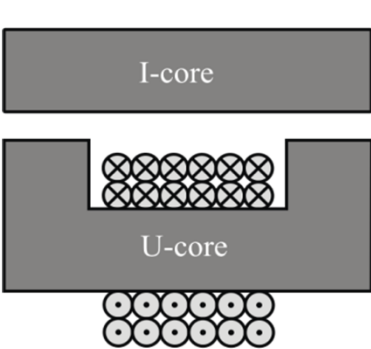
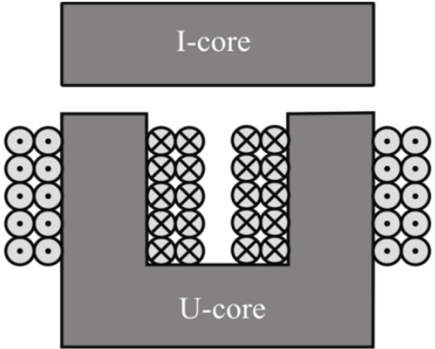

Power Magnetic Devices: A Multi-Objective Design Approach

Chapter 3: Introduction to Inductor Design

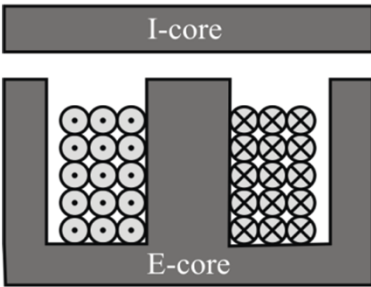
3.1 Common Inductor Architectures



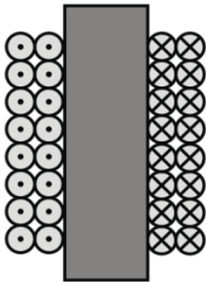
(a) UI-core



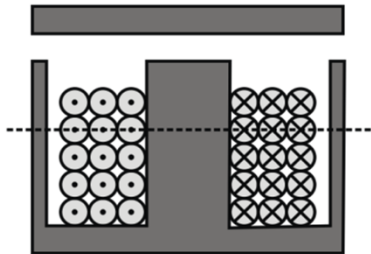
(b) alternate UI-core



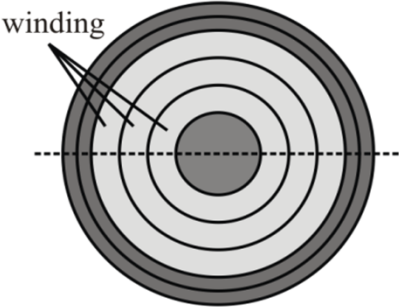
(c) EI-core



(d) solenoid



(e) bobbin core cross section



(f) bobbin core (bottom piece from above)

3.2 DC Coil Resistance

- Resistance of coil

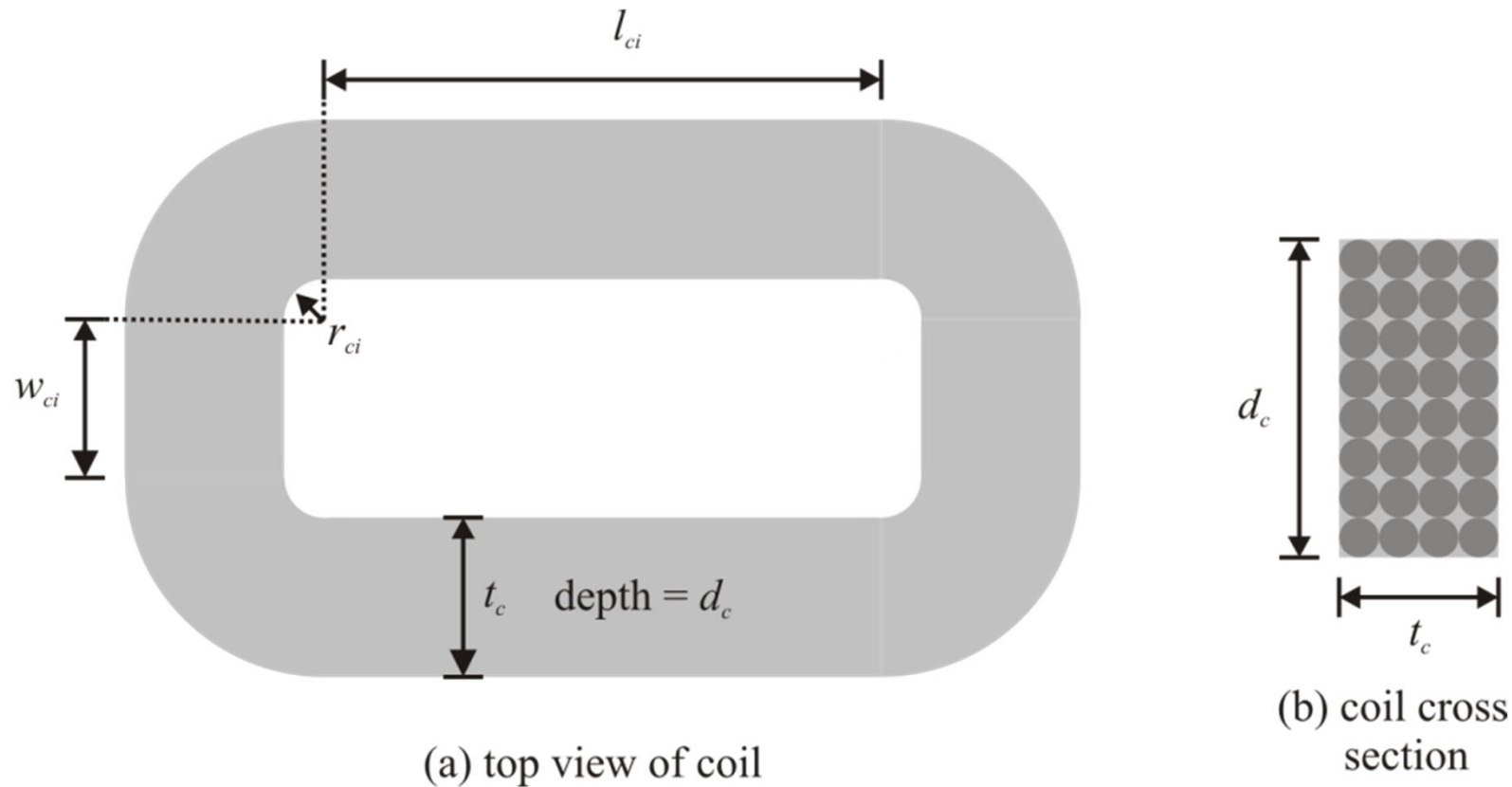
$$R = \frac{V_{cl} N^2}{k_{pf} A_{cl}^2 \sigma}$$

