

ECE 5984: Power Distribution System Analysis

Lecture 8: Voltage Regulation

Reference: Textbook, Chapter 7

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Outline

- Need for voltage regulation in distribution systems
- Voltage control mechanisms and standards
- Single-phase autotransformers and ABCD models
- Controlling single-phase voltage regulators
- Three-phase voltage regulators and models

Voltage regulation

- In distribution systems, voltage control is more challenging

$$\begin{aligned}\Delta V &\simeq \operatorname{Re} \{zI\} = \operatorname{Re} \left\{ z \left(\frac{s}{V} \right)^* \right\} \\ &\simeq \operatorname{Re} \{zs^*\} = rp + xq\end{aligned}$$

- Voltage drops due to both active and reactive loads (significant line resistances)

Effects of over-voltage

reduced light bulb life and electronic devices

Effects of under-voltage

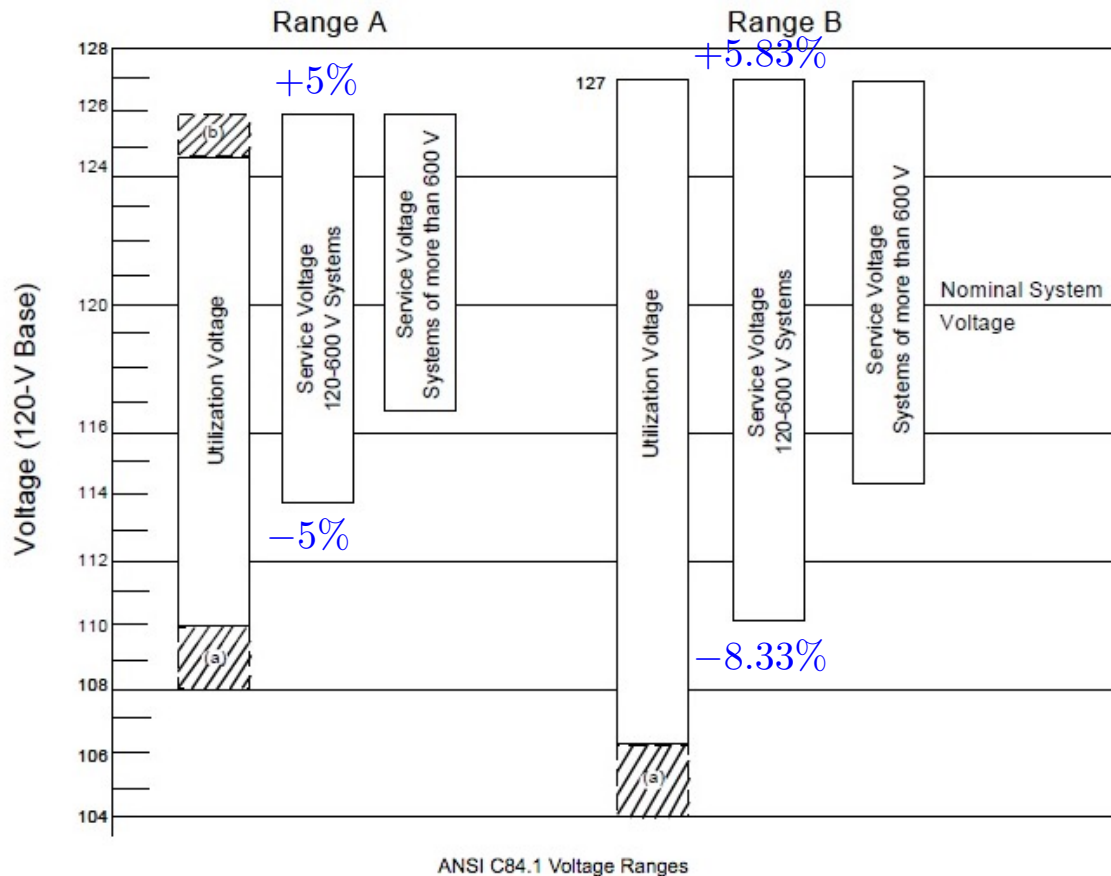
lower illumination

constant-z heating devices (e.g., water heaters) operate slower

higher starting currents on motors and overheating

Standard voltage ratings

- American National Standards Institute (ANSI) C84.1-1995 on *5-min RMS voltages*



Range A must be the basis for equipment's design and rating for satisfactory performance. Range B necessarily results from the practical operating conditions on supply and/or user systems. Such conditions should be limited in extent, duration and frequency. Corrective measures shall be undertaken within a reasonable time to bring back voltages within Range A limits.

