

ECE-382: Midterm #1

February 26, 2007

8:30--9:30PM

- Do not write your answers at the back of any page. Answers written at the back of page will NOT be graded.
- This is a closed book exam. You are only allowed to bring a pen (or pencil), a one-sided crib sheet, and an eraser to the exam. No calculator is allowed.
- Common Laplace transform pairs are provided at the end of the exam booklet.
- Write your name and PUID at the space provided below.
- This exam has two parts.

Part I consists of three problems. Unless otherwise instructed, justify your answers to these problems completely. **Show all intermediate steps to get full credits for problems from Part I.**

Part II consists of one question, with three sub-questions, for which no justification is required. Enter the answers to Part II in the spaces provided. **Partial credit will not be provided for problems from Part II.**
- The total points are 150. You have one hour to complete the exam.

Solution

Your Name

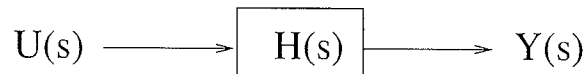
10-digit PUID

Questions for Part I

Justify your answer completely. Show all intermediate steps to get full credits.

Problem 1 (30 points)

Consider the following system



where the transfer function is given by

$$H(s) = \frac{s}{s^2 + s + 1}$$

- (a) (5 points) Assume that the input is $u(t) = e^{-t}1(t)$. What is its Laplace transform $U(s)$?
- (b) (5 points) Find the Laplace transform $Y(s)$ of the corresponding output signal.
- (c) (10 points) Find the output time-response $y(t)$ when the input is $u(t) = e^{-t}1(t)$. Your final answer should not contain imaginary parts.
- (d) (10 points) Find the final value of $y(t)$ as $t \rightarrow +\infty$.

Solution:

$$(a) \quad U(s) = \frac{1}{s+1}$$

$$(b) \quad Y(s) = H(s) U(s) = \frac{s}{(s+1)(s^2+s+1)}$$

$$(c) \quad \text{Let } Y(s) = \frac{a}{s+1} + \frac{bs+c}{s^2+s+1}$$

$$\begin{aligned} \Rightarrow \quad s &= a(s^2+s+1) + (bs+c)(s+1) \\ &= (a+b)s^2 + (a+b+c)s + (a+c) \end{aligned}$$

$$\therefore \begin{cases} a+b=0 \\ a+b+c=1 \\ a+c=0 \end{cases}$$

Solve for a , b , and c , we obtain

$$\begin{cases} a = -1 \\ b = 1 \\ c = 1 \end{cases}$$

$$\therefore Y(s) = \frac{-1}{s+1} + \frac{s+1}{s^2+s+1}$$

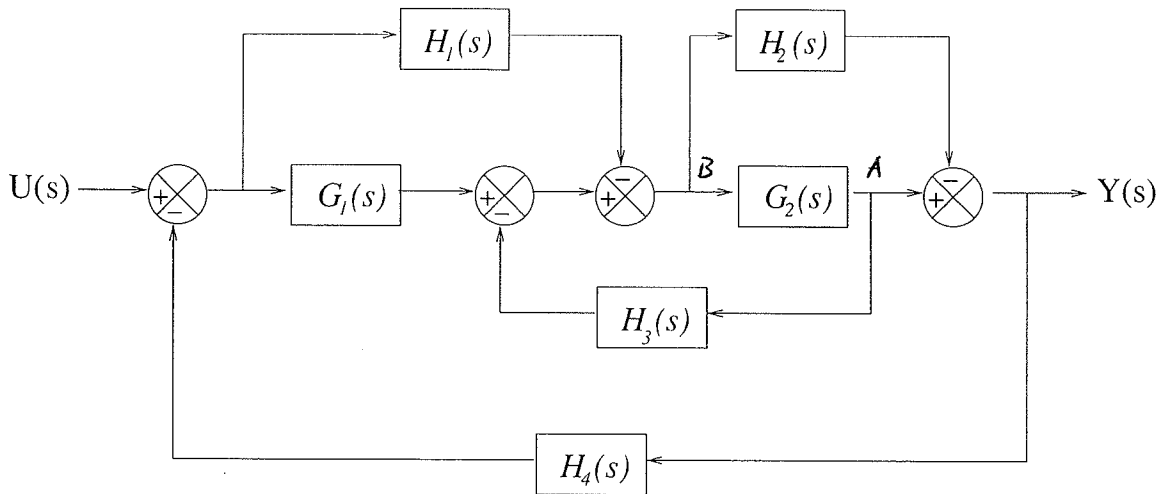
$$= \frac{-1}{s+1} + \frac{s + \frac{1}{2}}{(s + \frac{1}{2})^2 + (\frac{\sqrt{3}}{2})^2} + \frac{\frac{1}{2}}{(s + \frac{1}{2})^2 + (\frac{\sqrt{3}}{2})^2}$$

$$\therefore y(t) = -e^{-t} + e^{-\frac{t}{2}} \cos \frac{\sqrt{3}}{2} t + \frac{1}{\sqrt{3}} e^{-\frac{t}{2}} \sin \frac{\sqrt{3}}{2} t$$

