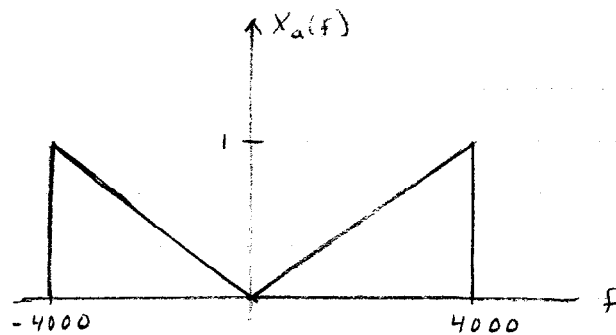


SPRING 2001

EE 438Assignment #3Solutions

1)



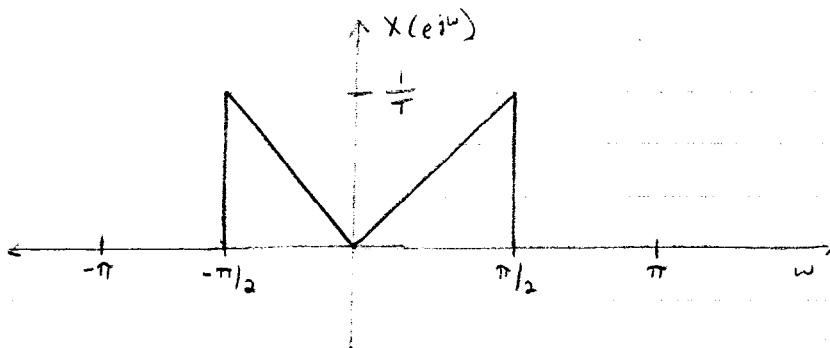
$$F_s = 16 \text{ kHz}$$

(a) Sampling frequency goes to 2π in ω domain.

$$F_s = 16 \text{ kHz} \rightarrow 2\pi$$

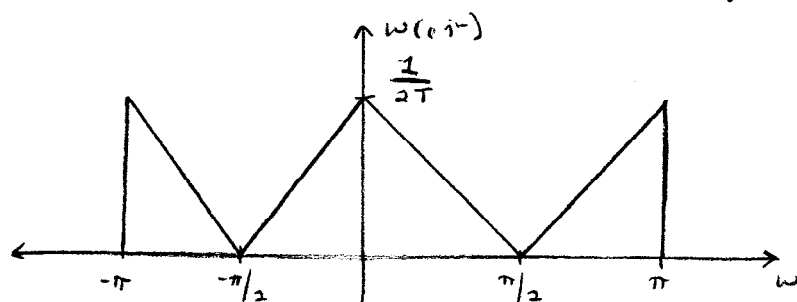
$$\therefore \text{CUTOFF}(X(e^{j\omega})) = \frac{4 \text{ kHz}}{16 \text{ kHz}} (2\pi) = \pi/2$$

AMPLITUDE IS SCALED BY $f_s = 1/T$

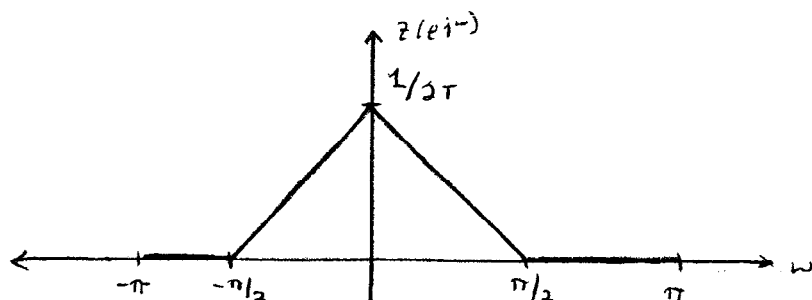


$$(b) \quad w(n) = \cos(\pi/2 n) x(n)$$

$$\begin{aligned} W(e^{j\omega}) &= \pi \{ \delta(\omega - \pi/2) + \delta(\omega + \pi/2) \} * X(e^{j\omega}) \\ &= 1/2 \{ X(e^{j(\omega - \pi/2)}) + X(e^{j(\omega + \pi/2)}) \} \end{aligned}$$