- Packing factor

$$k_{pf} = \frac{A_{cd}}{A_{cl}}$$

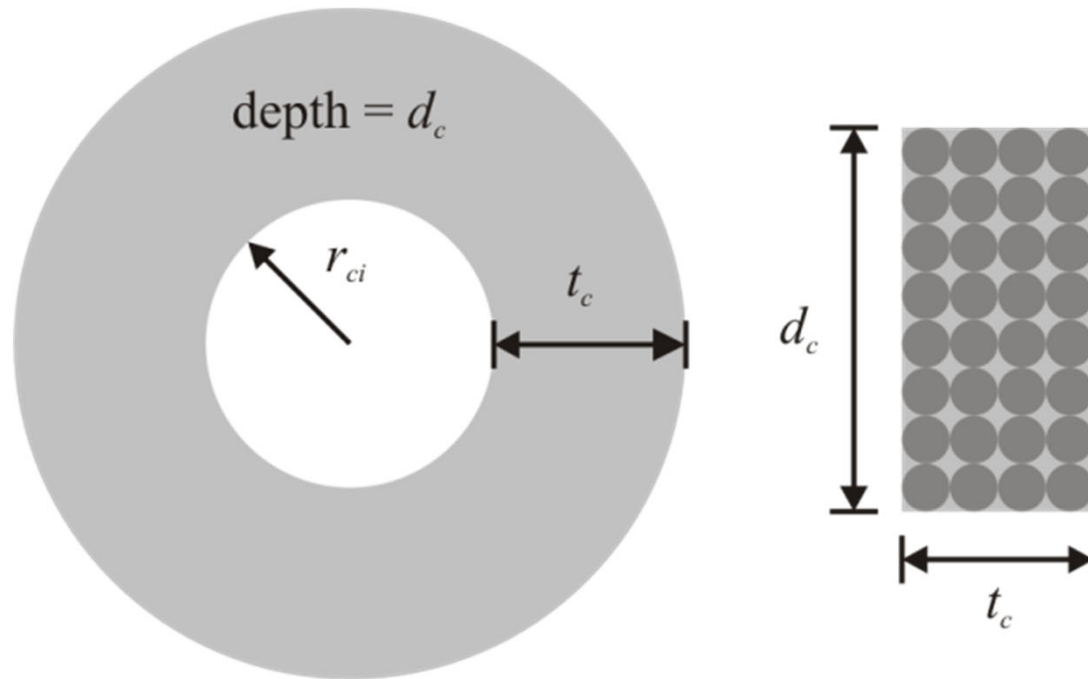
3.2 DC Coil Resistance

3.2 DC Coil Resistance



$$V_{cl} = d_c \left(\pi(t_c + r_{ci})^2 - \pi r_{ci}^2 + (2l_{ci} + 2w_{ci})t_c \right) \quad A_{cl} = t_c d_c$$

