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**Lecture Set 5:**  
**Distributed Windings and Rotating MMF**

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Spring 2021

# About this Lecture Set

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- Reading
  - *Power Magnetic Devices: A Multi-Objective Design Approach*, 8.1-8.5, (Available through IEEE)
- Goal
  - Set the stage to study ac machinery including permanent magnet synchronous machines as well as induction machines

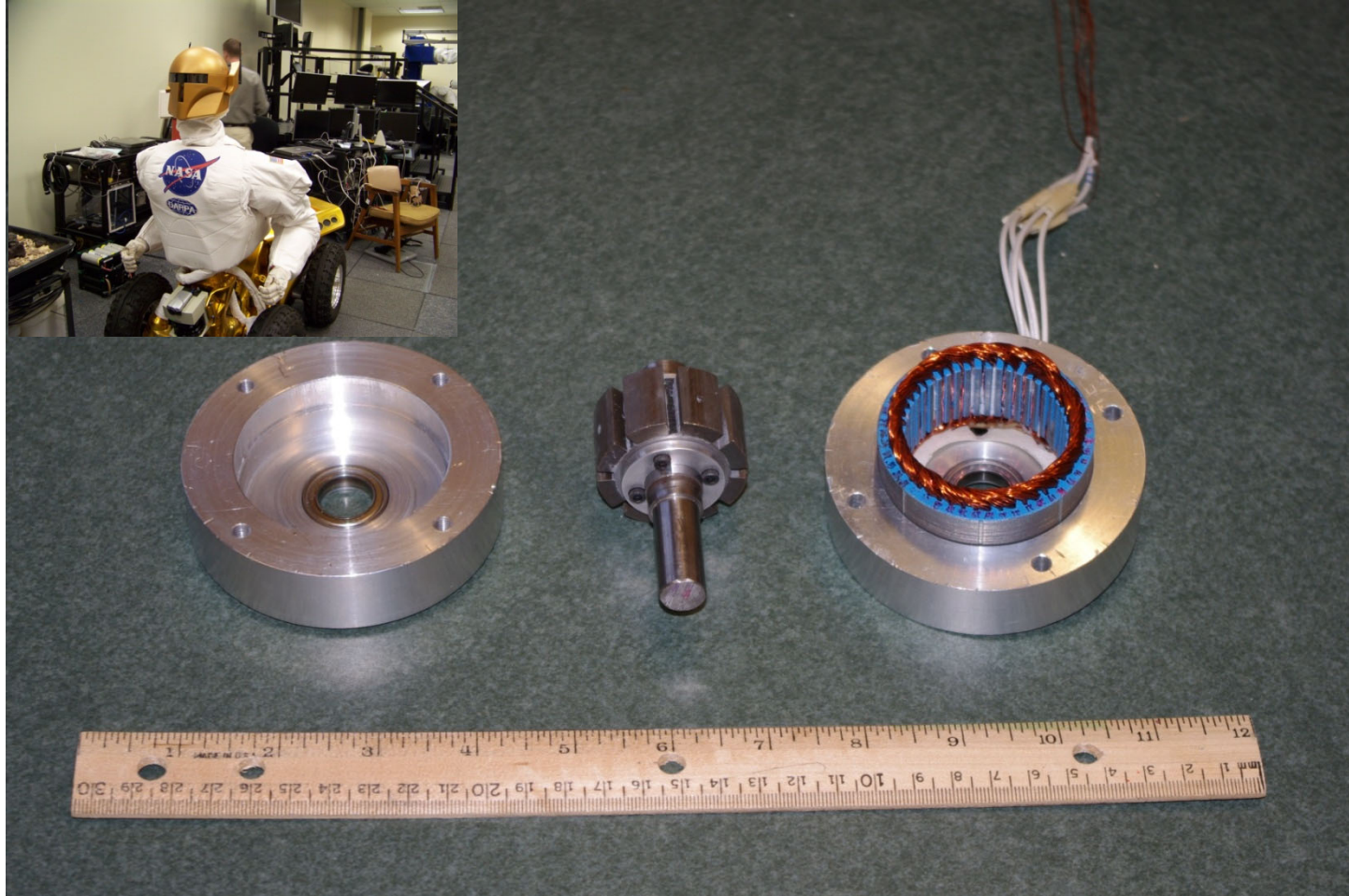
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## *Lecture 42*