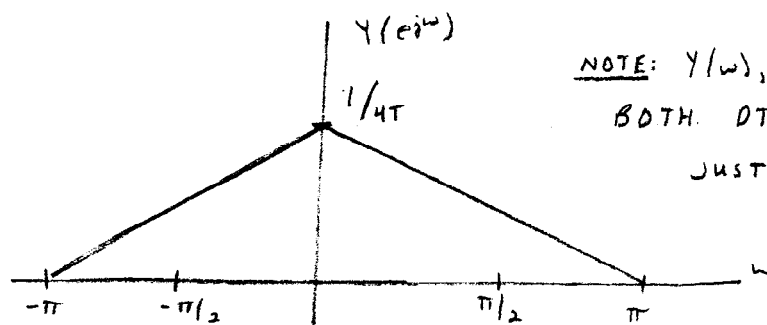


$$Z(e^{j\omega}) = H_{LP}(e^{j\omega}) W(e^{j\omega}) = \begin{cases} W(e^{j\omega}) & -\pi/2 \leq \omega \leq \pi/2 \\ 0, & \text{else.} \end{cases}$$



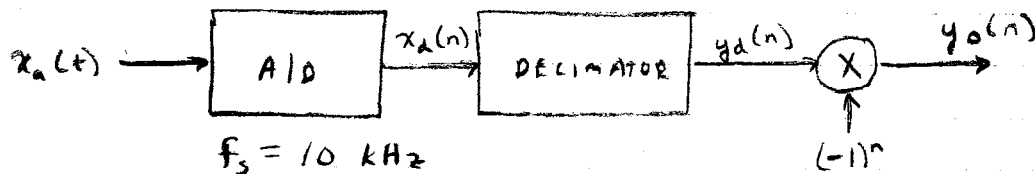
$$\text{DECIMATOR: } Y(\omega) = \frac{1}{D} \sum_{k=0}^{D-1} Z\left(\frac{\omega + 2\pi k}{D}\right) \quad k=0, 1, \dots, (D-1)$$

$$\text{with } D=2, \quad Y(e^{j\omega}) = \frac{1}{2} Z(e^{j\omega/2}) + \frac{1}{2} Z(e^{j(\omega - 2\pi)/2})$$



NOTE: $Y(\omega), Y(e^{j\omega})$ ARE
BOTH DFT $\{y(n)\}$,
JUST NOTATION.

$$2) \quad x_a(t) = \cos[2\pi(2000)t] + \cos[2\pi(6000)t]$$



$$(a) \quad f_s \rightarrow 2\pi$$

$$\therefore f_{a1} = 2 \text{ kHz} \rightarrow \frac{2 \text{ kHz}}{10 \text{ kHz}} (2\pi) = 2/5 \pi \text{ rad/sample}$$

$$f_{a2} = 4 \text{ kHz} \rightarrow \frac{4 \text{ kHz}}{10 \text{ kHz}} (2\pi) = 4/5 \pi \text{ rad/sample}$$

$$\text{Also } -2\pi/5, -4\pi/5$$

$$\text{IN RANGE } 0 \leq \omega \leq \pi : 2\pi/5, 4\pi/5$$

$$(b) \quad y_d(n) = x_d(2n)$$

i.e. just taking every other sample,
analogous to $x_a(t)$ sampled at 5 kHz.

$$\therefore f_{a1} = 2 \text{ kHz} \rightarrow \frac{2 \text{ kHz}}{10 \text{ kHz}} (2\pi) = 4\pi/5 \text{ rad/sample}$$

$$f_{a2} = 4 \text{ kHz} \rightarrow \frac{4 \text{ kHz}}{10 \text{ kHz}} (2\pi) - 2\pi = -2\pi/5 \text{ rad/sample}$$

$$\text{Also } -4\pi/5, 2\pi/5$$

$$\text{IN RANGE } 0 \leq \omega \leq \pi : 2\pi/5, 4\pi/5$$

NOTE ALIASING DUE TO THE NYQUIST CRITERIA NOT MET.