3.2 DC Coil Resistance



(a) top view of coil

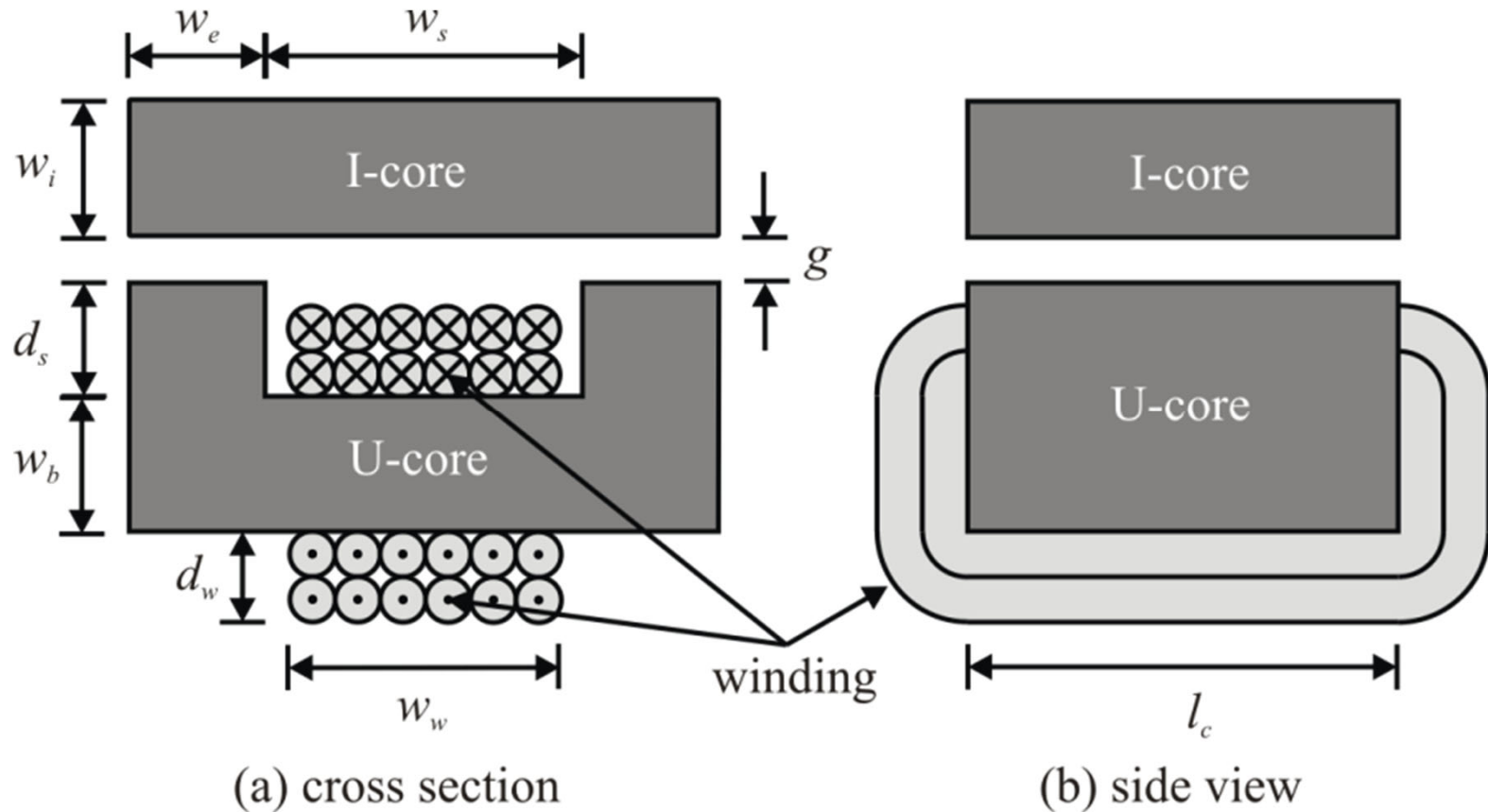
(b) cross section of coil

$$V_{cl} = \pi d_c \left((t_c + r_{ci})^2 - r_{ci}^2 \right)$$

$$A_{cl} = t_c d_c$$

3.3 DC Inductor Design

- Design space



$$\boldsymbol{\theta} = [T_{cr} \ T_{cd} \ g \ l_c \ w_e \ r_{ie} \ r_{be} \ a^* \ N^* \ N_w^* \ N_d^* \ c_w \ c_d]^T$$