# Introduction to AC Machines

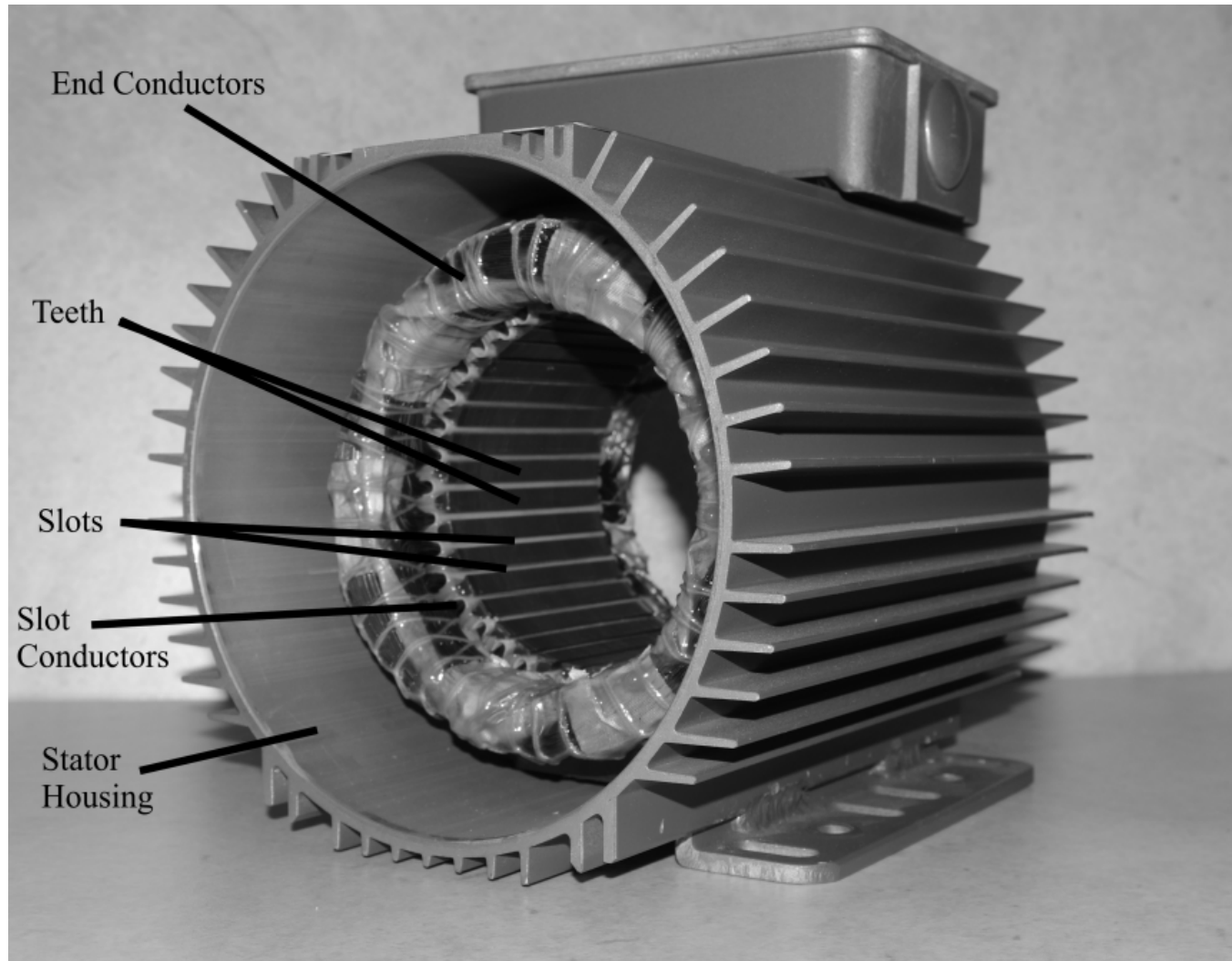
# PM AC Machine

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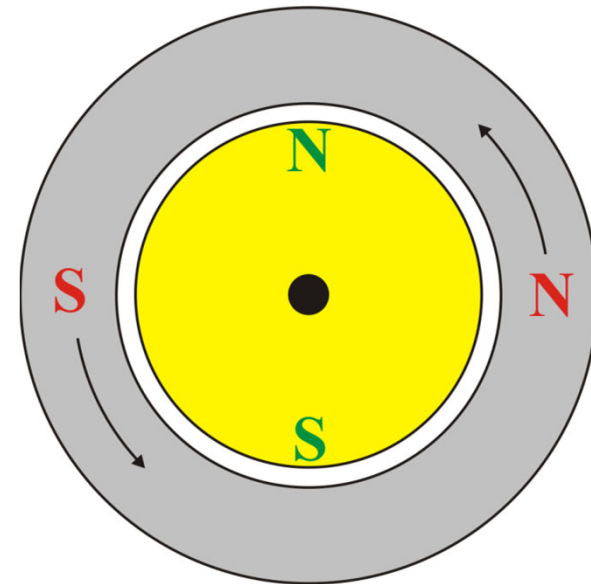
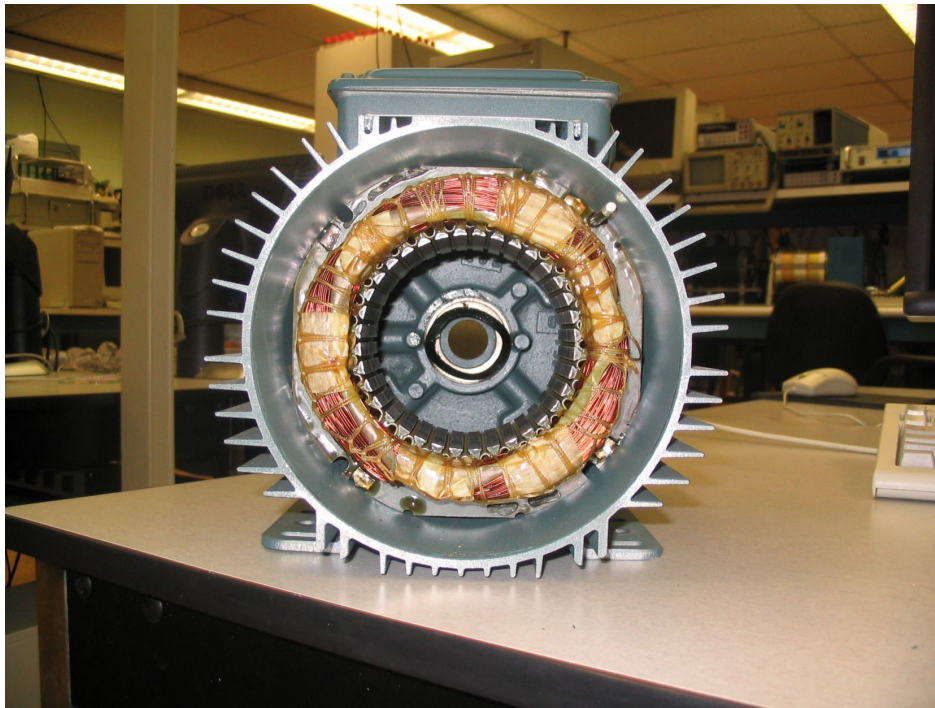
# Induction Machine Stator

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# Operation of AC Machines

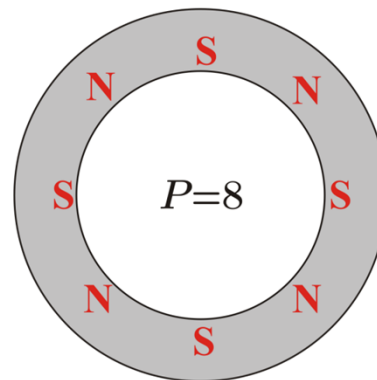
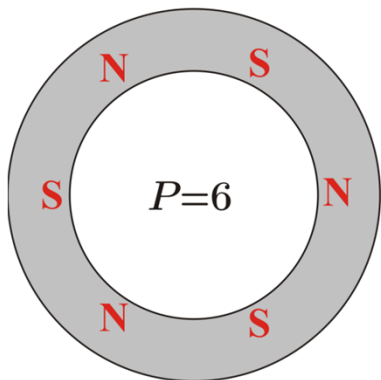
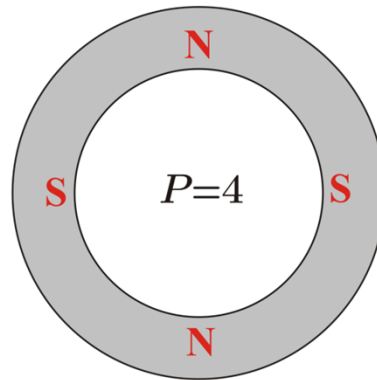
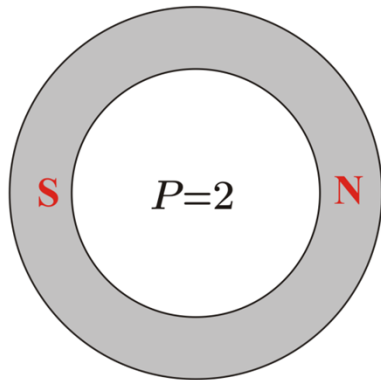
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N-Pole: Flux leaves material  
S-Pole: Flux enters material

# P-Pole Machines

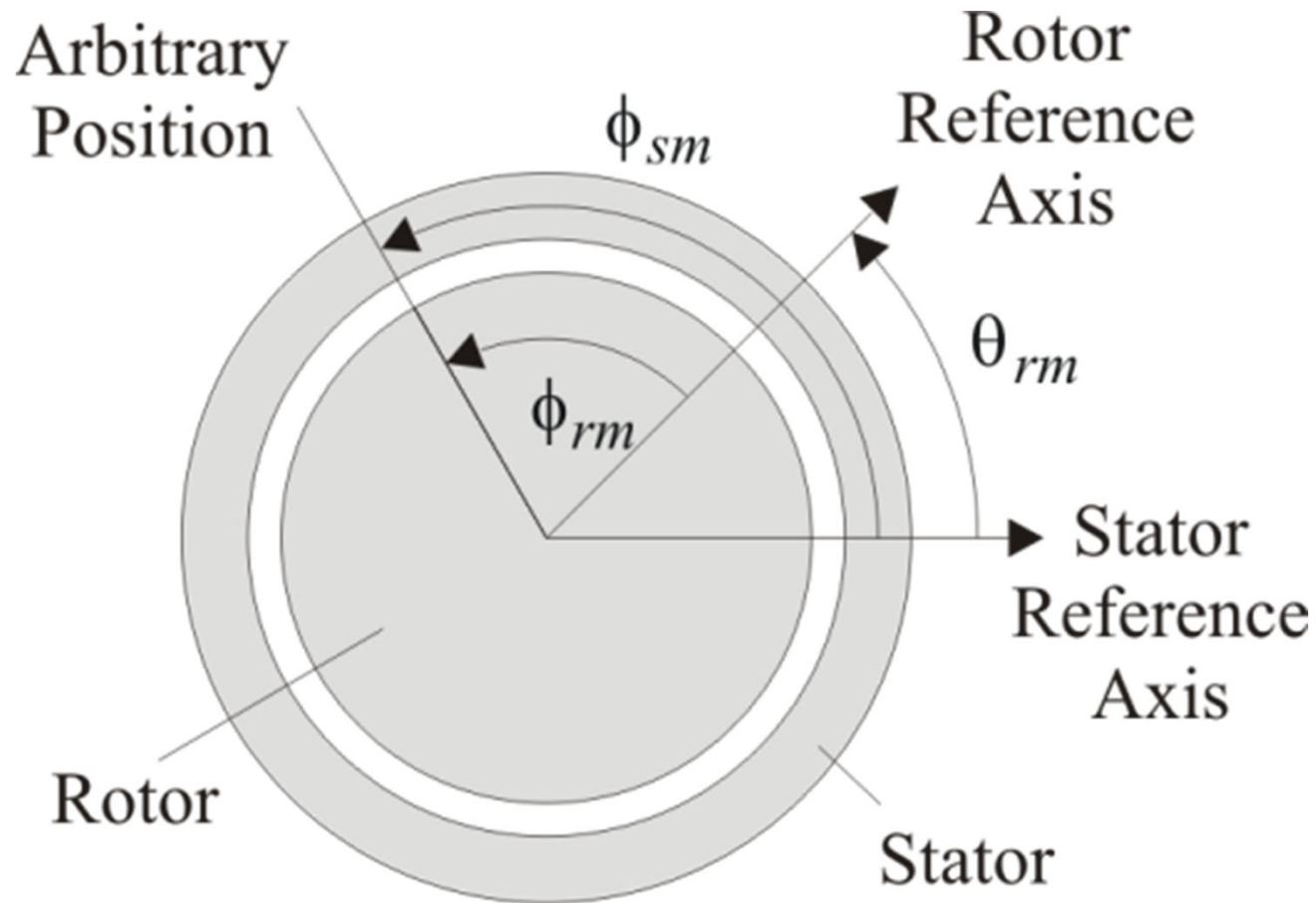
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For given ac frequency having more than 2 poles reduces machine speed by  $P/2$