4.

$$\begin{aligned}
 (c) \quad y_0(n) &= (-1)^n y_d(n) \quad \text{But } -1 = e^{j\pi} \\
 &= e^{j\pi n} y_d(n) \\
 &= \cos(\pi n) y_d(n) \\
 &= \cos(\pi n) \left\{ \cos\left[\frac{4\pi}{5}n\right] + \cos\left[\frac{2\pi}{5}n\right] \right\} \\
 &= \frac{1}{2} \left\{ \cos\left[\frac{9\pi}{5}n\right] + \cos\left[\frac{\pi}{5}n\right] + \cos\left[\frac{7\pi}{5}n\right] \right. \\
 &\quad \left. + \cos\left[\frac{3\pi}{5}n\right] \right\}
 \end{aligned}$$

NOTE: ABOVE STEP FROM $\cos A \cos B = \frac{1}{2} \{ \cos(A+B) + \cos(A-B) \}$

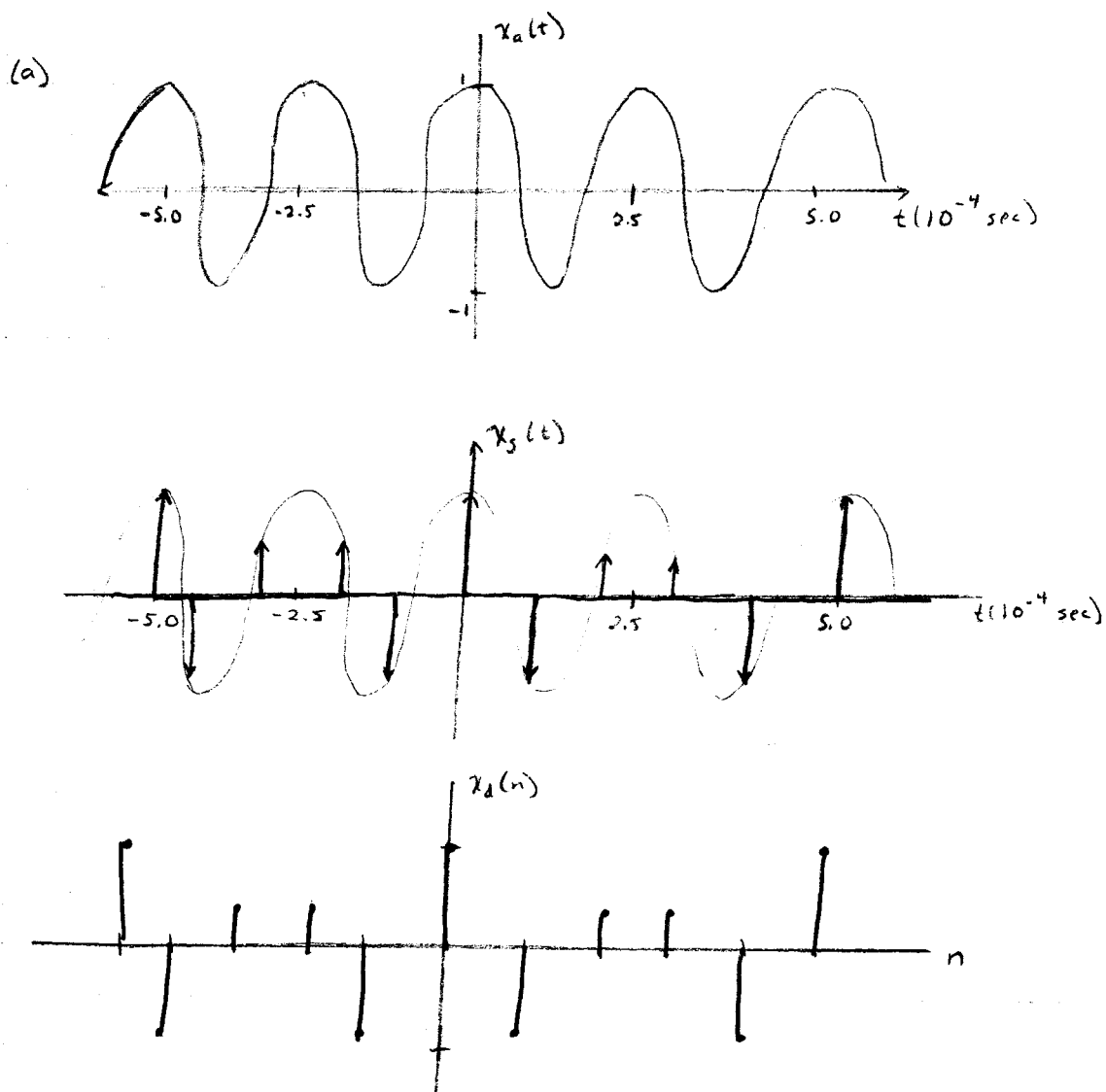
AS IN PART (b), ALIASING HAS OCCURRED $\rightarrow \omega_d > \pi$
 ARE ALIASED BACK INTO $(-\pi, \pi)$ WITH

