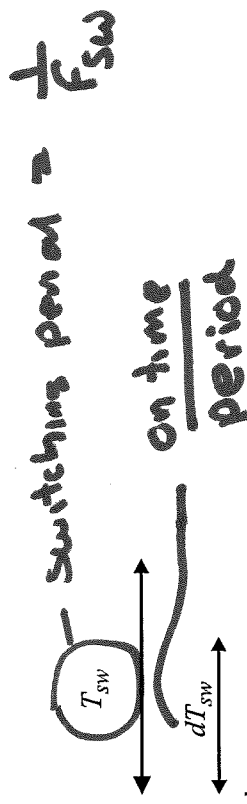


switch on

$$-V_c + V_{fsw} + v_s = 0 \quad v_s = V_c - V_{fsw}$$

switch off

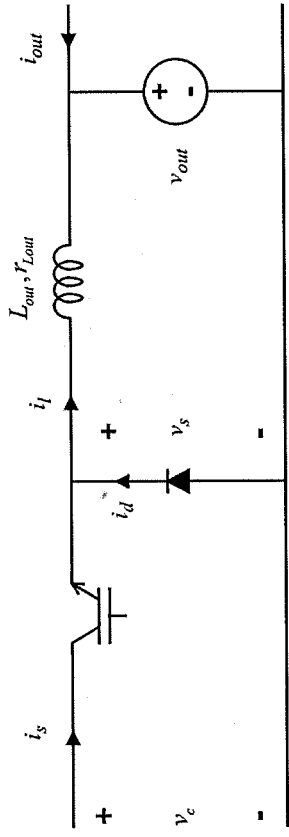
$$-V_{fd} + v_s = 0 \quad v_s = -V_{fd}$$



CC Buck Operation

- Definition of average:

$$\bar{x} = \frac{1}{T_{SW}} \int_{t_{SS}}^{t_{SS} + T_{SW}} x(t) dt$$



- Average switch point voltage

$$\begin{aligned} \bar{V}_s &= \frac{1}{T_{sw}} \int_0^{T_{sw}} V_s dt = \frac{1}{T_{sw}} \left[(V_c - V_{fsw}) \alpha T_{sw} + (-V_{fsw})(1-\alpha)T_{sw} \right] \\ &= (\bar{V}_c - V_{fsw}) \alpha - (1-\alpha)V_{fd} \end{aligned}$$

CC Buck Operation

- Average switch current

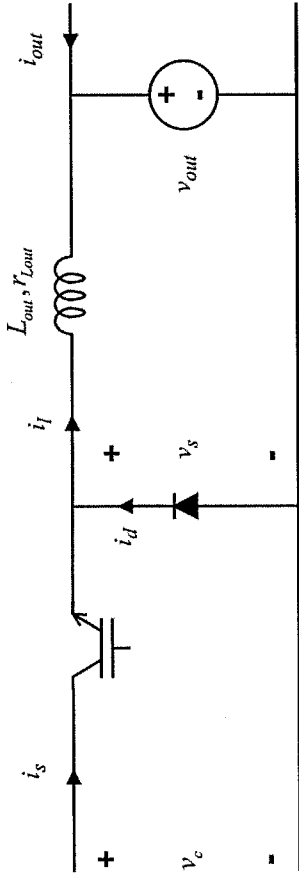
$$\bar{I}_Q = \frac{1}{T_{sw}} \int_0^{T_{sw}} I_e dt$$

$$\bar{I}_Q = \frac{1}{T_{sw}} \left[\frac{1}{2} (I_{max} + I_{min}) T_{sw} \right]$$