# Position Measurement

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# Electrical Angles

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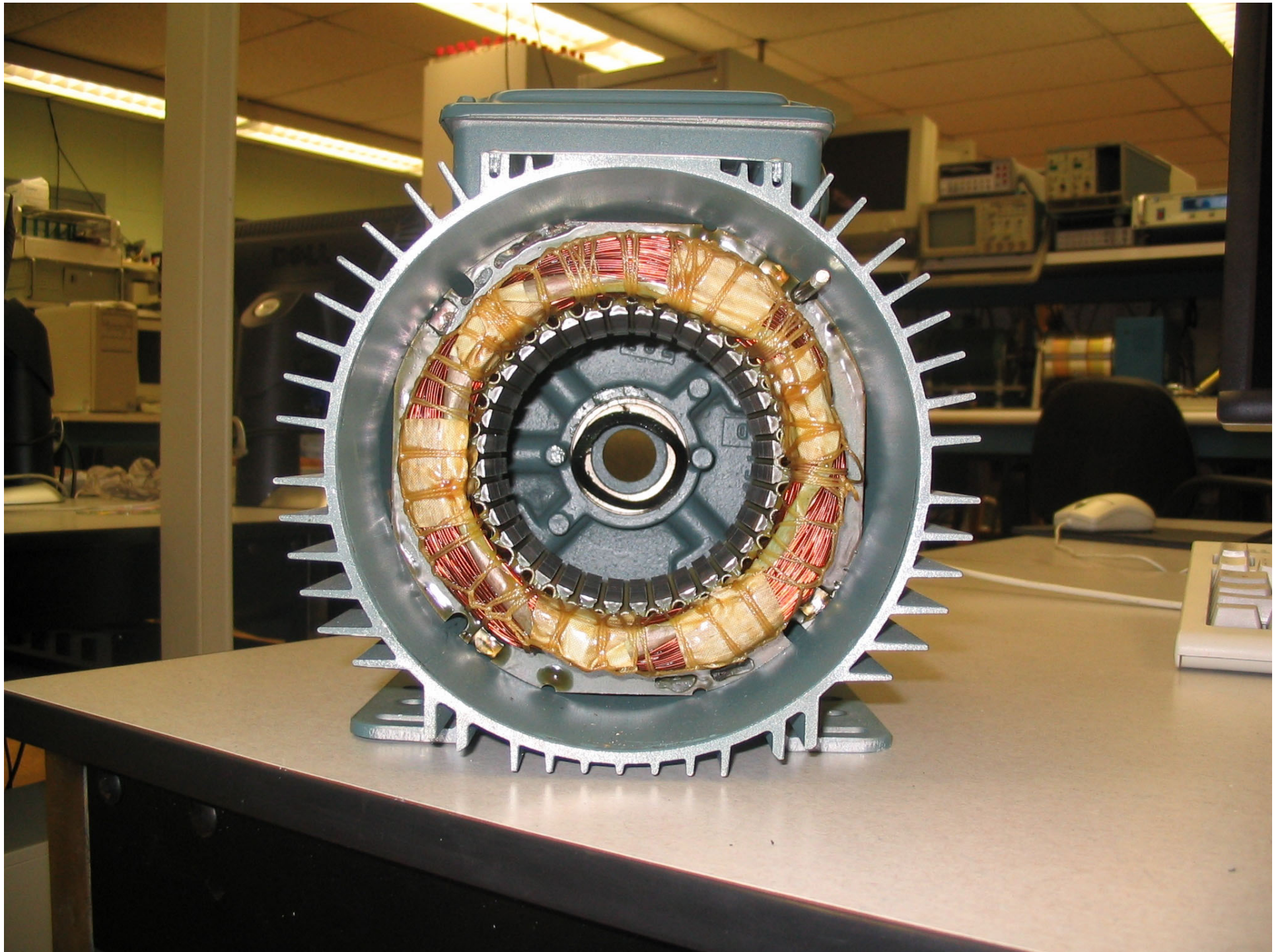
## *Lecture 43*

# Distributed Windings

# Distributed Windings

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- Thus far, we have mostly considered lumped windings. We must now consider distributed winding. This is a more difficult concept.
- Two Representations
  - Discrete Winding Representation
  - Continuous Winding Representation



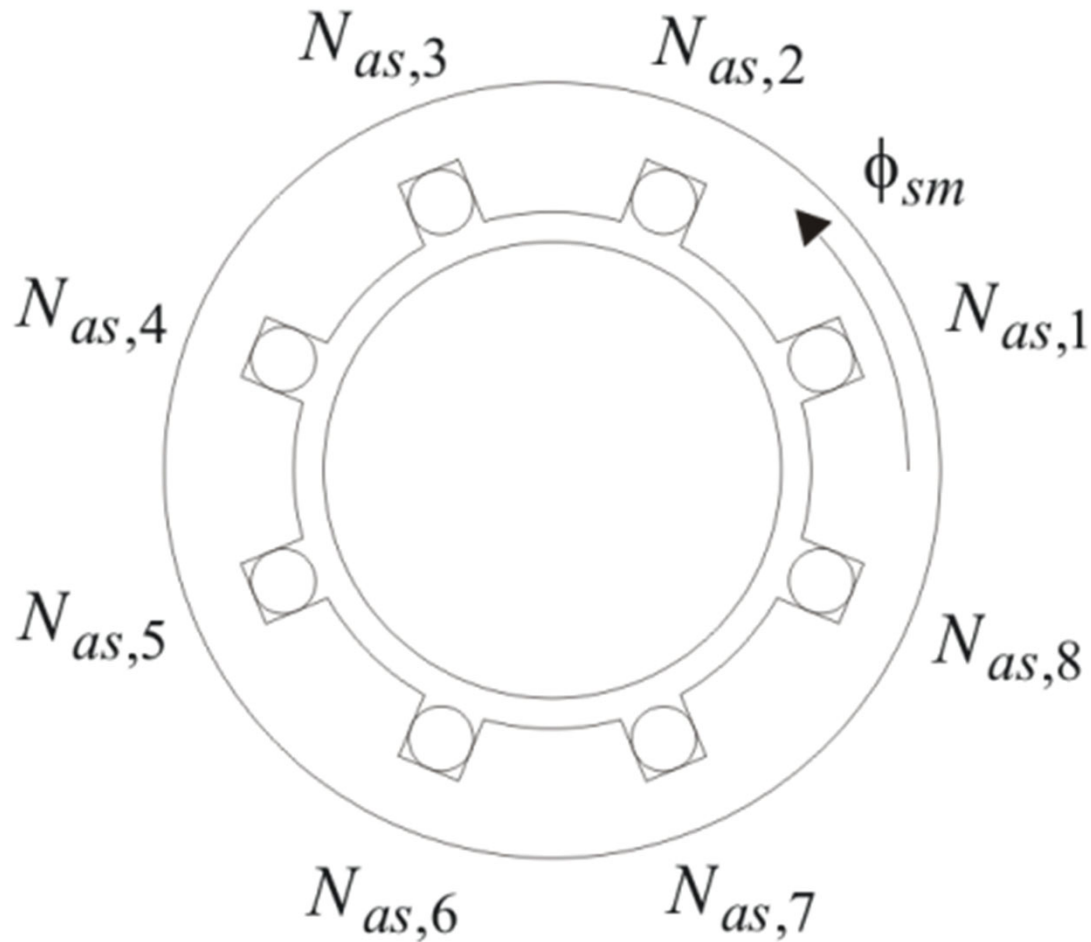
# Discrete Winding Representation

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- Let  $N_{x,i}$  denote number of conductors of a x-phase winding in the i'th slot
  - ‘i’ designates the slot number
  - ‘x’ denotes winding or phase and is typically ‘as’, ‘bs’, or ‘cs’
  - ‘y’ is ‘s’ for stator or ‘r’ for rotor
  - A positive value indicates the conductor directed out of the page