3.3 DC Inductor Design

- Widths

$$w_i = r_{ie} w_e$$

$$w_b = r_{be} w_e$$

- Conductors

$$a = \text{round}_{WG}(a^*)$$

$$r_c = \sqrt{\frac{a}{\pi}}$$

$$N = \text{round}(N^*)$$

3.3 DC Inductor Design

- Coil layout

$$N_w = \text{round}(N_w^*)$$

$$N_d = \text{round}(N_d^*)$$

$$w_w = 2r_c k_b N_w$$

$$d_w = 2r_c k_b N_d$$

- More geometry

$$w_s = w_w + 2c_w$$

$$d_s = d_w + c_d$$

3.3 DC Inductor Design

- Mass and volumes

$$M_{cr} = \rho_{cr} \left((w_b + w_i)(w_s + 2w_e) + 2d_s w_e \right) l_c$$

$$V_{cl} = w_w \left(\pi d_w^2 + (2l_c + 2w_b) d_w \right)$$

$$k_{pf} = \frac{Na}{d_w w_w}$$

$$V_{cd} = k_{pf} V_{cl}$$

$$M_{cd} = \rho_{cd} V_{cd}$$

3.3 DC Inductor Design

- Total mass

$$M_L = M_{cr} + M_{cd}$$

- Loss

$$A_{cl} = d_w w_w$$

$$R = \frac{V_{cl} N^2}{k_{pf} d_w^2 w_w^2 \sigma_{cd}}$$

$$P_L = R i_{dc0}^2$$

3.3 DC Inductor Design

- Constraints

$$c_1 = \text{gte}(N_d N_w, N)$$

$$c_2 = \text{lte}(k_{pf}, k_{pf, mx})$$

$$J = \frac{i_{dc0}}{a}$$

$$c_3 = \text{lte}(J, J_{mx})$$

3.3 DC Inductor Design

- More constraints

$$h_L = d_w + w_b + d_s + g + w_i$$

$$w_L = 2w_e + w_s$$

$$l_L = 2d_w + l_c$$

$$\alpha_L = \frac{\max(h_L, w_L, l_L)}{\min(h_L, w_L, l_L)}$$

$$c_4 = \text{lte}(\alpha_L, \alpha_{mx})$$

3.3 DC Inductor Design

- Even more constraints

$$c_5 = \text{lte}(M_L, M_{Lmxa})$$

$$c_6 = \text{lte}(P_L, P_{Lmxa})$$

$$L_{inc} = \frac{\lambda|_{i_{dc0}+\Delta i} - \lambda|_{i_{dc0}-\Delta i}}{2\Delta i}$$

$$c_7 =$$

$$c_8 = \text{gte}(L_{inc}, L_{mn})$$

3.3 DC Inductor Design

- A final constraint

$$r_{\Phi} = \left| \frac{N \Phi_3 \big|_{i=i_{dc0} + \Delta i}}{\lambda \big|_{i=i_{dc0} + \Delta i}} \right|$$

$$c_9 = \text{gte}(r_{\Phi}, r_{\Phi r_q})$$

3.3 DC Inductor Design

- Pseudo-code for fitness function

```
calculate  $M_L$ ,  $P_L$ 
calculate constraints  $c_1$  through  $c_6$ 
 $C_s = c_1 + c_2 + c_3 + c_4 + c_5 + c_6$ 
 $C_I = 6$ 
if ( $C_s < C_I$ )
     $\mathbf{f} = \varepsilon \left( \frac{C_s - C}{C} \right) \begin{bmatrix} 1 \\ 1 \end{bmatrix}$ 
    return
end
construct MEC and calculate  $c_7$ ,  $c_8$ , and  $c_9$ 
 $C_s = C_I + c_7 + c_8 + c_9$ 
 $C_I = C_I + 1$ 
if ( $C_s < C_I$ )
     $\mathbf{f} = \varepsilon \left( \frac{C_s - C}{C} \right) \begin{bmatrix} 1 \\ 1 \end{bmatrix}$ 
    return
end
 $\mathbf{f} = \begin{bmatrix} 1 \\ M_L \\ 1 \\ P_L \end{bmatrix}$ 
return
```


3.4 Case Study

- Specifications

Table 3.4-1 Inductor specifications.

| Symbol | Description | Value |
|----------------|--|-----------------------|
| i_{dc0} | nominal dc operating current | 10 A |
| L_{rqi} | required incremental inductance at nominal current | 5 mH |
| J_{mxa} | Maximum allowed current density at nominal current | 7.6 MA/m ² |
| α_{mxa} | maximum allowed aspect ratio | 3 |
| M_{Lmxa} | maximum allowed mass | 5 kg |
| P_{Lmxa} | maximum allowed loss at nominal current | 100 W |
| k_{pf-mxa} | maximum allowed packing factor | 0.7 |
| k_b | winding build factor | 1.05 |
| Δi | perturbation in current for inductance calculation | 0.1 A |
| $r_{\Phi rq}$ | minimum required magnetizing flux ratio | 0.9 |