- Service/delivery vs. utilization voltage (~4V drop within customer's wiring)*
- Imbalance less than 3% at the utility meter

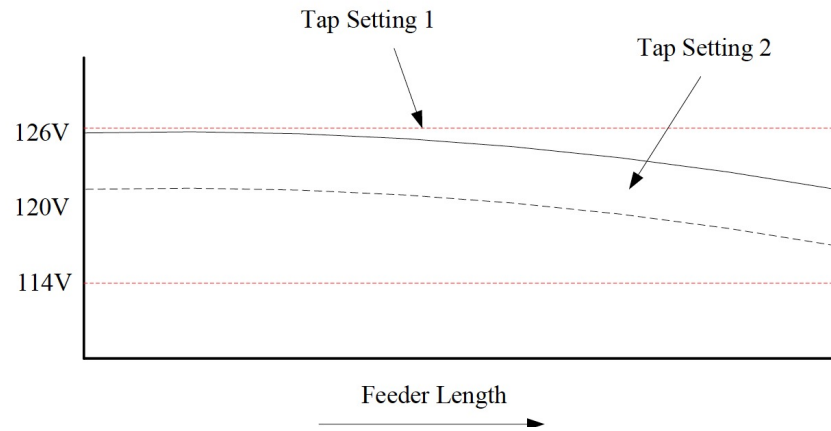
Means to improve voltage profile

- Increase feeder conductor size (reconducting)
- Increase primary voltage level
- Converting feeder sections from single-phase to multiphase
- Load balancing across feeders
- Build new substations and feeders
- Install regulators at substation and primary feeders
- Install capacitors at substation and primary feeders

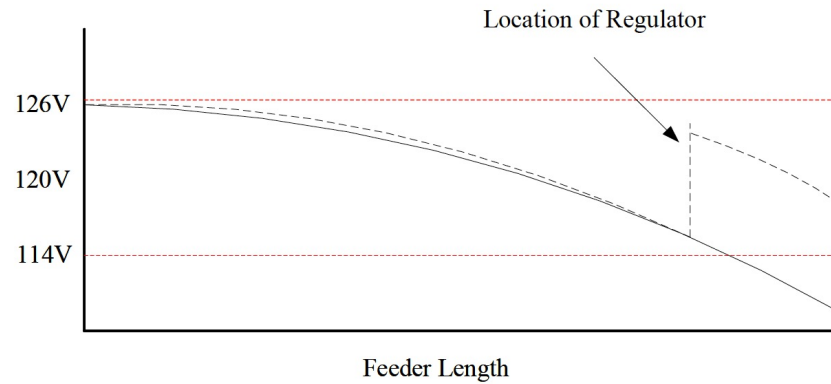
Voltage regulation mechanisms



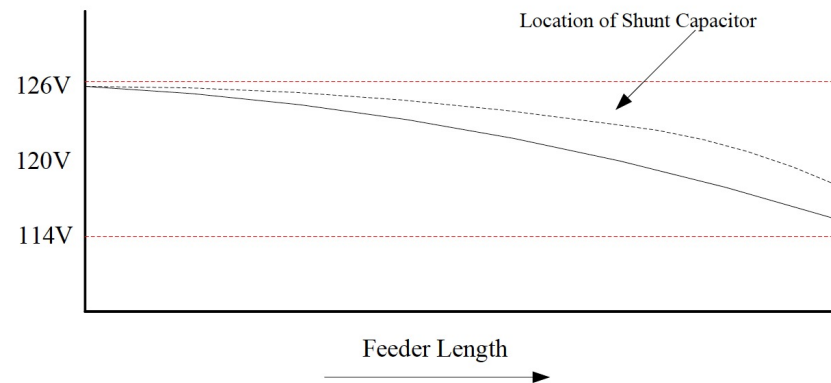
*LTC transformer
(substation
regulator)*