# Slot and Tooth Locations

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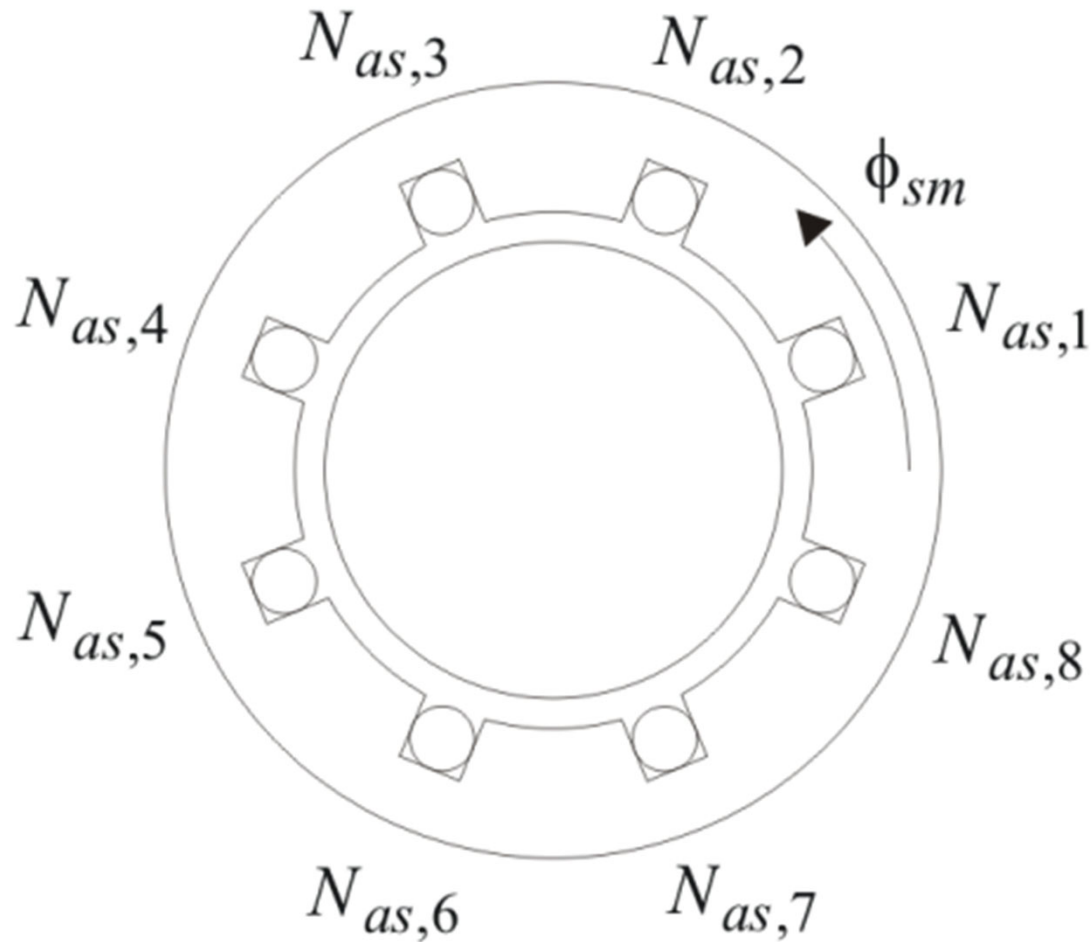


$$\phi_{ys,i} = \pi(2i - 2) / S_y + \phi_{ys,1}$$

$$\phi_{yt,i} = \pi(2i - 3) / S_y + \phi_{ys,1}$$

# Discrete Winding Representation

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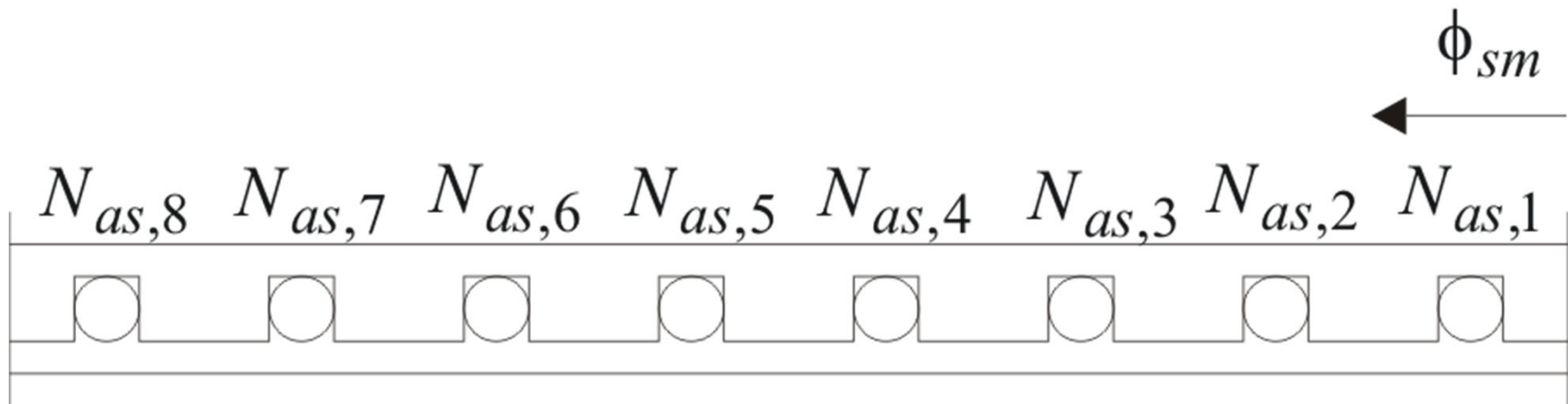


$$\sum_{i=1}^{S_y} N_{x,i} = 0$$

$$N_x = \sum_{i=1}^{S_y} N_{x,i} \mathbf{u}(N_{x,i})$$

# Developed Diagram

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# Continuous Winding Description

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- Conductor density (conductors/rad):  $n_x(\phi_{ym})$ 
  - ‘x’ denotes winding/phase ‘as’, ‘bs’, or ‘cs’
  - ‘y’ is location (‘s’ for stator, ‘r’ for rotor)
  - positive value indicates conductors out of the page
- For example, the a-phase conductor density may be given by

$$n_{as}(\phi_{sm}) = N_{s1} \sin(P\phi_{sm} / 2) - N_{s3} \sin(3P\phi_{sm} / 2)$$

- Total conductors per phase

$$N_{xy} = \int_0^{2\pi} n_{xy}(\phi) u(n_{xy}(\phi)) d\phi$$

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## *Lecture 44*

# Properties of Distributed Windings

# Symmetry Conditions

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$$N_{x,i+2S_y/P} = N_{x,i}$$

$$N_{x,i+S_y/P} = -N_{x,i}$$

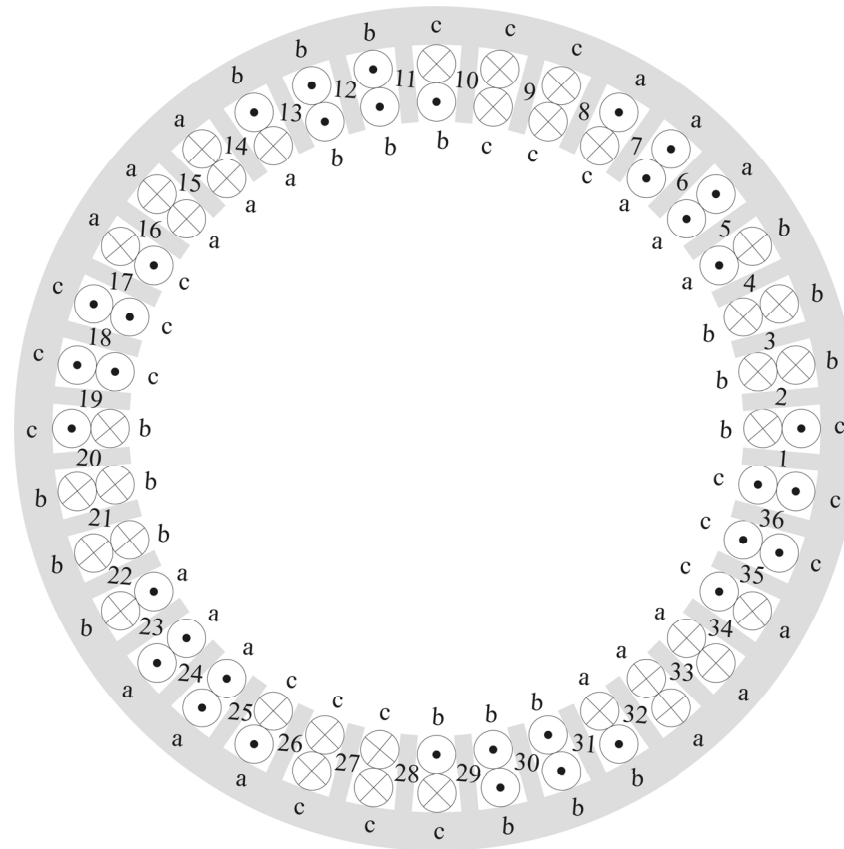
$$\mathbf{n}_x(\phi_m + 4\pi/P) = \mathbf{n}_x(\phi_m)$$