$$\bar{I}_S = \frac{1}{T_{sw}} \int_0^{dT_{sw}} I_e dt$$

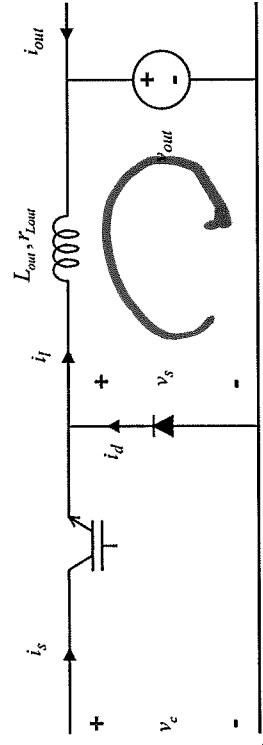
$$= \frac{1}{T_{sw}} \left[\frac{1}{2} (I_{max} + I_{min}) dT_{sw} \right]$$

$$\bar{I}_S = dI_Q$$



CC Buck Operation

- Relationship of output voltage and switch point voltage



$$-V_s + L_{out} \frac{di_L}{dt} + \int_{L_{out}} i_L + V_{out} = 0$$

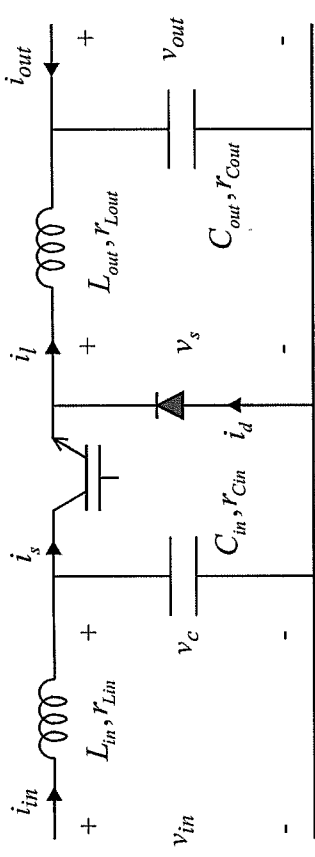
$$\bar{X} = \frac{1}{T_{sw}} \int_0^{T_{sw}} x \, dt$$

$$-\bar{V}_s + \frac{1}{T_{sw}} \int_0^{T_{sw}} L_{out} \frac{di_L}{dt} \, dt + \int_{L_{out}} \bar{i}_L + \bar{V}_{out} = 0$$