*step-voltage
regulator*



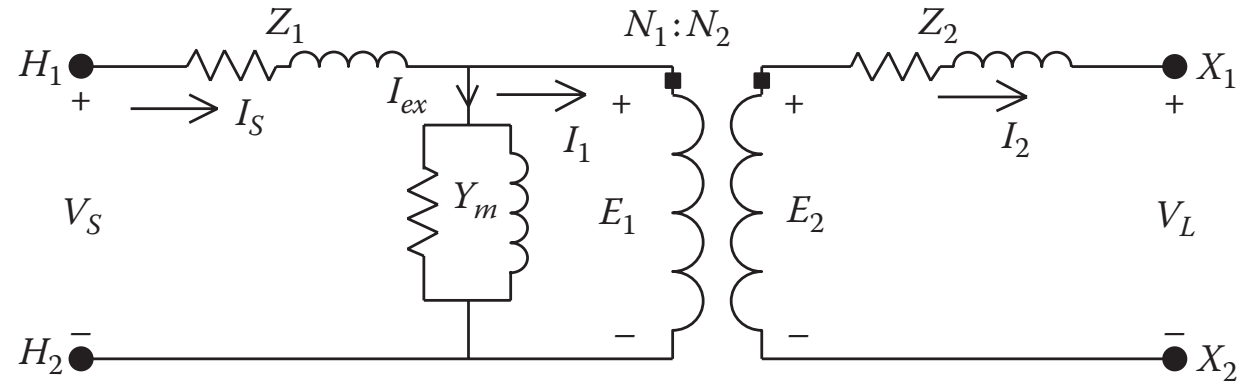
*shunt
capacitors*



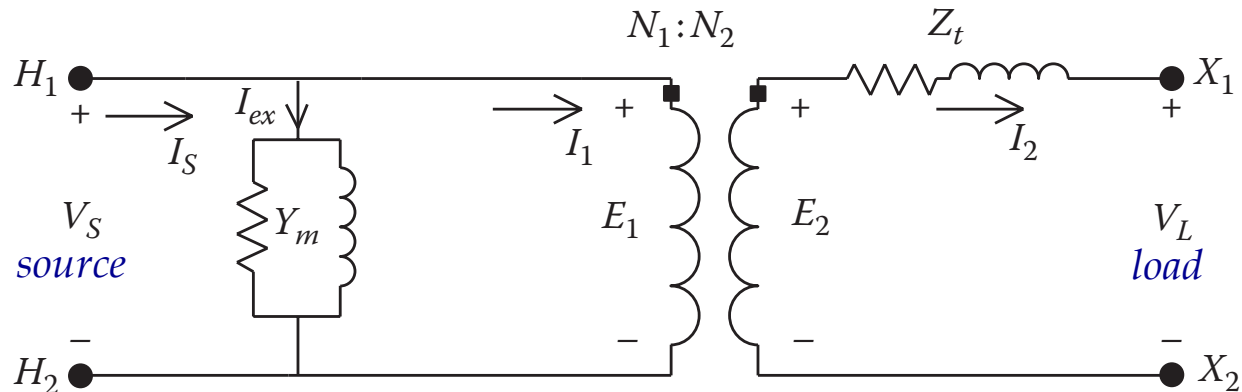
Single-phase transformer

- Detailed model

turns ratio $n_t = \frac{N_2}{N_1}$



- Approximate model

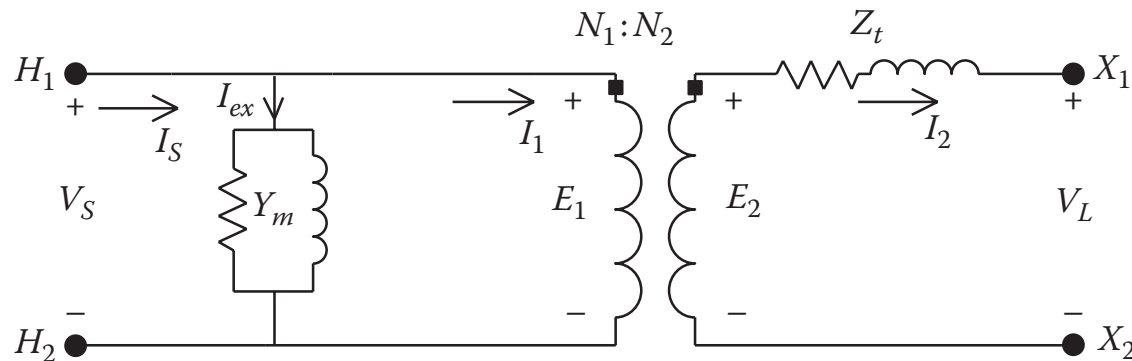


- High-voltage side impedance referred to low-voltage side

$$Z_t = n_t^2 Z_1 + Z_2$$

- The dot convention for voltages and currents

Single-phase transformer ABCD

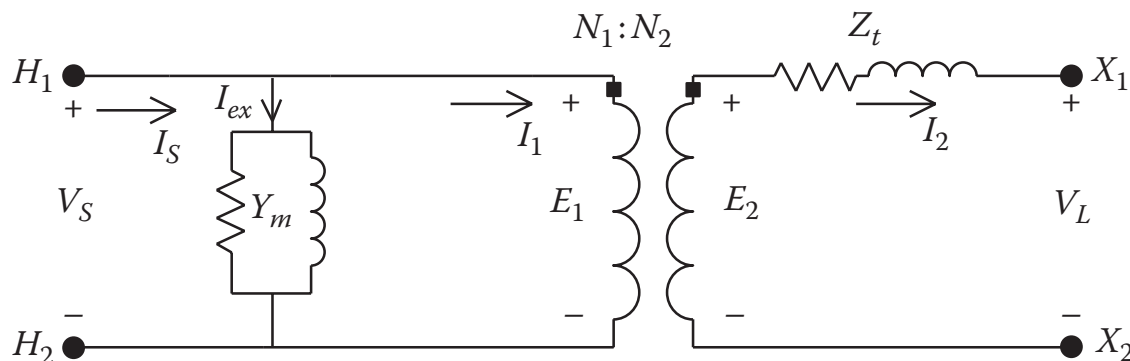


- Transformation ratios $\frac{E_1}{E_2} = \frac{N_1}{N_2} = \frac{1}{n_t}$ and $\frac{I_1}{I_2} = n_t$

- AB model $V_S = E_1 = \frac{E_2}{n_t} = \frac{1}{n_t}V_L + \frac{Z_t}{n_t}I_L$

- CD model
$$\begin{aligned} I_S &= Y_m V_S + I_1 \\ &= \frac{Y_m}{n_t} V_L + \frac{Y_m Z_t}{n_t} I_L + n_t I_L \\ &= \frac{Y_m}{n_t} V_L + \left(n_t + \frac{Y_m Z_t}{n_t} \right) I_L \end{aligned}$$

Single-phase transformer ABCD (cont'd)



$$\begin{bmatrix} V_S \\ I_S \end{bmatrix} = \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} V_L \\ I_L \end{bmatrix}$$