$$\mathbf{n}_x(\phi_m + 2\pi/P) = -\mathbf{n}_x(\phi_m)$$

# Symmetry Conditions

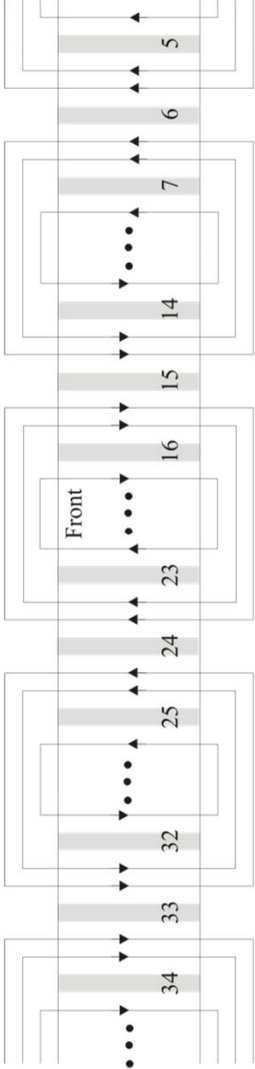
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# Winding Arrangements

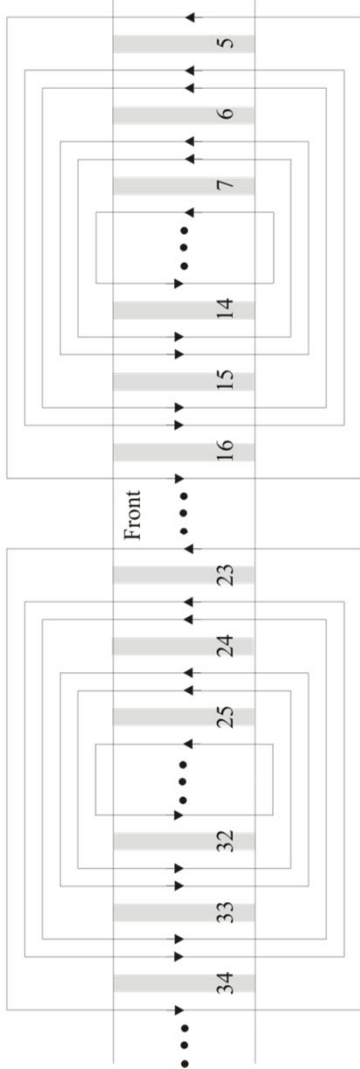


$$\mathbf{N}_{as} \Big|_{1-18} = N[000122100000-1-2-2-100]$$

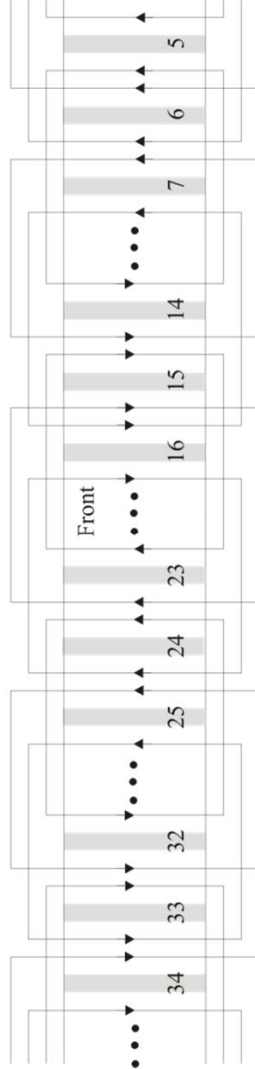
# Winding Arrangements



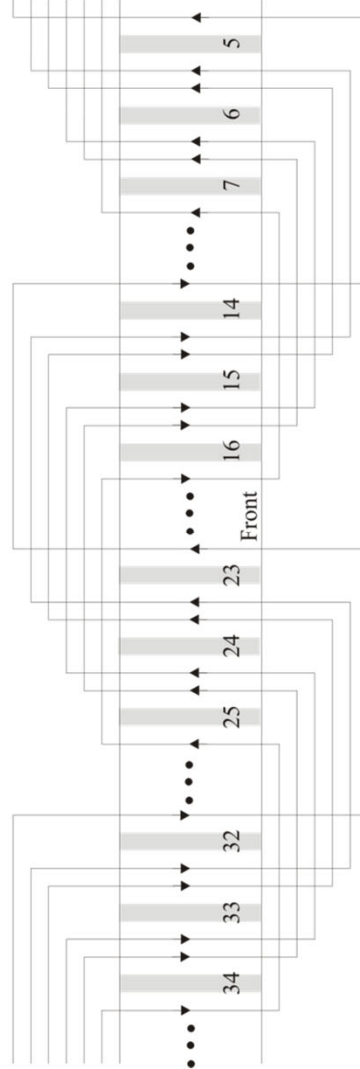
(a) Concentric winding arrangement



(b) Consequent pole winding arrangement



(c) Lap winding arrangement



(d) Wave winding arrangement

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## *Lecture 45*

# The Discrete Winding Function

# Introduction to Winding Functions

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- The winding function is a measure of how many times a winding links flux at a given place
- The winding function will be used to
  - Find self-inductance of a distributed winding
  - Find mutual-inductance between distributed windings
  - Find the mmf associated with a distributed winding
  - It can be expressed in terms of both discrete and continuous winding descriptions



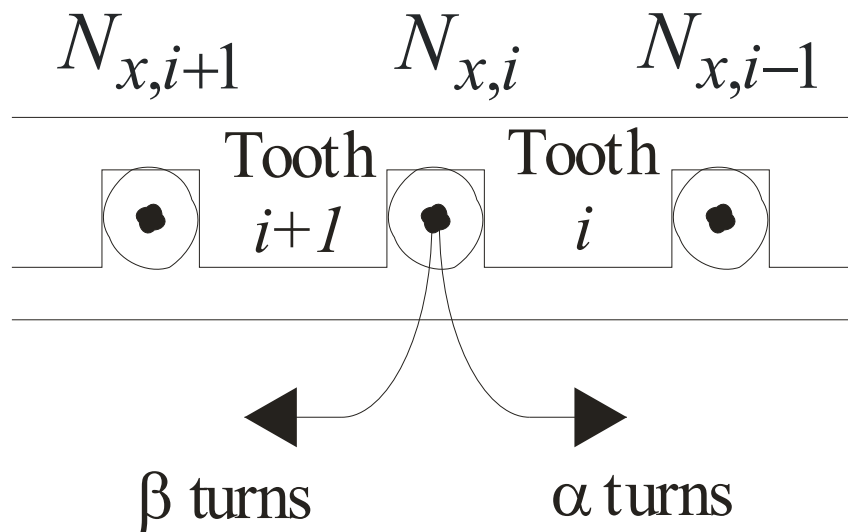
# The Discrete Winding Function

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- $W_{x,i}$  is the discrete winding function of phase 'x' of the stator or rotor
- It is the number of times the winding links flux through the  $i$ 'th tooth
- It is the number of turns around tooth  $i$

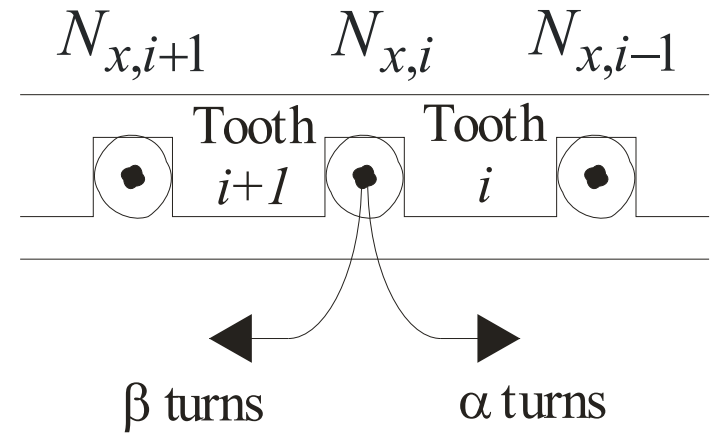
# Derivation of Discrete Winding Function