$$\hat{\omega}_d = |\omega_d - 2\pi|$$

$$\text{GIVING: } \pm \pi/5, \pm 3\pi/5$$

$$\text{IN RANGE } 0 \leq \omega_d \leq \pi : \pi/5, 3\pi/5$$

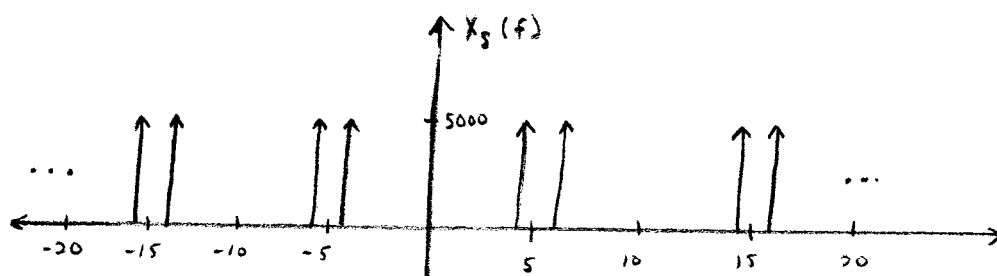
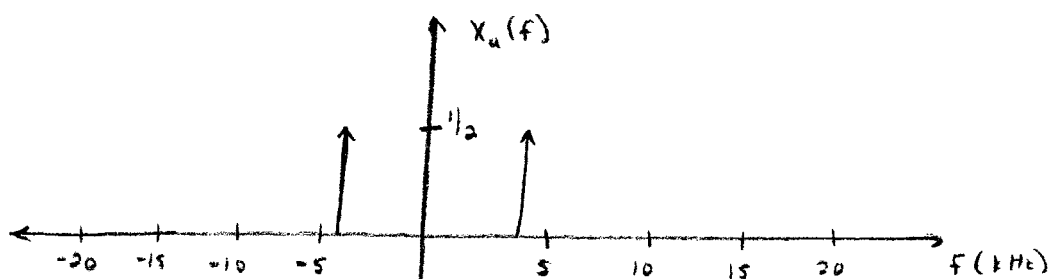
3) $x_a(t) = \cos(2\pi 4000 t)$ $f_s = 10 \text{ kHz}$
 $x_s(t) = \text{comb}_T [x_a(t)]$
 $x_d(n) = x_a(nT)$



$$(b) \quad X_a(f) = \frac{1}{2} [\delta(f - 4000) + \delta(f + 4000)]$$

$$X_s(f) = \frac{1}{T} \text{rep}_{\frac{1}{T}} [X_a(f)]$$

$$= 10000 \text{rep}_{10000} \left\{ \frac{1}{2} [\delta(f - 4000) + \delta(f + 4000)] \right\}$$



$$(c) \quad X_d(\omega) = X_s\left(\frac{\omega}{2\pi T}\right)$$

$$X_s(f) = 5000 \text{rep}_{10000} [\delta(f - 4000) + \delta(f + 4000)]$$

$$= 5000 \sum_{k=-\infty}^{\infty} [\delta(f - 4000 - 10000k) + \delta(f + 4000 - 10000k)]$$

$$X_d(\omega) = 5000 \sum_k [\delta\left(\frac{\omega}{2\pi T} - 4000 - 10000k\right) + \delta\left(\frac{\omega}{2\pi T} + 4000 - 10000k\right)]$$