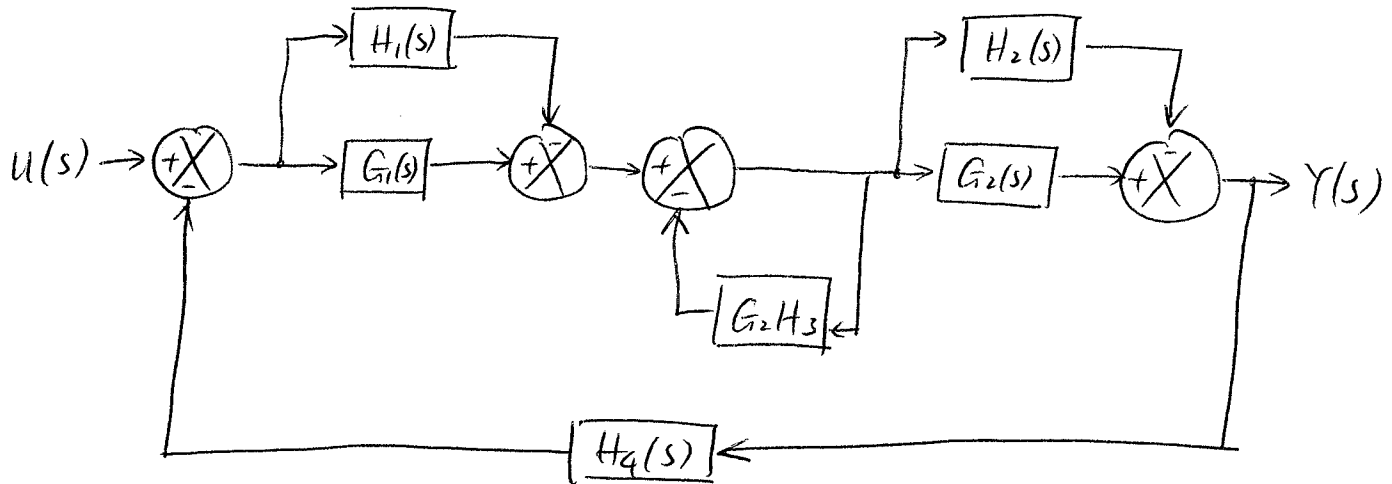
(d) As $t \rightarrow +\infty$, $y(t) \rightarrow 0$.

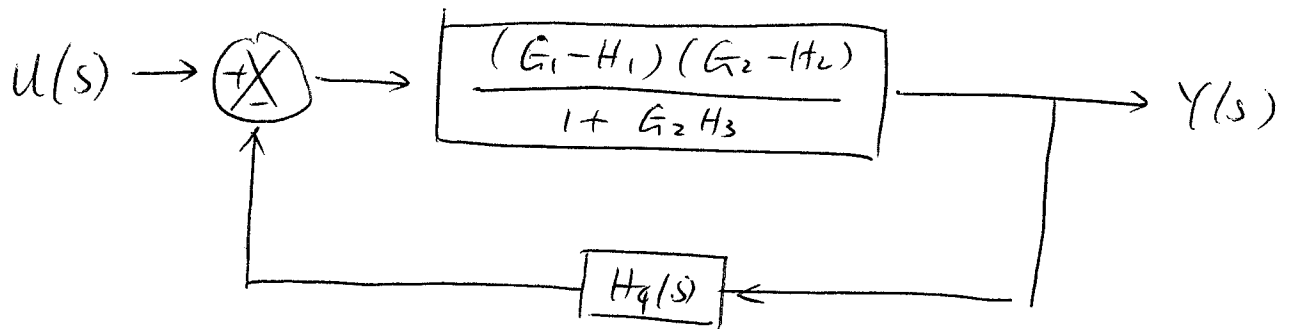
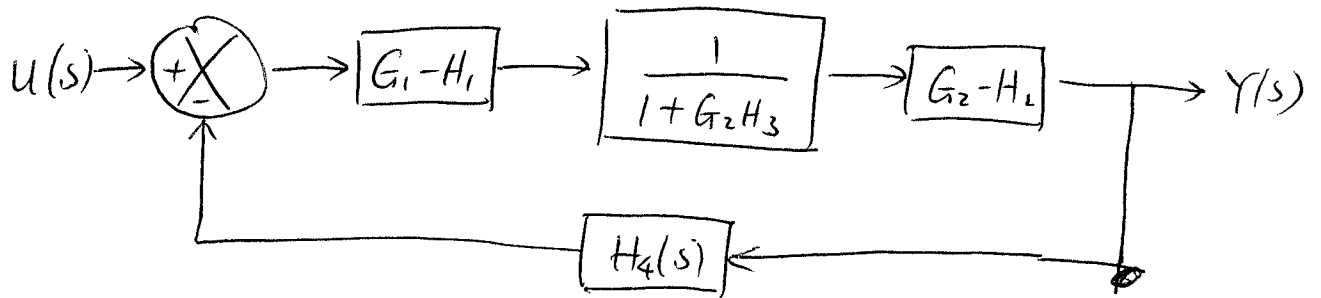
Problem 2 (40 points)