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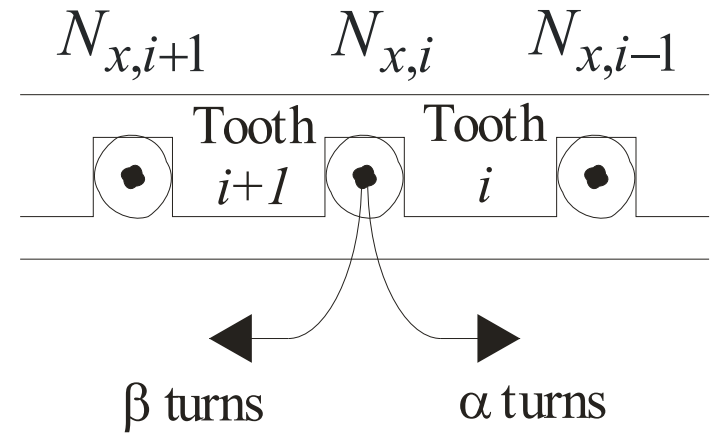
# Derivation of Discrete Winding Function

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# Derivation of Discrete Winding Function

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# Derivation of Discrete Winding Function

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# Derivation of Discrete Winding Function

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- Summarizing:

$$W_{x,1} = \frac{1}{2} \sum_{i=1}^{S_y/P} N_{x,i}$$

$$W_{x,i+1} = W_{x,i} - N_{x,i}$$

# Example

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- Suppose we have

$$\mathbf{N}_x = [10 \quad 20 \quad 10 \quad -10 \quad -20 \quad -10 \quad 10 \quad 20 \quad 10 \quad -10 \quad -20 \quad -10]^T$$

# Example

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- Finally we obtain ...

$$\mathbf{w}_x = [20 \ 10 \ -10 \ -20 \ -10 \ 10 \ 20 \ 10 \ -10 \ -20 \ -10 \ 10]^T$$



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## *Lecture 46*

# The Continuous Winding Function

# Notation

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- $\phi = \phi_{rm}$  for rotor windings
- $\phi = \phi_{sm}$  for stator windings
- $w_x(\phi) =$  winding function
  - Number of times a winding links flux at position  $\phi$

# Derivation of The Cont. Winding Function

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# Derivation of The Cont. Winding Function

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# Derivation of The Cont. Winding Function

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# Derivation of The Cont. Winding Function

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- Thus we have

$$w_x(\phi_m) = \frac{1}{2} \int_0^{2\pi/P} n_x(\phi_m) d\phi_m - \int_0^{\phi_m} n_x(\phi_m) d\phi_m$$

# Example

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- Consider the following geometry

- We might have

$$n_{as} = 100 \sin(\phi_{sm})$$

- Find the winding function

# Example

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# Example

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## *Lecture 47*

# Comparison of Winding Functions

# Example

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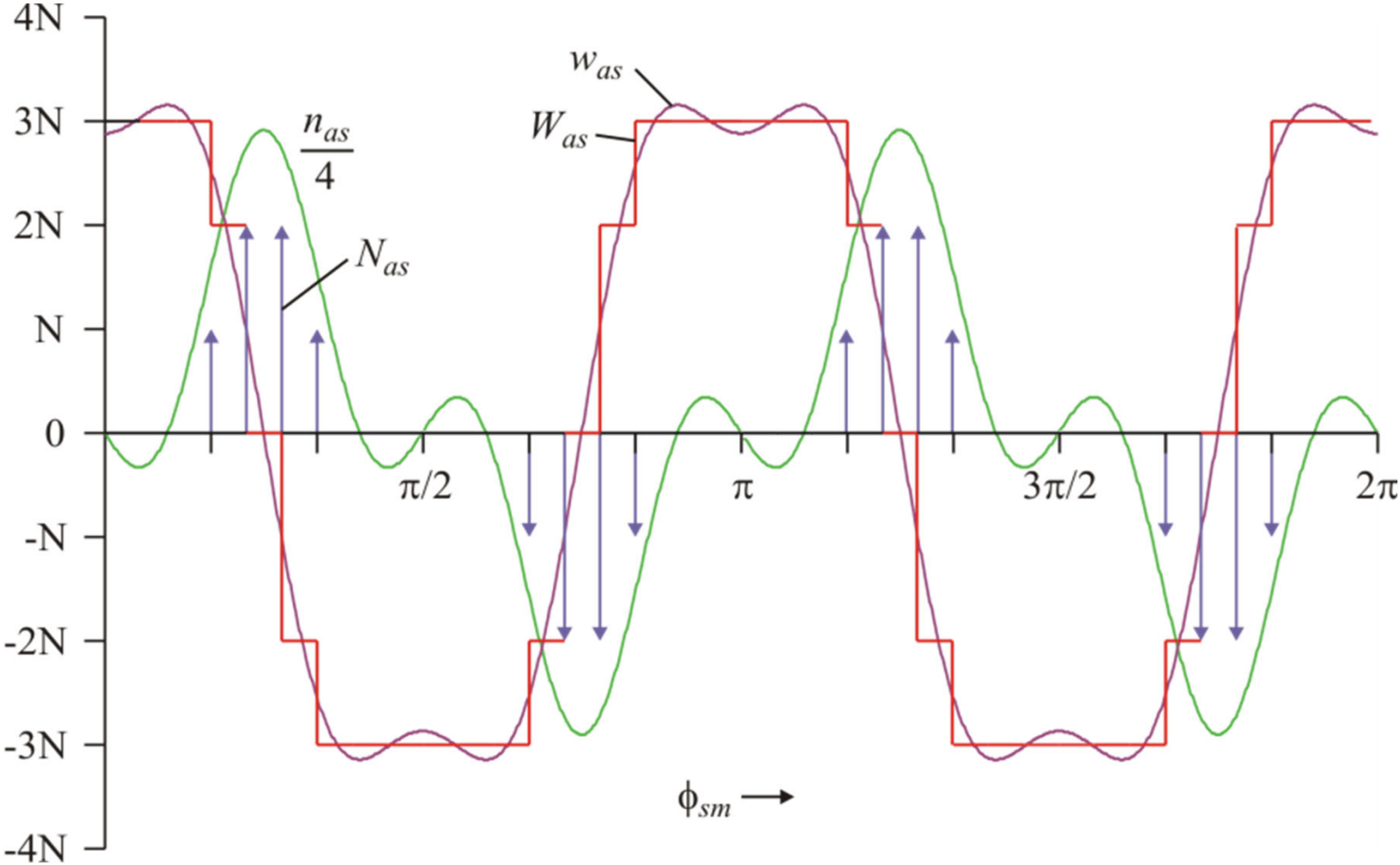
$$\mathbf{N}_{as}|_{1-18} = N [0 \ 0 \ 0 \ 1 \ 2 \ 2 \ 1 \ 0 \ 0 \ 0 \ 0 \ 0 \ -1 \ -2 \ -2 \ -1 \ 0 \ 0]$$

$$\mathbf{W}_{as}|_{1-18} = N [3 \ 3 \ 3 \ 3 \ 2 \ 0 \ -2 \ -3 \ -3 \ -3 \ -3 \ -3 \ -3 \ -3 \ -2 \ 0 \ 2 \ 3 \ 3]$$

$$n_{as} = N (7.221 \sin(2\phi_{sm}) - 4.4106 \sin(6\phi_{sm}))$$

$$w_{as} = N \left( \frac{7.221}{2} \cos(2\phi_{sm}) - \frac{4.4106}{6} \cos(6\phi_{sm}) \right)$$