$$-\bar{V}_s + \int_{L_{out}} \bar{i}_L + \bar{V}_{out} = 0$$

CC Buck Operation

- Relationship of input and input capacitor voltages



$$-V_{in} + L_{in} \frac{di_{in}}{dt} + r_{Lin} i_{in} + V_C = 0$$

\Downarrow

$$-V_{in} + r_{Lin} I_{in} + V_C = 0$$

CC Buck Operation

- Summary of average-value steady-state CC operation

$$\bar{i}_l = -\bar{i}_{out}$$

$$\bar{i}_s = d\bar{i}_l$$

$$\bar{v}_{out} = \bar{v}_s - r_{Lout}\bar{i}_l$$

$$\bar{v}_s = d(\bar{v}_c - v_{fsw}) - (1-d)v_{fd}$$

$$\bar{v}_c = \bar{v}_{in} - r_{Lins}\bar{i}_s$$

$$\underline{I_{in} > 0}$$

CC Input Capacitor Current & Voltage Ripple

- Waveforms

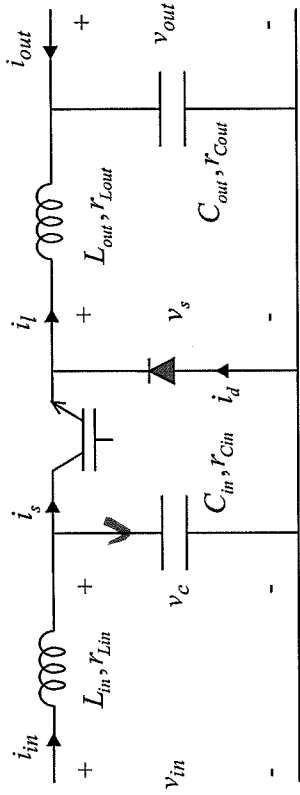


$$i_{cin} = i_{in} - i_s$$

$$X_{rms} = \sqrt{\frac{1}{T} \int_{T_{on}}^{T_{off}} x^2 dt}$$

$$T_{sw} |i_{c,rms}|^2 = \int_0^{T_{sw}} |i_{in}|^2 dt$$

$$= \int_0^{T_{sw}} (dI_0 - I_0)^2 dt + \int_{T_{off}}^{T_{sw}} 0^2 dt$$



$$i_{cin} = i_{in} - i_s$$

$$= dI_0 - i_s$$

CC Input Capacitor Current & Voltage Ripple

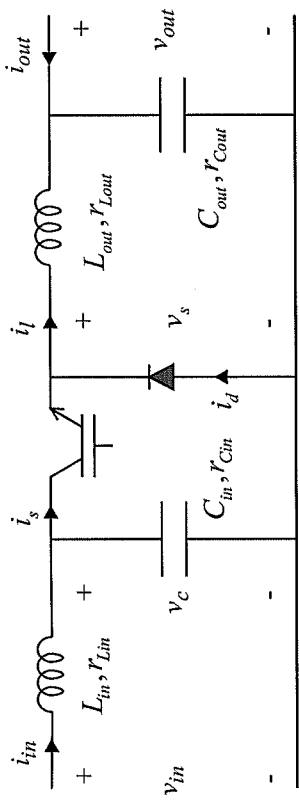
- Analysis

$$I_{C_{in, rms}}^2 = I_0^2 (d-1)^2 d I_{L_1}^2 + I_0^2 d^2 (1-d) I_{L_2}^2$$

$$= I_0^2 d [d^2 - 2d + 1 + d - d^2]$$

$$= I_0^2 d (1-d)$$

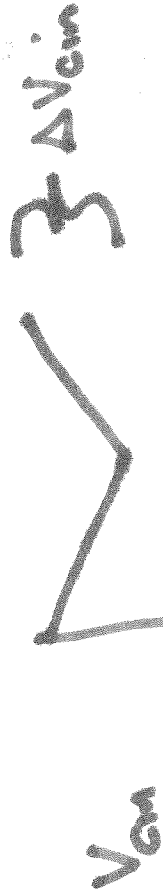
$$I_{C_{in, rms}} = I_0 \sqrt{d(1-d)}$$



CC Input Capacitor Current & Voltage Ripple

- Analysis

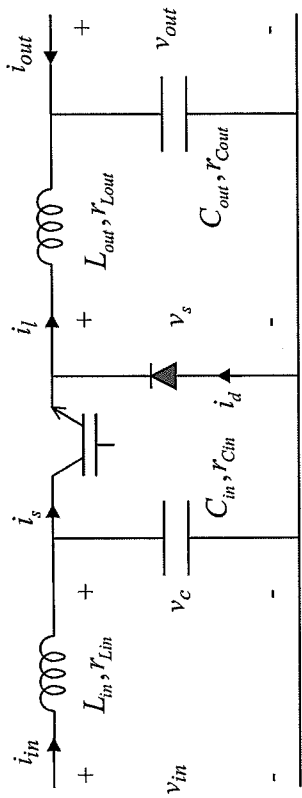
$$i_{cm} = C_{in} \frac{dV_{in}}{dt}$$



$$V_{in}(T_{sw}) = V_{in}(0) + \int_0^{T_{sw}} i_{cm} dt$$

$$\Delta V_{in} = \frac{1}{C_{in}} \int_0^{T_{sw}} i_{cm} dt = \frac{1}{C_{in}} \int_0^{T_{sw}} (1-d) I_{sw} dt$$