3.4 Case Study

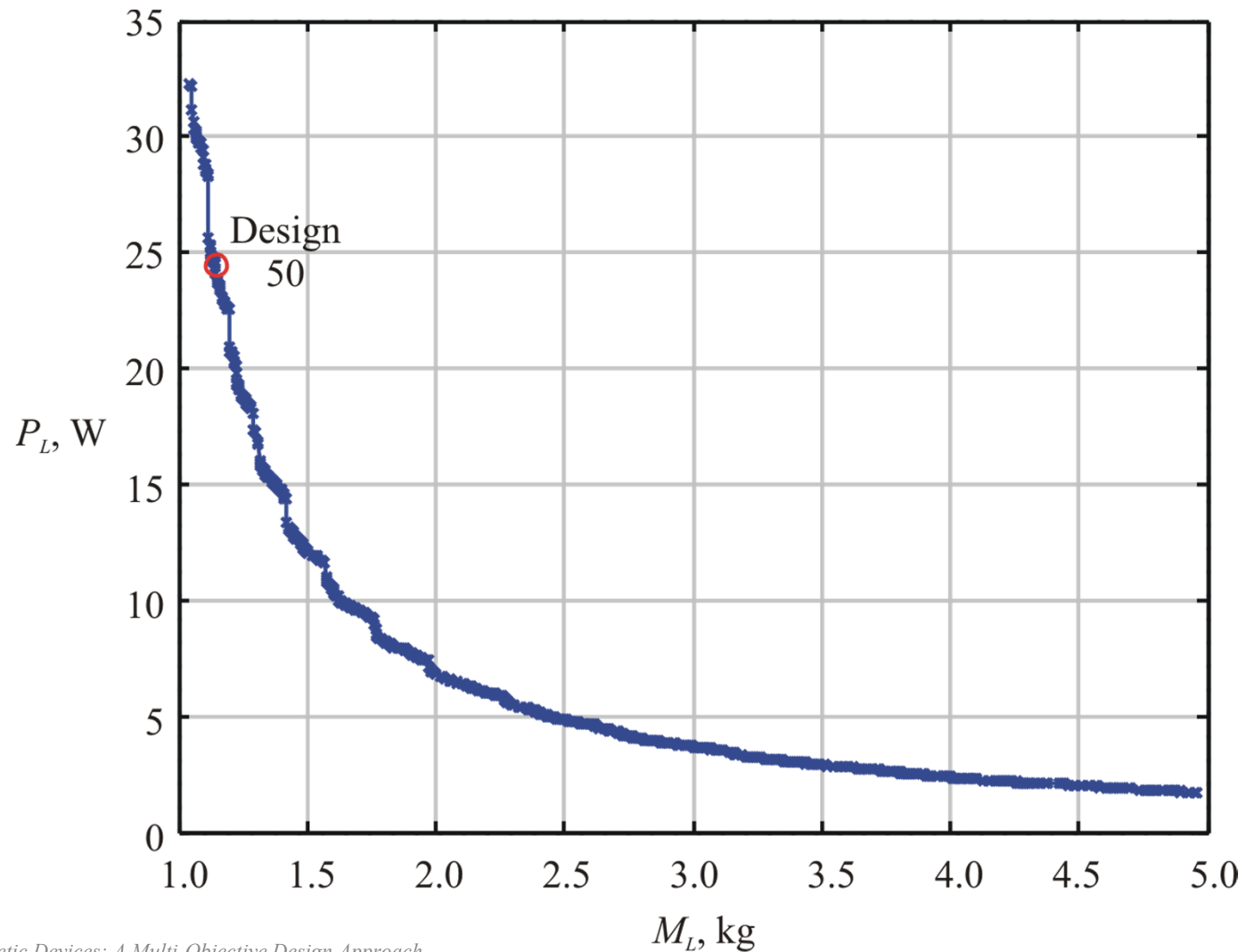
- Design space

Table 3.4-2 Design space.

| Par. | Description | Min. | Max. | Enc. | Gene |
|----------|--|-------------------|-----------|------|------|
| T_{cr} | Core material | 1 | 5 | int | 1 |
| T_{cd} | Conductor material | 1 | 2 | int | 2 |
| g | Air gap (m) | 10^{-4} | 10^{-2} | log | 3 |
| l_c | Core length | 10^{-2} | 1 | log | 4 |
| w_e | U-core end width (m) | $2 \cdot 10^{-3}$ | 10^{-1} | log | 5 |
| r_{ie} | w_i to w_e ratio | 0.5 | 1.5 | lin | 6 |
| r_{be} | w_b to w_e ratio | 0.5 | 1.5 | lin | 7 |
| a_c | Cross sectional conductor area (m ²) | 10^{-9} | 10^{-4} | log | 8 |
| N^* | Desired number of turns | 1 | 10^3 | log | 9 |
| N_w^* | Desired slot width in conductors | 1 | 10^3 | log | 10 |
| N_d^* | Desired slot depth in conductors | 1 | 10^3 | log | 11 |
| c_w | Slot clearance in width (m) | 10^{-6} | 10^{-2} | log | 12 |
| c_d | Slot clearance in depth (m) | 10^{-6} | 10^{-2} | log | 13 |