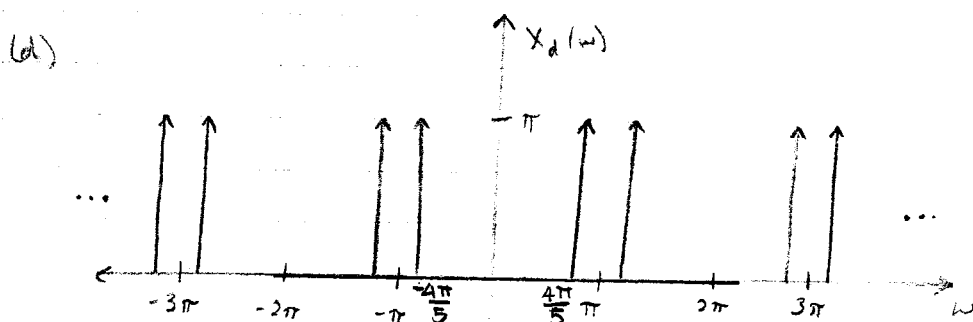
$$= 5000 \sum_k [\delta\left(\frac{10000\omega}{2\pi} - 4000 - 10000k\right) + \delta\left(\frac{10000\omega}{2\pi} + 4000 - 10000k\right)]$$

(next)

FROM HINT: $\delta(A\omega - b) = \frac{1}{|A|} \delta(\omega - \frac{b}{|A|})$

$$X_d(\omega) = 5000 \sum_k \left\{ \frac{2\pi}{10000} \left[\delta\left(\omega - \frac{2\pi}{10000} (10000k + 4000)\right) + \delta\left(\omega - \frac{2\pi}{10000} (10000k - 4000)\right) \right] \right\}$$

$$= \pi \sum_k \left[\delta\left(\omega - 2\pi k - \frac{4\pi}{5}\right) + \delta\left(\omega - 2\pi k + \frac{4\pi}{5}\right) \right]$$



(e) $x_a(t) = \cos(2\pi 4000t)$

$f_s = 10 \text{ kHz}$

$$x_d(n) = x_a\left(\frac{n}{f_s}\right) = \cos\left(2\pi \frac{4000}{10,000} n\right) = \cos\left(\frac{4\pi}{5} n\right)$$

in class, we showed:

$$\cos(\omega_0 n) \xrightarrow{\text{DTFT}} \pi \sum_{k=-\infty}^{\infty} \left\{ \delta(\omega - \omega_0 - 2\pi k) + \delta(\omega + \omega_0 - 2\pi k) \right\}$$

Thus,

$$\cos\left(\frac{4\pi}{5} n\right) \xrightarrow{\text{DTFT}} \pi \sum_{k=-\infty}^{\infty} \left\{ \delta\left(\omega - \frac{4\pi}{5} - 2\pi k\right) + \delta\left(\omega + \frac{4\pi}{5} - 2\pi k\right) \right\}$$

for $|\omega| < \pi$: $\text{DTFT}\left\{\cos\left(\frac{4\pi}{5} n\right)\right\} = \pi \left\{ \delta\left(\omega - \frac{4\pi}{5}\right) + \delta\left(\omega + \frac{4\pi}{5}\right) \right\}$

SAME AS PARTS (c) & (d).

$$4) \quad E\{x(t)\} = 0 \quad (\text{ZERO MEAN})$$

$$E\{x^2(t)\} = \sigma^2 \quad (\text{VARIANCE})$$

For the Gaussian

$$f_x(x) = \frac{1}{\sqrt{2\pi}\sigma} e^{-x^2/2\sigma^2}$$

• desire overload probability < 0.02

• $M \approx 2^B$ levels (actually 2^{B-1})