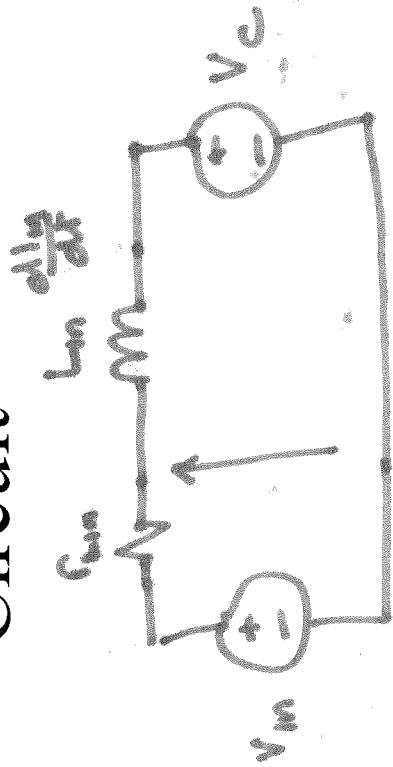
$$\Delta V_{in} = \frac{1}{C_{in} f_{sw}} \int_0^{T_{sw}} I_{sw} dt (1-d)$$



$$\frac{i_{cm}}{C_{in}} dt$$

CC Input Inductor Current Ripple

- Circuit

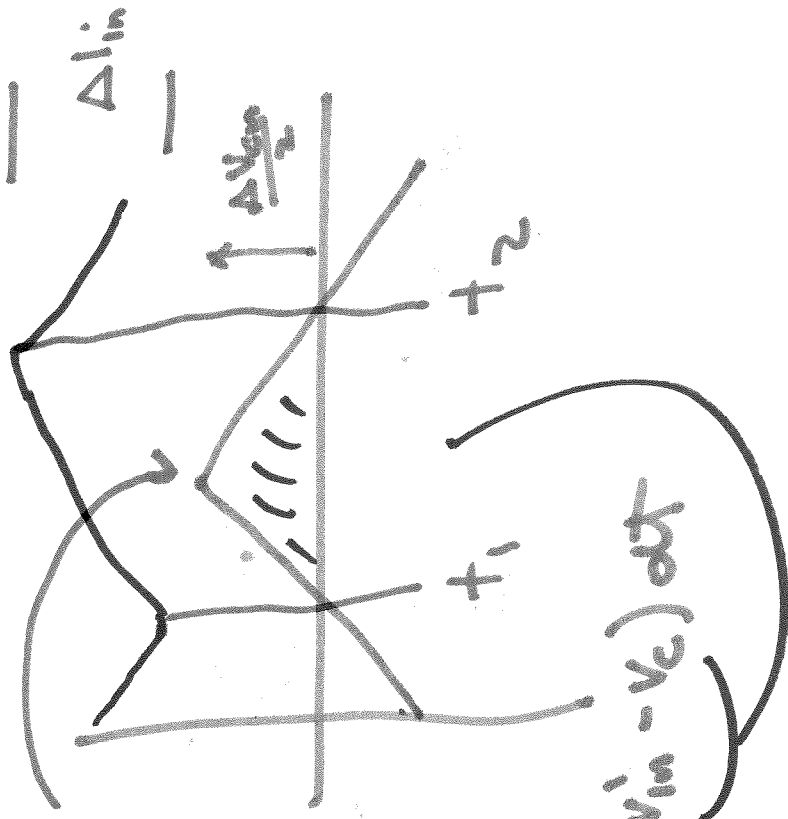
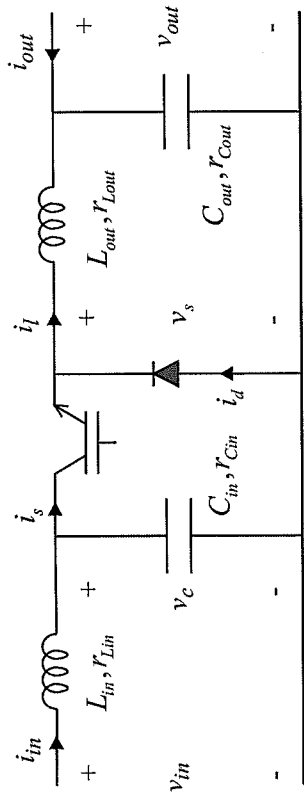


$$v_{in}^i = v_{in} - L_{in} \frac{di_{in}}{dt}$$

$$-v_{in}^i + L_{in} \frac{di_{in}}{dt} + v_c = 0$$

$$\frac{di_{in}}{dt} = \frac{1}{L_{in}} (v_{in}^i - v_c)$$

$$i_{in}(t_2) = i_{in}(t_1) + \int_{t_1}^{t_2} \frac{1}{L_{in}} (v_{in}^i - v_c) dt$$