$$a = \frac{1}{n_t}; \quad b = \frac{Z_t}{n_t}$$

$$c = \frac{Y_m}{n_t}; \quad d = \frac{Y_m Z_t}{n_t} + n_t$$

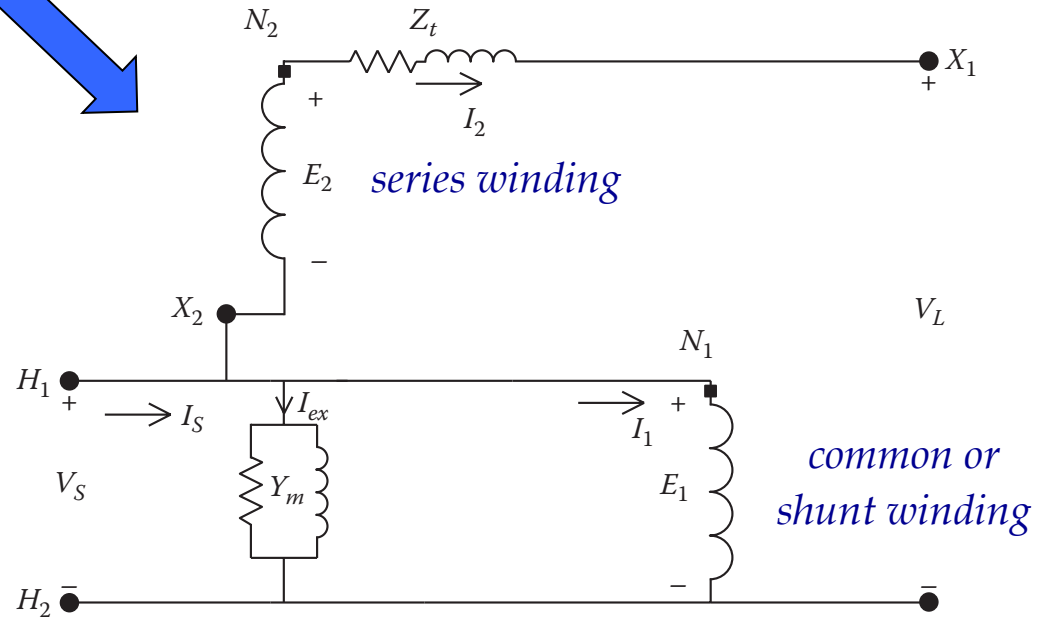
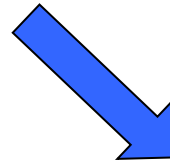
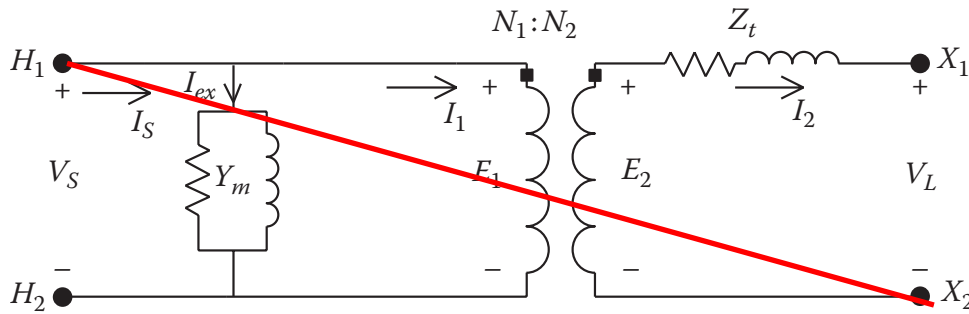
- Ideal transformer $Y_m = Z_t = 0$
- Receiving voltage in terms of its current and sending voltage (E/F model)

$$V_L = n_t V_S - Z_t I_L$$

- *Lecture 9* covers network models for all transformer connections

Autotransformer

- A single-winding tapped, or a two-winding transformer reconnected



- Windings magnetically and electrically coupled

Step-up autotransformer ABCD

- Transformation ratios

$$\frac{E_1}{E_2} = \frac{N_1}{N_2} = \frac{1}{n_t} \quad \text{and} \quad \frac{I_1}{I_2} = n_t$$