# Comparison of Winding Functions



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# *Lecture 48*

## Distributed MMF

# The Idea

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- In a distributed machine, the MMF source is associated with a winding is a function of position as well as the current
- We will focus on the continuous winding distribution
- We can show

$$F_S(\phi_m) = \sum_{x \in X_S} w_x(\phi_m) i_x \quad F_R(\phi_m) = \sum_{x \in X_R} w_x(\phi_m) i_x$$

$$F_g(\phi_m) = F_S(\phi_m) + F_R(\phi_m)$$

# Lumped vs Distributed MMF

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# Airgap MMF

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- Before we start, define airgap MMF in a rotating machine:

$$F_g(\phi_m) = \int_{rotor}^{stator} \mathbf{H}(\phi_m) \cdot d\mathbf{l}$$

- Since the integration is in radial direction

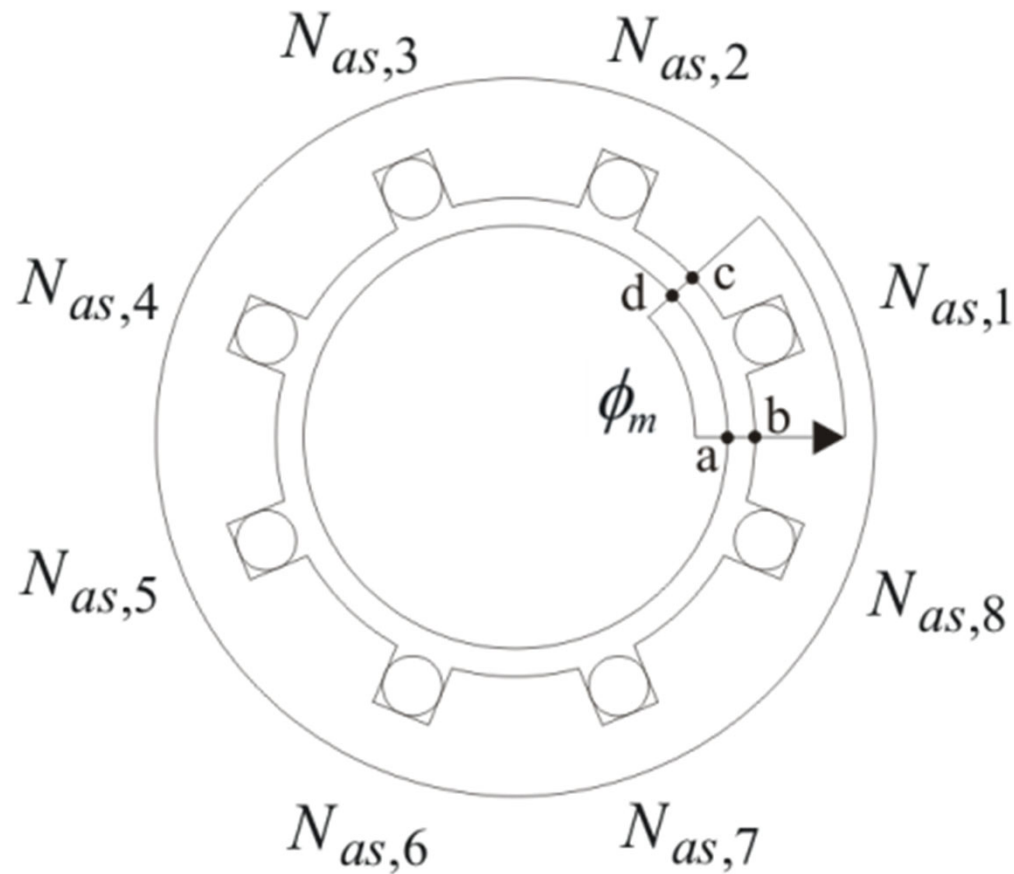
$$F_g(\phi_m) = \int_{rotor}^{stator} H_r(\phi_m) dl$$



# Air Gap MMF

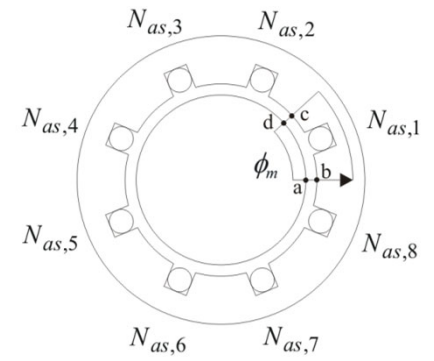
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- Now consider the following



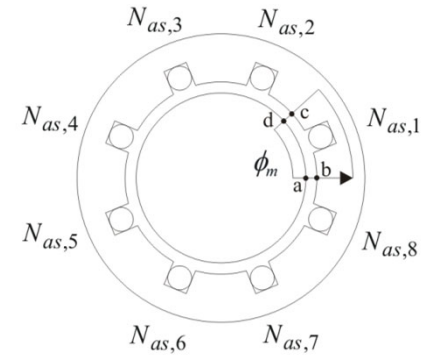
# Air Gap MMF

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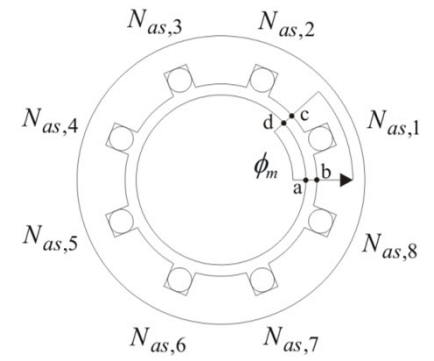
# Air Gap MMF

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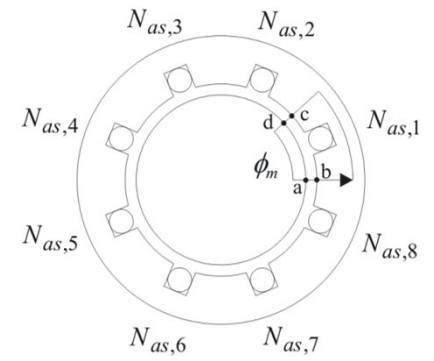
# Air Gap MMF

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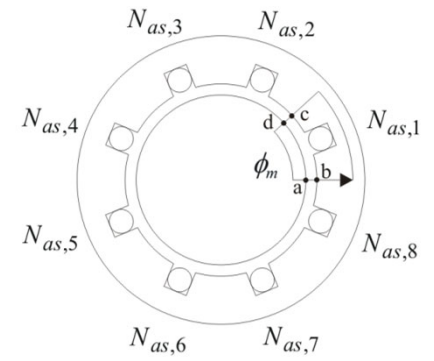
# Air Gap MMF

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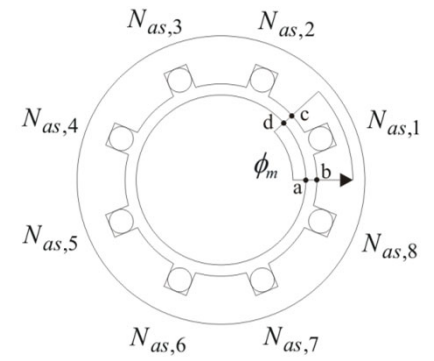
# Air Gap MMF

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# Air Gap MMF

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## *Lecture 49*

# Reasons to Care About Airgap MMF



# Air Gap Flux Density

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- Thus we have

$$B_g(\phi_m) = \frac{\mu_0 F_g(\phi_m)}{g(\phi_m)}$$

# Total Field

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$$F(\phi) = \sum_{x \in X} F_x(\phi)$$

$$B = \frac{\mu_0 F(\phi)}{g(\phi)}$$

# Field Associate With a Winding

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$$B_x = \frac{\mu_0 F_x(\phi)}{g(\phi)}$$

# Example

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- Suppose
  - $w_{as} = 100 \cos(4\phi_{sm}); i_{as} = 5 \text{ A}$
  - $w_{bs} = 50 \sin(4\phi_{sm}); i_{bs} = 10 \text{ A}$
  - $g = 1 \text{ mm}$
- What is the peak B-field ?
- At what position is the peak B-field ?

# Example

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# Example

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# *Lecture 50*

## Rotating MMF

# Rotating MMF

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- Another concept from the MMF is that the fields rotate



# Rotating MMF in Two-Phase Systems

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- Suppose

$$n_{as} = N_s \sin(P\phi_{sm} / 2)$$

$$n_{bs} = -N_s \cos(P\phi_{sm} / 2)$$

$$w_{as} = \frac{2N_s}{P} \cos(P\phi_{sm} / 2)$$

$$w_{bs} = \frac{2N_s}{P} \sin(P\phi_{sm} / 2)$$

- And that

$$i_{as} = \sqrt{2}I_s \cos(\omega_e t + \phi_i)$$

$$i_{bs} = \sqrt{2}I_s \sin(\omega_e t + \phi_i)$$

- What is the MMF ?