• assume symmetry $\Rightarrow$ swing from $-\frac{1}{2}2^B = -2^{B-1}$ to $+2^{B-1}$

$$\therefore P\{|x| > 2^{B-1}\Delta\} = 0.02 \quad (\text{NOTE: } 2^B \text{ levels at } \Delta \text{ each})$$

$$\Rightarrow P\{x > 2^{B-1}\Delta\} = 0.01 \quad (\text{by symmetry})$$

$$P\{0 \leq x \leq 2^{B-1}\Delta\} = 0.5 - 0.01 = 0.49$$

$$\Rightarrow \frac{1}{\sigma\sqrt{2\pi}} \int_0^{2^{B-1}\Delta} e^{-t^2/2\sigma^2} dt = 0.49$$

$$\text{Setting } y = t/\sigma$$

$$\frac{1}{\sqrt{2\pi}} \int_0^{\frac{2^{B-1}\Delta}{\sigma}} e^{-y^2/2} dy = 0.49$$

From Gaussian data tables,

$$\frac{2^{B-1}\Delta}{\sigma} = 2.35$$

$$\Delta = 2.35\sigma \cdot 2^{1-B}$$

(next page)

Now, analyzing the noise:

$n_q(t)$ is uniformly distributed between $-\Delta/2$ and $\Delta/2$.

$$E \{ n_q(t) \} = 0 \quad E \{ n_q^2(t) \} = \Delta^2/12$$

JUST ABOUT ANY COMMUNICATIONS
OR SIGNAL PROCESSING BOOK HAS THIS,
ALSO EASY TO SHOW.

$$SNR = 10 \log_{10} \frac{E \{ x^2(t) \}}{E \{ n_q^2(t) \}} = 10 \log_{10} \frac{\sigma^2}{(2.35 \cdot 2^{-8})^2 / 12}$$

$$= 10 \log_{10} \frac{2^{28} \cdot 12}{4 \cdot (2.35)^2} = -2.65 + 68$$

5)

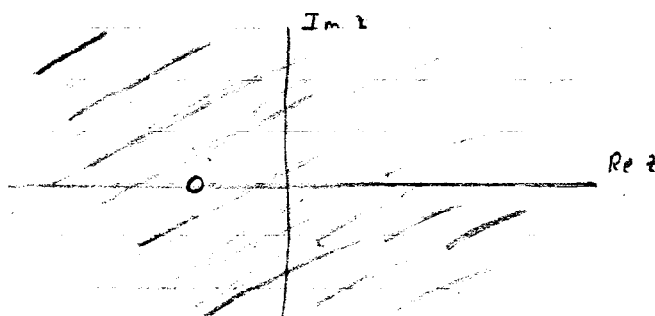
$$(a) \quad x(n) = \delta(n+2) + 2\delta(n+1) + \delta(n)$$

$$X(z) = \sum_{n=0}^2 x(n) z^{-n}$$

$$= z^2 + 2z + 1 \quad \text{ROC: } |z| < \infty$$

POLES: ∞

$$\text{ZEROS: } z^2 + 2z + 1 = (z+1)^2 \Rightarrow z = -1$$



$$(b) \quad x(n) = \delta(n+1) + 2\delta(n) + \delta(n-1)$$

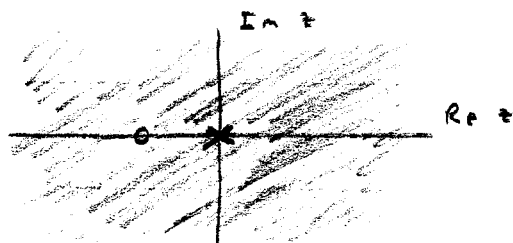
$$X(z) = \sum_{n=-1}^1 x(n) z^{-n} = z + 2 + z^{-1}$$