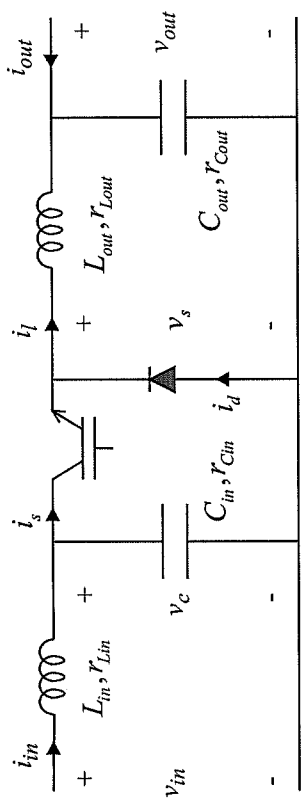


CC Input Inductor Current Ripple

- Waveforms

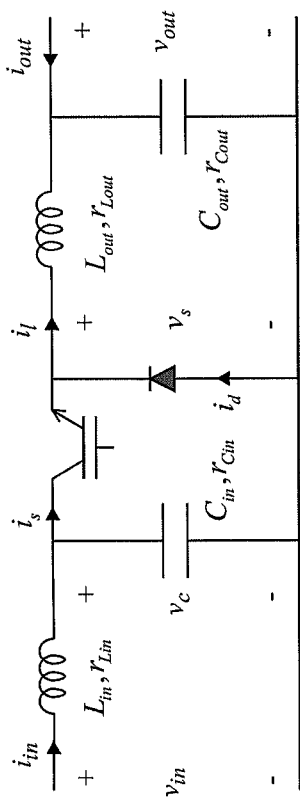
$$\Delta I_{in} = \frac{1}{L_{in}} \cdot \frac{1}{2} \cdot \Delta V_{cm} \cdot T_{sw}$$

$$\Delta I_{in} = \frac{1}{8} \frac{V_{cm}}{L_{in} f_{sw}}$$



CC Input Inductor Current Ripple

- Analysis

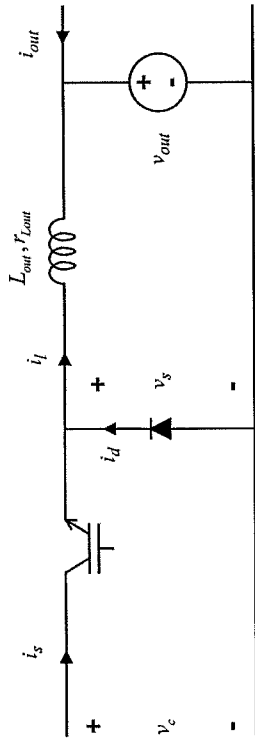


CC Output Inductor Current Ripple

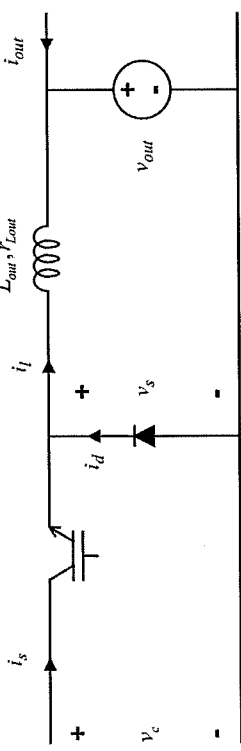
- On state $\approx \bar{I}_s$
 $-V_C + V_{fsw} + r_{Lout} \bar{I}_s + L_{out} \frac{dI_s}{dt}$
 $+ V_{out} = 0$

$$L_{out} \frac{dI_s}{dt} = V_C - V_{fsw} - r_{Lout} \bar{I}_s - V_{out}$$

$$L_{out} \frac{I_{ms} - I_{ms}}{\Delta T_{sw}} = V_C - V_{fsw} - r_{Lout} \bar{I}_s - V_{out}$$