Using block-diagram reduction to simplify the following block diagram and derive the transfer function $\frac{Y(s)}{U(s)}$ from the input $U(s)$ to the output $Y(s)$. Clearly show all intermediate steps. Simplify your final transfer function so that it does NOT contain fraction of fractions. (Note: you must use the block-diagram reduction method. Using other methods will result into zero credits.)



Solution: Move the branch point A ahead of $G_2(s)$.

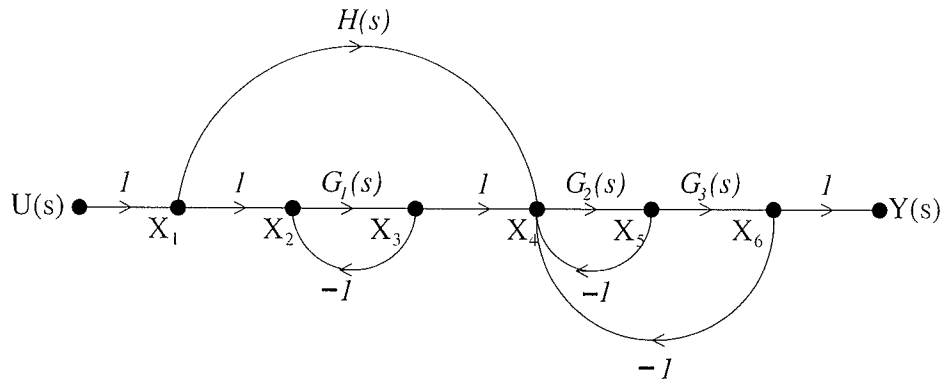




$$\begin{aligned}
 \therefore \frac{Y(s)}{U(s)} &= \frac{\frac{(G_1-H_1)(G_2-H_2)}{1+G_2H_3}}{1 + \frac{(G_1-H_1)(G_2-H_2)}{1+G_2H_3} \cdot H_4} \\
 &= \frac{(G_1-H_1)(G_2-H_2)}{1+G_2H_3 + \cancel{G_2H_3} H_4 (G_1-H_1)(G_2-H_2)}
 \end{aligned}$$

Problem 3 (50 points)

Follow the steps below and use Mason's Rule to derive the transfer function $\frac{Y(s)}{U(s)}$ for the following signal-flow-graph.



- (10 points) Identify all the loops and write down their individual loop gains.
- (10 points) Find the determinant of the graph.
- (10 points) Identify all the forward paths and write down their path gains.
- (10 points) Find the co-factors for each forward path.
- (10 points) Find the transfer function $\frac{Y(s)}{U(s)}$ using Mason's Rule.

Solution:

(a) The loops are

$$L1: X_2 X_3 X_2$$

$$L2: X_4 X_5 X_4$$

$$L3: X_4 X_5 X_6 X_4$$

Gain

$$-G_1$$

$$-G_2$$

$$-G_2 G_3$$

(b) Two non-touching Loops

$$L1 \& L2$$

$$L1 \& L3$$

Gains

$$G_1 G_2$$

$$G_1 G_2 G_3$$

$$\therefore \Delta = 1 + G_1 + G_2 + G_2 G_3 + G_1 G_2 + G_1 G_2 G_3$$

(c)	<u>Forward Paths</u>	<u>Gains</u>
	$P_1: X_1 X_2 X_3 X_4 X_5 X_6$	$G_1 G_2 G_3$
	$P_2: X_1 X_4 X_5 X_6$	$G_2 G_3 H$

(d) Cofactor for $P_1: 1$
 Cofactor for $P_2: 1 + G_1$

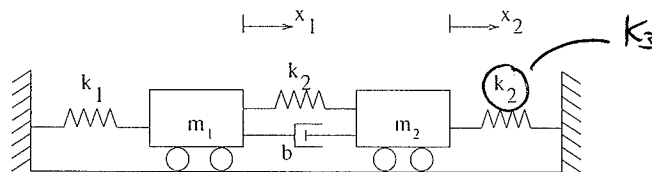
$$(e) \frac{Y(s)}{U(s)} = \frac{G_1 G_2 G_3 + G_2 G_3 H (1 + G_1)}{1 + G_1 + G_2 + G_2 G_3 + G_1 G_2 + G_1 G_2 G_3}$$