$$= z^{-1}(z^2 + 2z + 1) = \frac{(z+1)(z+1)}{z}$$

$$\text{ROC: } 0 < |z| < \infty$$

$$\text{POLES: } 0, \infty$$

$$\text{ZEROS: } (z+1) = 0 \Rightarrow z = -1$$



$$(c) \quad x(n) = e^{j\omega_0 n} \quad -\infty < n < \infty$$

$$e^{j\omega_0 n} u(n) \xrightarrow{\mathcal{ZT}} \frac{1}{1 - e^{j\omega_0} z^{-1}}, \quad |z| > |e^{j\omega_0}| = 1 \quad \left. \vphantom{\frac{1}{1 - e^{j\omega_0} z^{-1}}} \right\} \text{outside unit circle}$$

$$e^{j\omega_0 n} u(-n-1) \xrightarrow{\mathcal{ZT}} \frac{-1}{1 - e^{j\omega_0} z^{-1}}, \quad |z| < |e^{j\omega_0}| = 1 \quad \left. \vphantom{\frac{-1}{1 - e^{j\omega_0} z^{-1}}} \right\} \text{inside unit circle}$$

ROC's do not overlap

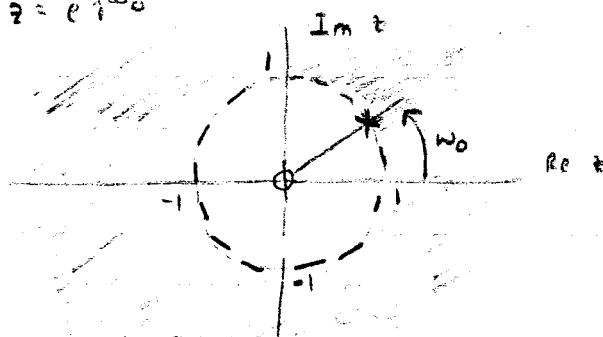
$\Rightarrow X(z)$ does not exist

(d)

$$x(n) = e^{j\omega_0 n} u(n)$$

$$e^{j\omega_0 n} u(n) \xleftrightarrow{zT} \frac{1}{1 - e^{j\omega_0} z^{-1}} \quad |z| > |e^{j\omega_0}| = 1$$

$$X(z) = \frac{z}{z - e^{j\omega_0}}$$

ZERO: $z = 0$ POLE: $z = e^{j\omega_0}$ 

$$\begin{aligned} \text{(e)} \quad x(n) &= \sin(\omega_0 n) u(n) \\ &= \frac{1}{2j} (e^{j\omega_0 n} - e^{-j\omega_0 n}) \end{aligned}$$

$$e^{j\omega_0 n} u(n) \xleftrightarrow{zT} \frac{1}{1 - e^{j\omega_0} z^{-1}} \quad |z| > |e^{j\omega_0}| = 1$$

$$e^{-j\omega_0 n} u(n) \xleftrightarrow{zT} \frac{1}{1 - e^{-j\omega_0} z^{-1}} \quad |z| > |e^{-j\omega_0}| = 1$$

$$X(z) = \frac{1}{2j} \left[\frac{z}{z - e^{j\omega_0}} - \frac{z}{z - e^{-j\omega_0}} \right]$$

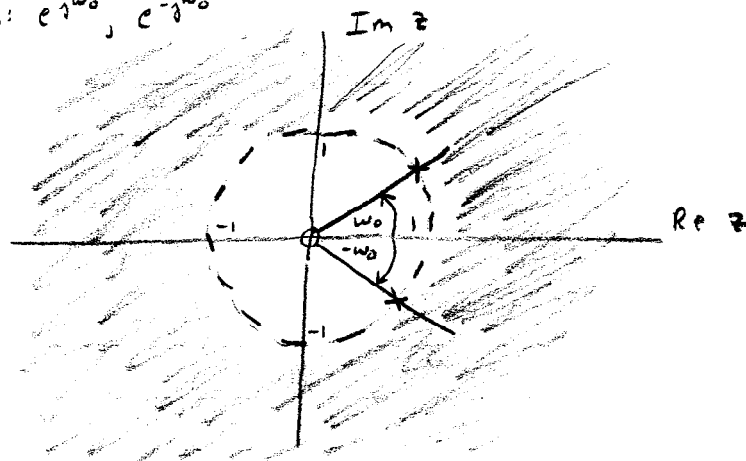
$$= \frac{1}{2j} \left[\frac{z^2 - z e^{-j\omega_0} - z^2 + z e^{j\omega_0}}{z^2 - z e^{-j\omega_0} - z e^{j\omega_0} + 1} \right]$$