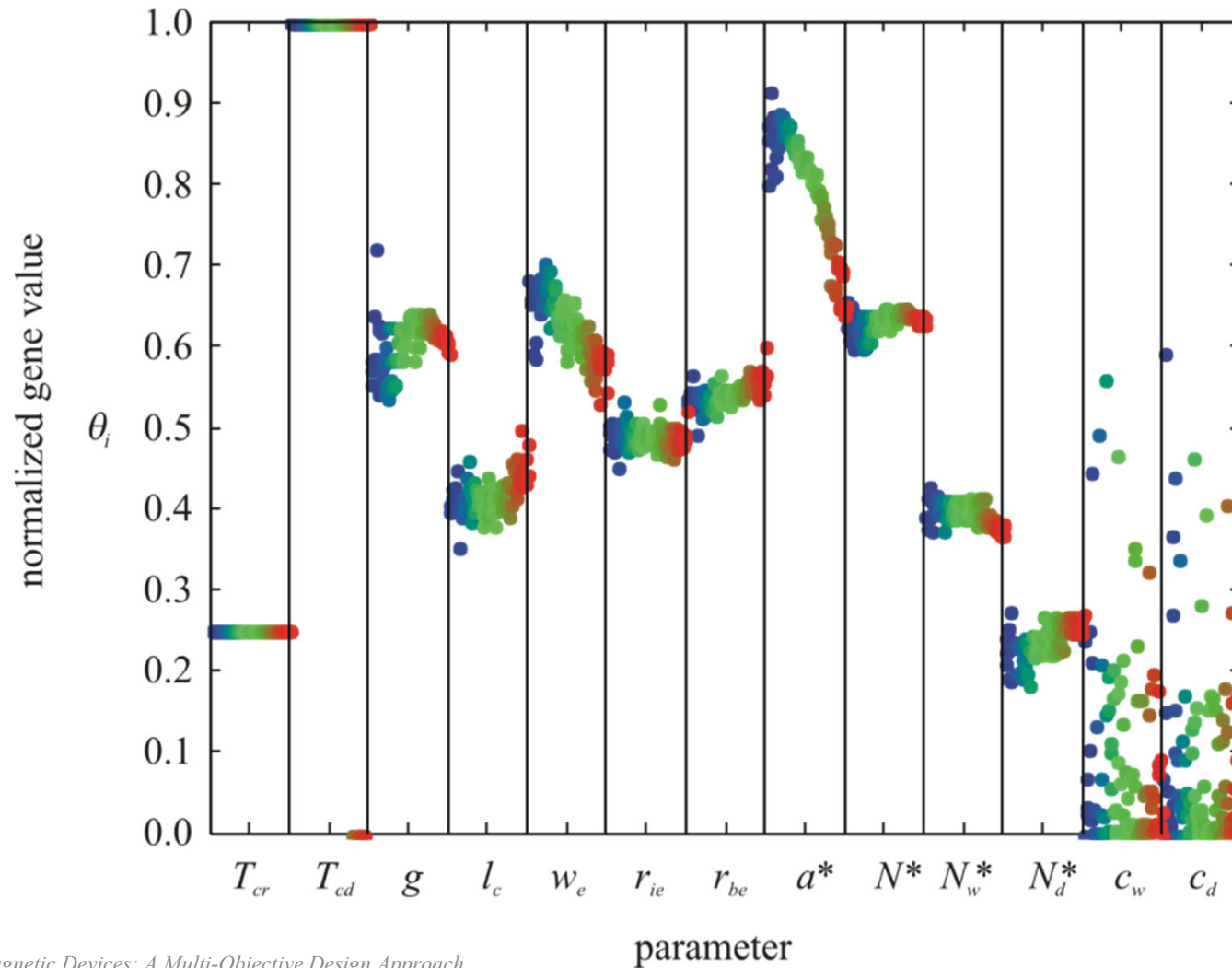
3.4 Case Study

- Pareto-optimal front



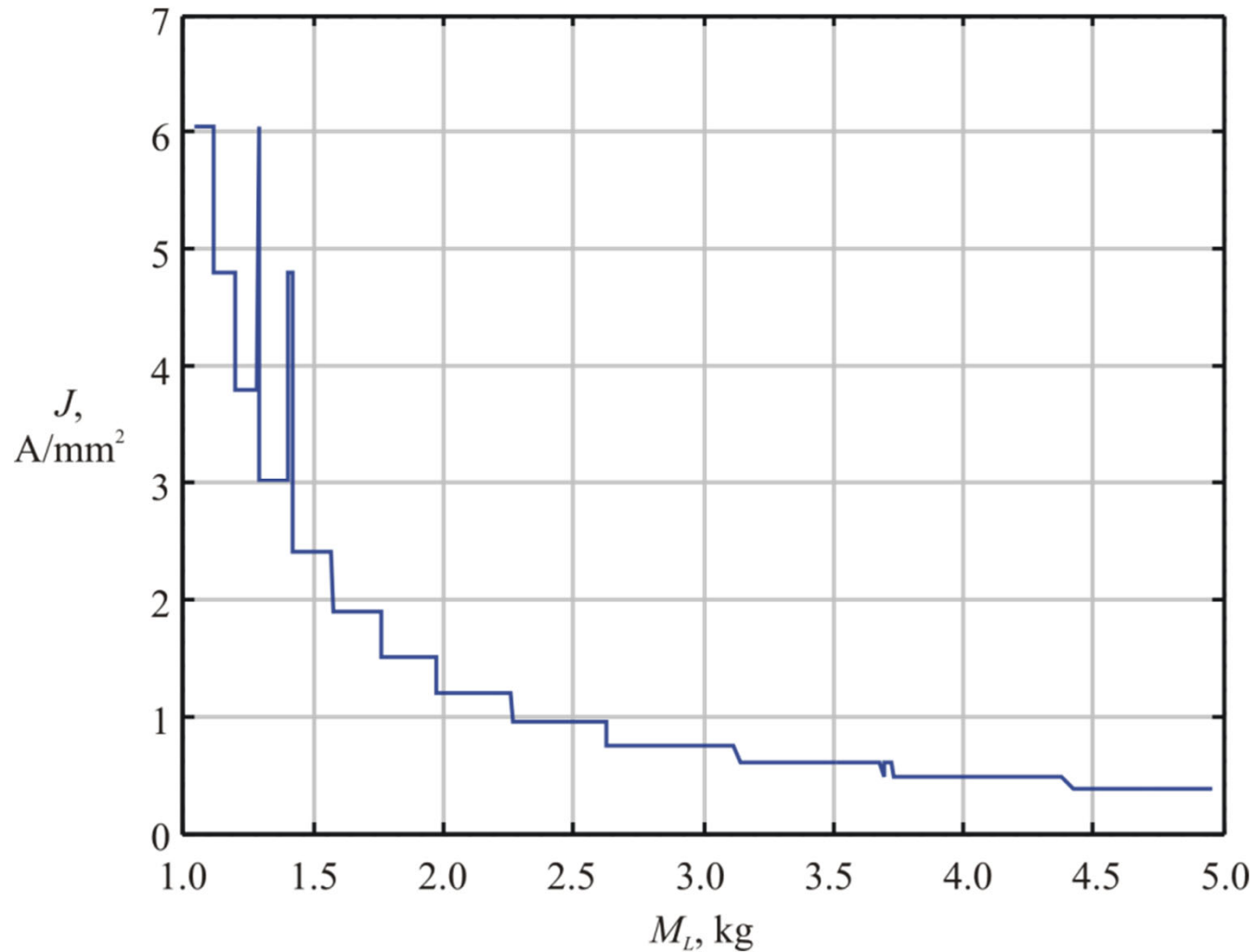
3.4 Case Study

- Population distribution (sorted by first objective)