Questions for Part II

Do not justify your answer
Partial credit is NOT available

Problem 4 (30 points)

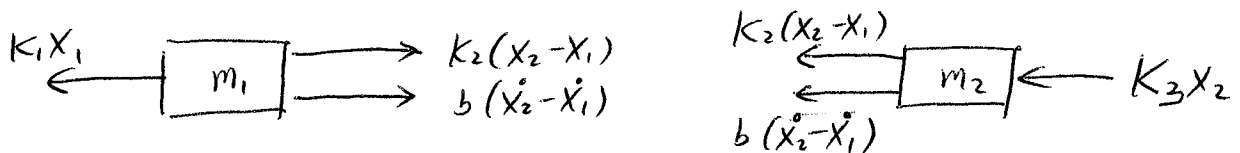
- (a) (10 points) Consider the following mechanical system. Assume that each spring (with constants k_1 , k_2 and k_3 , respectively) produces a force that is proportional to the amount of extension/compression. Assume that the damper (with constant b) produces a force that is proportional to the velocity that it is extended/compressed. Assume that the springs are relaxed when $x_1 = x_2 = 0$. Use Newton's Second Law to write down the differential equations that characterize the dynamics of the objects m_1 and m_2 . (Note: You do NOT need to solve these equations!)



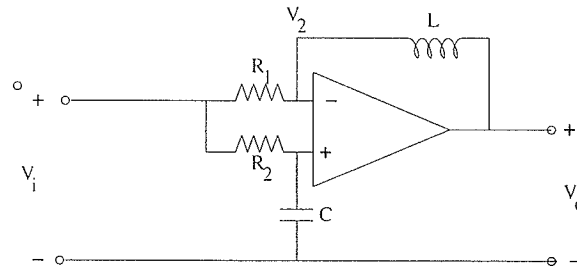
Your answer:

$$\begin{aligned}
 k_2(x_2 - x_1) + b(\dot{x}_2 - \dot{x}_1) - k_1 x_1 &= m_1 \ddot{x}_1 \\
 -k_3 x_2 - k_2(x_2 - x_1) - b(\dot{x}_2 - \dot{x}_1) &= m_2 \ddot{x}_2
 \end{aligned}$$

The free-body diagrams:



- (b) (10 points) Consider the following electrical system. Assume zero initial conditions for the inductor L and the capacitor C . Write down the two equations that govern the relationship between $V_i(s)$, $V_o(s)$ and $V_2(s)$. The unknowns in your equations should only contain $V_i(s)$, $V_o(s)$ and $V_2(s)$. (Note: You do NOT need to solve these equations!)



Your answer:

$$\frac{V_2(s) - V_i(s)}{R_2} + sC \cdot V_2(s) = 0$$

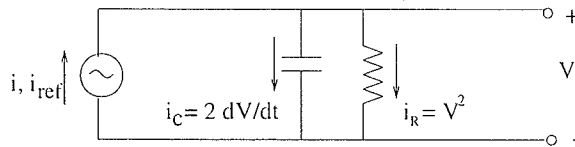
$$\frac{V_2(s) - V_i(s)}{R_1} + \frac{V_2(s) - V_o(s)}{sL} = 0$$

Note that the voltage at both inputs of the ideal op-amp is $V_2(s)$.

- (c) (10 points) Consider the following non-linear circuit. Assume that the reference input from the current source i is $i_{\text{ref}} = 1$. The linear approximation of the system is of the following form:

$$(i - 1) = a(V - 1) + b \frac{d}{dt}(V - 1),$$

where a and b are two constants. Find the values of a and b .



Your answer:

$$a = \underline{\quad 2 \quad}, \quad b = \underline{\quad 2 \quad}$$

First write down the equation that govern the dynamics of the circuit:

$$i = 2 \frac{dv}{dt} + V^2$$

When the reference input is $i_{\text{ref}} = 1$, set $\frac{dv}{dt} = 0$

$$1 = V_{\text{eq}}^2 \Rightarrow V_{\text{eq}} = 1$$

Note that the only non-linear term is $f(V) = V^2$, then

$$f(V) \approx V_{\text{eq}}^2 + \left. \frac{\partial f}{\partial V} \right|_{V=V_{\text{eq}}} (V - V_{\text{eq}})$$

$$\left. \frac{\partial f}{\partial V} \right|_{V=V_{\text{eq}}} = 2V \big|_{V=1} = 2$$

\therefore The linear approximation is

$$i = 2 \frac{dv}{dt} + 1 + 2(V - 1)$$

$$\Rightarrow (i - 1) = 2 \frac{d}{dt}(V - 1) + 2(V - 1)$$