$$= \frac{1}{2j} \left[\frac{z(e^{j\omega_0} - e^{-j\omega_0})}{z^2 - 2z \cos \omega_0 + 1} \right] = \frac{z \sin \omega_0}{z^2 - 2z \cos \omega_0 + 1}$$

(next)

$$z \neq 0: z = 0$$

$$\text{POLES: } e^{j\omega_0}, e^{-j\omega_0}$$



f)

$$x(n) = 2^n u(-n) + 4^n u(n)$$

$$4^n u(n) \xleftrightarrow{ZT} \frac{1}{1-4z^{-1}} \quad |z| > 4$$

$$2^n u(-n) = 2^n u(-n-1) + \delta(n) \xleftrightarrow{ZT} \frac{-1}{1-2z^{-1}} + 1 \quad |z| < 2$$

$$\{|z| < 4\} \cap \{|z| < 2\} = \emptyset$$

$\therefore$ No ROC $\Rightarrow$ ZT does not exist

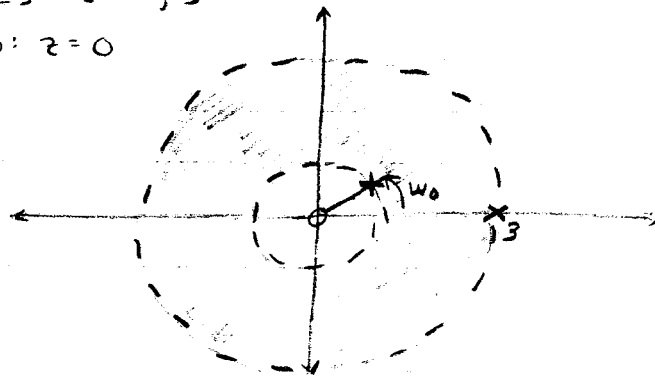
(g)

$$x(n) = -3^n u(-n-1) + e^{j\omega_0 n} u(n)$$

$$-3^n u(-n-1) \xleftrightarrow{zT} \frac{1}{1-3z^{-1}} \quad |z| < 3$$

$$e^{j\omega_0 n} u(n) \xleftrightarrow{zT} \frac{1}{1-e^{j\omega_0} z^{-1}} \quad |z| > 1$$

$$\begin{aligned} X(z) &= \frac{z}{z-3} + \frac{z}{z-e^{j\omega_0}} \\ &= \frac{z^2 - z e^{j\omega_0} + z^2 - 3z}{z^2 - 3z - z e^{j\omega_0} - 3 e^{j\omega_0}} \\ &= \frac{2z^2 - (e^{j\omega_0} + 3)z}{z^2 - (e^{j\omega_0} + 3)z - 3 e^{j\omega_0}} \end{aligned}$$

POLES: $e^{j\omega_0}, 3$ ZERO: $z=0$ (h) USING PROPERTY $nx(n) \xleftrightarrow{zT} -z \frac{dX(z)}{dz}$

$$u[n] \xleftrightarrow{zT} \frac{1}{1-z^{-1}}$$

$$\begin{aligned} n u[n] &\xleftrightarrow{zT} -z \frac{d}{dz} \left(\frac{1}{1-z^{-1}} \right) = -z \left(\frac{-z^{-2}}{(1-z^{-1})^2} \right) \\ &= \frac{z^{-1}}{(1-z^{-1})^2} \quad \text{ROC: } |z| > 1 \end{aligned}$$

(next)

ZERO: $z=0$

POLES: 1