# Rotating MMF in Two-Phase Systems

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# Rotating MMF in Two-Phase Systems

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# Rotating MMF in Two-Phase Systems

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- Result

$$F_s = \frac{2\sqrt{2}N_s I_s}{P} \cos\left(\frac{P\phi_{sm}}{2} - \omega_e t - \varphi_i\right)$$

# Affect of Number of Poles

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$$F_s = \frac{2\sqrt{2}N_s I_s}{P} \cos(P\phi_{sm} / 2 - \omega_e t - \varphi_i)$$

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# *Lecture 51*

## Unbalance

# Unbalanced Two-Phase Systems

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- Let's consider another case. Again starting with

$$F_s = \frac{2N_s}{P} \left( \cos(P\phi_{sm} / 2) i_{as} + \sin(P\phi_{sm} / 2) i_{bs} \right)$$

- Suppose the b-phase current is zero; the a-phase current given by

$$i_{as} = \sqrt{2} I_s \cos(\omega_e t + \phi_i)$$

# Unbalanced Two-Phase Systems

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# Unbalanced Two-Phase Systems

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# Unbalanced Two-Phase Systems

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## *Lecture 52*

# Three Phase Systems

# Rotating MMF in Three-Phase Systems

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- In this case suppose

$$n_{as}(\phi_{sm}) = N_{s1} \sin(P\phi_{sm} / 2) - N_{s3} \sin(3P\phi_{sm} / 2)$$

$$n_{bs}(\phi_{sm}) = N_{s1} \sin(P\phi_{sm} / 2 - 2\pi / 3) - N_{s3} \sin(3P\phi_{sm} / 2)$$

$$n_{cs}(\phi_{sm}) = N_{s1} \sin(P\phi_{sm} / 2 + 2\pi / 3) - N_{s3} \sin(3P\phi_{sm} / 2)$$

- Question: why would I do this ?

# Rotating MMF in Three-Phase Systems

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- Proceeding, we have

$$w_{as}(\phi_{sm}) = \frac{2N_{s1}}{P} \cos(P\phi_{sm} / 2) - \frac{2N_{s3}}{3P} \cos(3P\phi_{sm} / 2)$$

$$w_{bs}(\phi_{sm}) = \frac{2N_{s1}}{P} \cos(P\phi_{sm} / 2 - 2\pi / 3) - \frac{2N_{s3}}{3P} \cos(3P\phi_{sm} / 2)$$

$$w_{cs}(\phi_{sm}) = \frac{2N_{s1}}{P} \cos(P\phi_{sm} / 2 + 2\pi / 3) - \frac{2N_{s3}}{3P} \cos(3P\phi_{sm} / 2)$$

- Now

$$F_S(\phi_{sm}) = w_{as}(\phi_{sm})i_{as} + w_{bs}(\phi_{sm})i_{bs} + w_{cs}(\phi_{sm})i_{cs}$$

# Rotating MMF in Three-Phase Systems

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- Now let us assume a three-phase balanced set of currents

$$i_{as} = \sqrt{2}I_s \cos(\omega_e t + \phi_i)$$

$$i_{bs} = \sqrt{2}I_s \cos(\omega_e t + \phi_i - 2\pi/3)$$

$$i_{cs} = \sqrt{2}I_s \cos(\omega_e t + \phi_i + 2\pi/3)$$

# Rotating MMF in Three-Phase Systems

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- Combining yields

$$F_S = \frac{3\sqrt{2}N_{s1}I_s}{P} \cos(P\phi_{sm} / 2 - \omega_e t - \phi_i)$$

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## *Lecture 53*

# Flux Linkage and Inductance



# Flux Linkage and Magnetizing Inductance

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- Flux linkage:

$$\lambda_x = \lambda_{xl} + \lambda_{xm}$$

- Inductance

$$L_{xy} = \frac{\lambda_x |_{\text{due to } i_y}}{i_y}$$

# Flux Linkage and Magnetizing Inductance

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- Partitioning Inductance

$$L_{xy} = L_{xyl} + L_{xym}$$

- Inductance Terms

$$L_{xyl} = \frac{\lambda_{xl} \big|_{\text{due to } i_y}}{i_y}$$

$$L_{xym} = \frac{\lambda_{xm} \big|_{\text{due to } i_y}}{i_y}$$

# Flux Linkage and Magnetizing Inductance

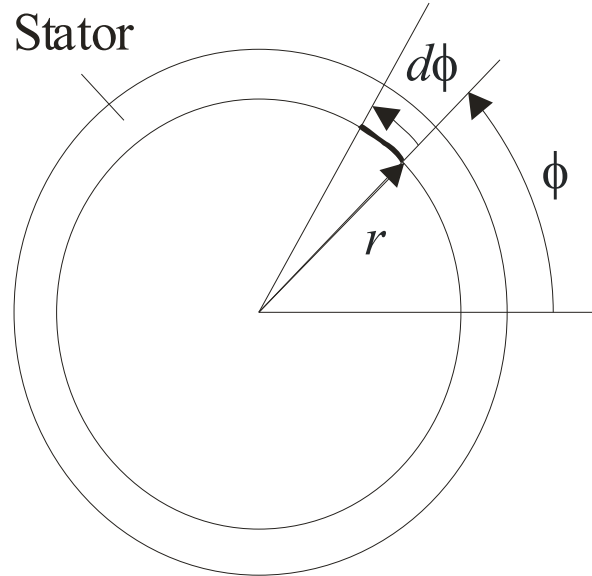
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- We can show

$$\frac{\lambda_{xym}}{i_y} = L_{xym} = \mu_0 r l \int_0^{2\pi} \frac{w_x(\phi_m) w_y(\phi_m)}{g(\phi_m)} d\phi_m$$

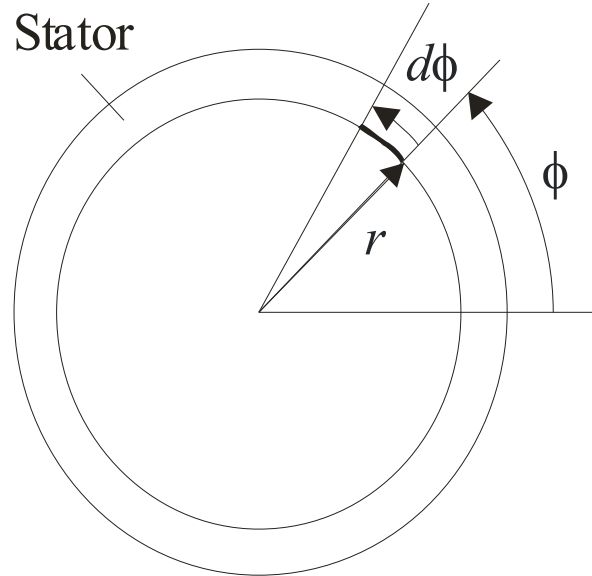
# Flux Linkage and Magnetizing Inductance

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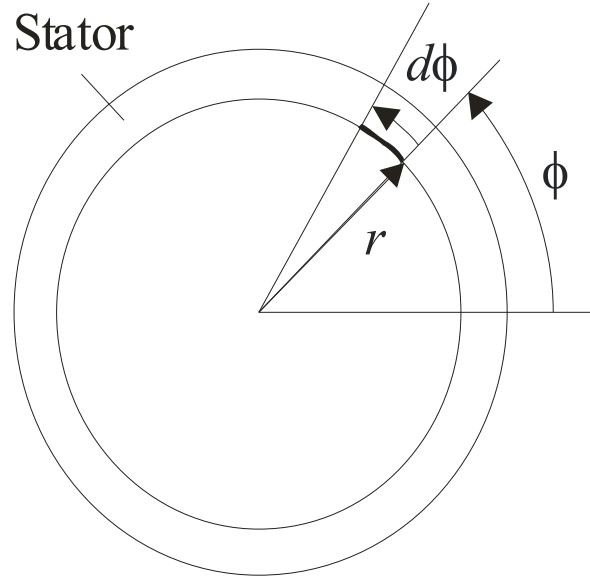
# Flux Linkage and Magnetizing Inductance

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# Flux Linkage and Magnetizing Inductance

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## *Lecture 54*

### Example Inductance Calculations

# Example 1

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- Suppose

$$n_{as} = N_s \sin(P\phi_{sm} / 2)$$

$$w_{as} = \frac{2N_s}{P} \cos(P\phi_{sm} / 2)$$

$$n_{bs} = N_s \sin(P\phi_{sm} / 2 - 2\pi/3)$$

$$w_{bs} = \frac{2N_s}{P} \cos(P\phi_{sm} / 2 - 2\pi/3)$$

and that the air gap is uniform. Find the magnetizing inductance between the two phases



# Example 1

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- We obtain

$$L_{asbs} = -\frac{2\pi\mu_0 r L N_s^2}{P^2 g}$$

## Example 2 (IM)

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$$n_{as}(\phi_{sm}) = N_{s1} \sin(P\phi_{sm} / 2)$$

$$n_{ar}(\phi_{rm}) = N_{r1} \sin(P\phi_{rm} / 2)$$

## Example 2 (Cont.)

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## Example 2 (Cont.)

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## Example 2 (Cont.)

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