CC Output Inductor Current Ripple



- Off state

$$-V_{fd} + L_{out} \frac{di_l}{dt} + r_{Lout} i_l + V_{out} = 0 \quad T_2$$

$$L_{out} \frac{di_l}{dt} = -V_{out} + V_{fd} - r_{Lout} i_l$$

$$+ L_{out} \frac{i_{max} - i_{min}}{(1-D)T_{sw}} = +V_{out} - V_{fd} + r_{Lout} I_o$$

CC Output Inductor Current Ripple

- In conclusion:

$$(i_{mx} - i_{mn}) \frac{L_{out}}{dT_{sw}} = V_{in} - V_{fsw} - r_{Lout} \bar{i}_l - V_{out} \quad \text{on}$$

$$(i_{mx} - i_{mn}) \frac{L_{out}}{(1-d)T_{sw}} = V_{fd} + r_{Lout} \bar{i}_l + V_{out} \quad \text{off}$$

- Combining

$$\frac{L_{out}}{T_{sw}} (i_{mx} - i_{mn}) \left[\frac{1}{d} + \frac{1}{1-d} \right] = V_{in} - V_{fsw} + V_{fd}$$

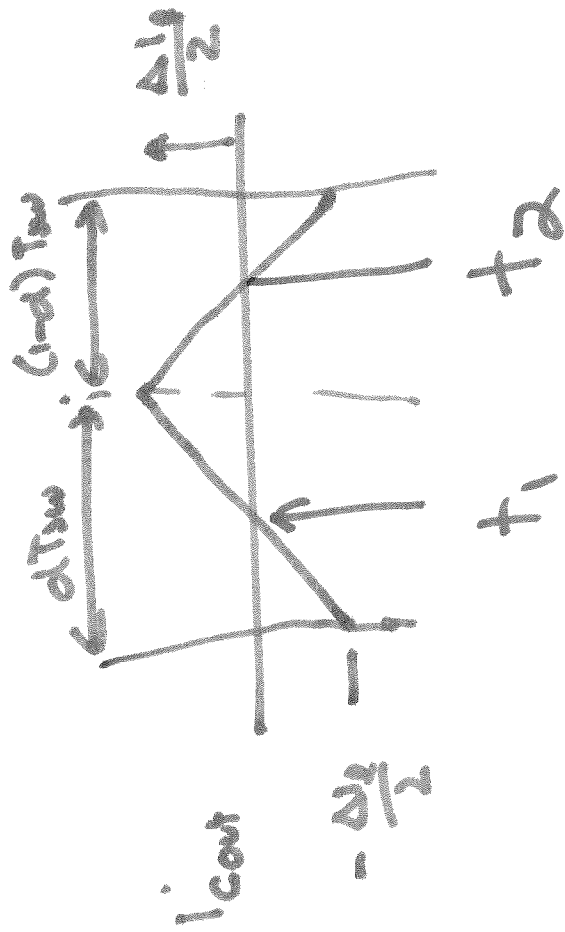
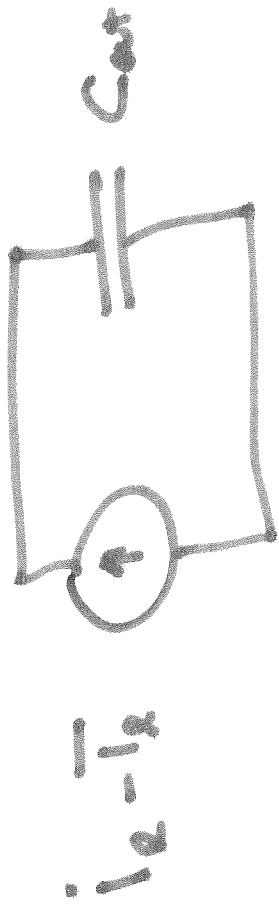
$$\frac{1}{d} + \frac{1}{1-d} = \frac{1-d+d}{d(1-d)} = \frac{1}{d(1-d)}$$