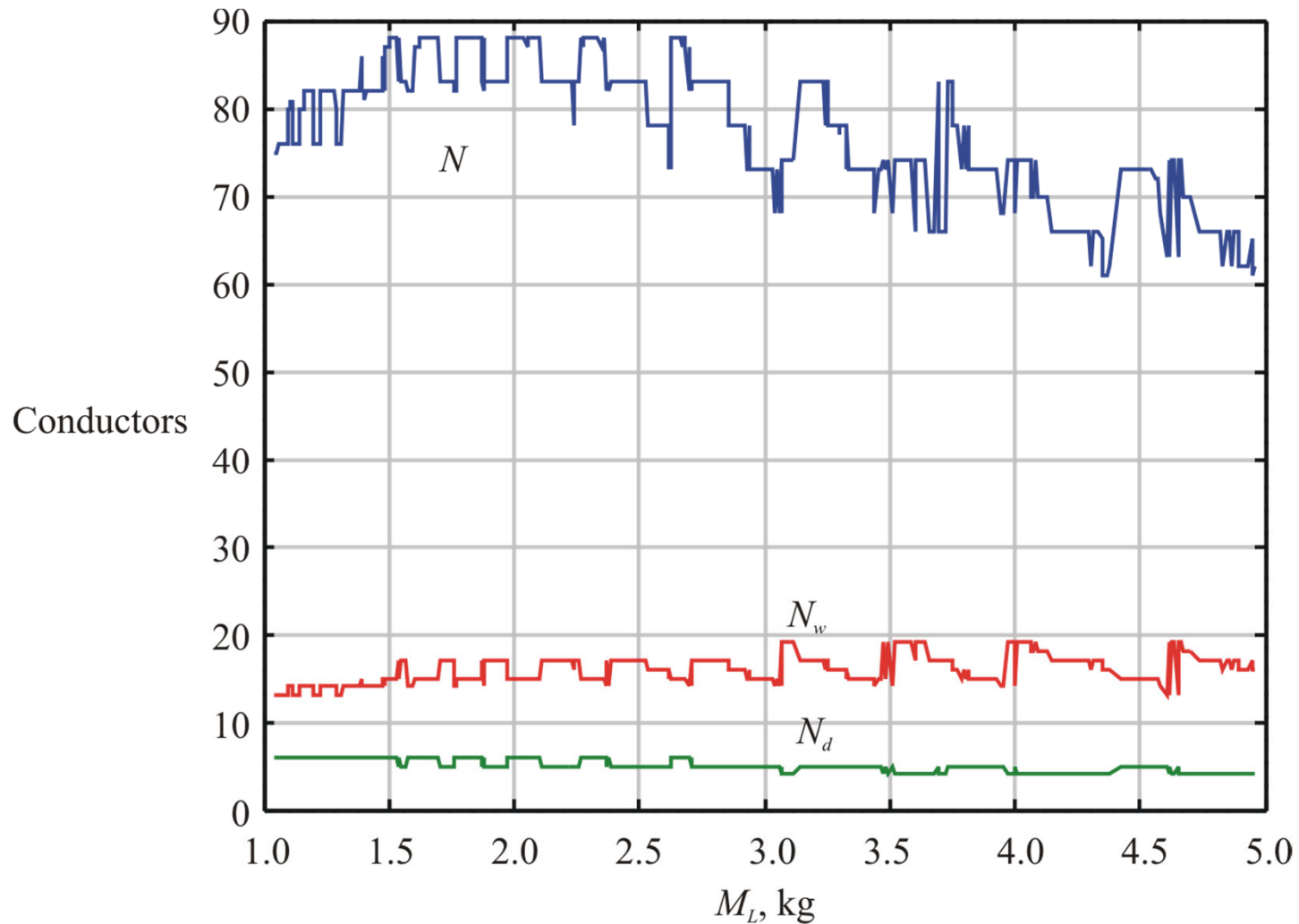
3.4 Case Study

- Current density



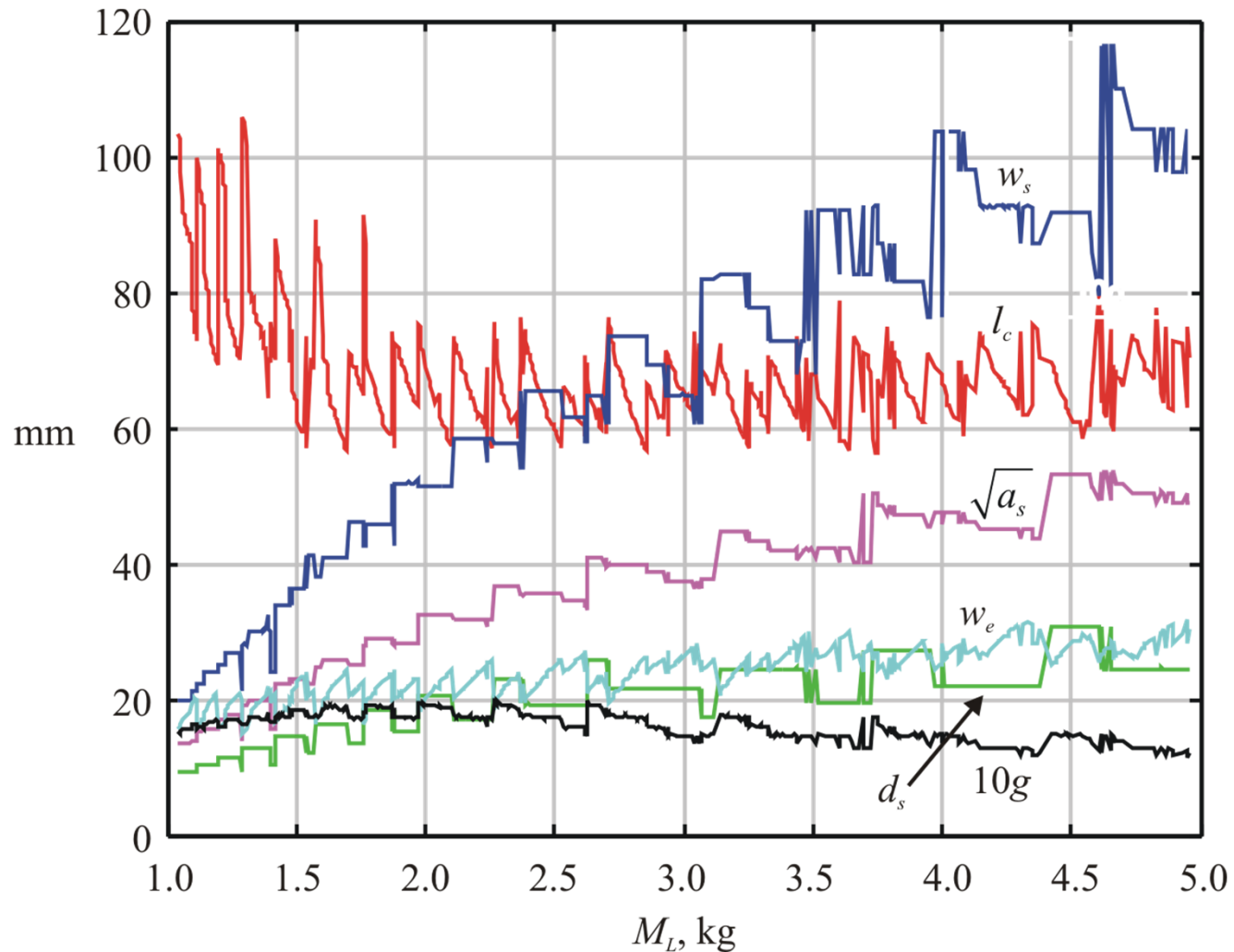
3.4 Case Study

- Conductor count



3.4 Case Study

- Dimensions



3.4 Case Study

- Design 50

Table 3.4-3 Design 50 data.

| <u>UI Core Data</u> | <u>Winding Data</u> | <u>Metrics</u> |
|---------------------|------------------------------|------------------------------|
| Material = MN60LL | Material = Aluminum | $M_L = 1.14$ kg |
| $w_e = 1.72$ cm | AWG = 14 | $P_L = 24.5$ W |
| $w_i = 1.67$ cm | $a_c = 2.08$ mm ² | $J = 4.80$ A/mm ² |
| $w_b = 1.82$ cm | $N = 76$ | $L_{inc} = 5.00$ mH |
| $w_s = 2.22$ cm | $N_w = 13$ | $R_L = 245$ mΩ |
| $d_s = 1.03$ cm | $N_d = 6$ | $h_L = 5.69$ cm |
| $l_c = 9.24$ cm | $d_w = 1.03$ cm | $w_L = 5.66$ cm |
| $M_{cr} = 1.03$ kg | $w_w = 2.22$ cm | $l_L = 11.3$ cm |
| $g = 1.58$ mm | $M_{cd} = 0.108$ kg | |