$$i_{mx} - i_{mn} = \Delta i_{Lout} = \frac{(V_{in} - V_{fsw} + V_{fd}) d(1-d)}{f_{sw} L_{out}}$$

CC Output Inductor Current Ripple

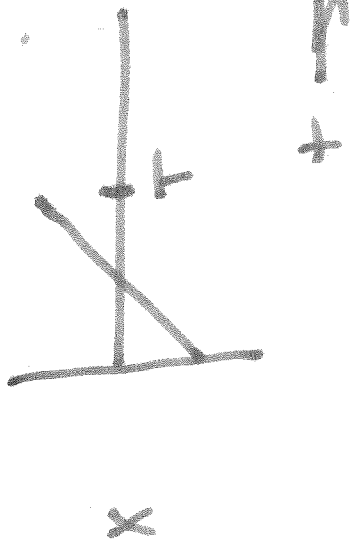
CC Output Capacitor Current & Voltage Ripple

- Circuit



CC Output Capacitor Current & Voltage Ripple

- Waveforms



$$X = \alpha + \beta t$$

$$\int_0^T X^2 dt = \int_0^T \alpha^2 + 2\alpha\beta t + \beta^2 t^2 dt$$

$$= \alpha^2 T + \alpha\beta T^2 + \frac{1}{3}\beta^2 T^3$$

$$T_{sw} I_{C_{out,rms}}^2 = \int_0^{T_{sw}} I_{C_{out}}^2 dt$$

$$= \int_0^{t_{off}} \left[-\frac{\Delta I_C}{2} + \frac{\Delta I_C}{t_{off}} t \right]^2 dt$$

$$+ \int_0^{t_{(1-off)T_{sw}}} \left[\frac{\Delta I_C}{2} - \frac{\Delta I_C}{(1-off)T_{sw}} t \right]^2 dt$$

shifted
time

CC Output Capacitor Current & Voltage Ripple

- Analysis

$$\begin{aligned} I_{Sw}^2 I_{Cap,rms} &= \left(\frac{\Delta I_L}{2}\right)^2 \alpha T_{SW} - \frac{\Delta I_L}{2} \frac{\Delta I_L}{\alpha T_{SW}} \alpha^2 T_{SW} + \frac{(\Delta I_L)^2}{3} \frac{(\alpha T_{SW})^3}{(\alpha T_{SW})^2} \\ &+ \left(\frac{\Delta I_L}{2}\right)^2 (1-\alpha) T_{SW} - \frac{\Delta I_L}{2} \frac{\Delta I_L}{(1-\alpha) T_{SW}} (\alpha T_{SW})^2 \\ &+ \frac{1}{3} \left(\frac{\Delta I_L}{(1-\alpha) T_{SW}}\right)^2 (1-\alpha)^3 T_{SW} \\ &= \Delta I_L^2 \left[\frac{1}{4} \alpha - \frac{1}{2} \alpha + \frac{1}{3} \alpha \right] \\ &+ \Delta I_L^2 \left[\frac{1}{4} (1-\alpha) - \frac{1}{2} (1-\alpha) + \frac{1}{3} (1-\alpha) \right] \\ &= \frac{\Delta I_L^2}{12} \left[\alpha + (1-\alpha) \right] = \frac{\Delta I_L^2}{12} \end{aligned}$$

$$I_{Cap,rms} = \frac{\Delta I_L}{2\sqrt{3}}$$