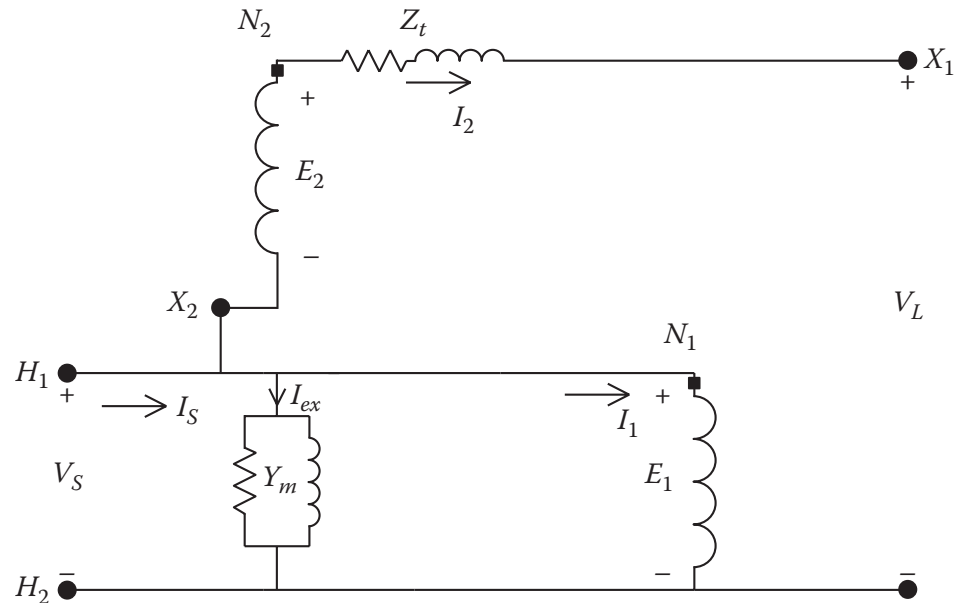
- AB model

$$\begin{aligned} V_L &= E_1 + E_2 - Z_t I_L \\ &= E_1(1 + n_t) - Z_t I_L \\ &= V_S(1 + n_t) - Z_t I_L \Rightarrow V_S = \frac{1}{1 + n_t} V_L + \frac{Z_t}{1 + n_t} I_L \end{aligned}$$

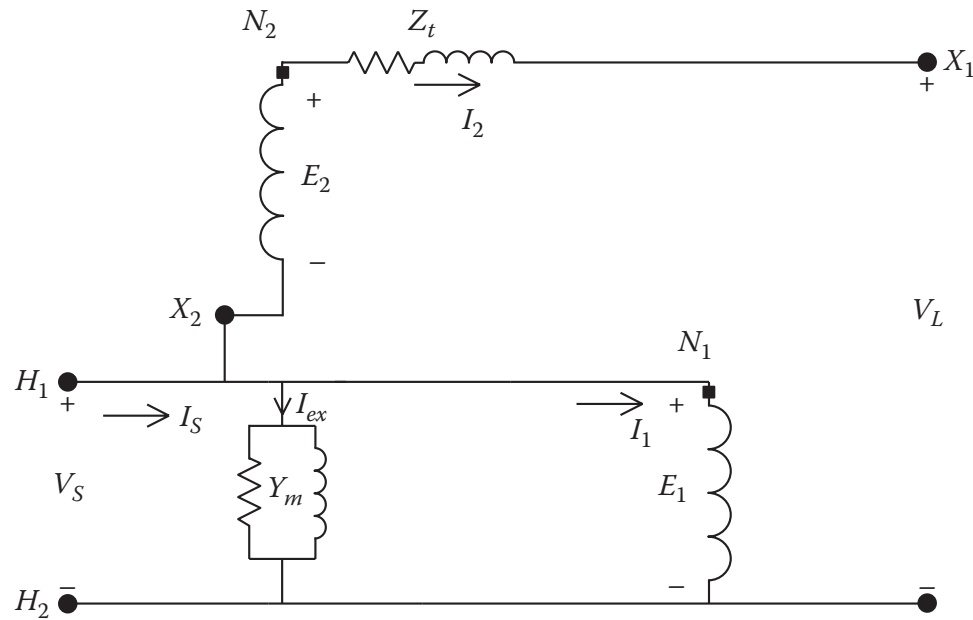
- Step-up because for small Z_t : $V_L \simeq (1 + n_t)V_S$

- CD model

$$\begin{aligned} I_S &= Y_m V_S + I_1 + I_2 \\ &= \frac{Y_m}{1 + n_t} V_L + \frac{Y_m Z_t}{1 + n_t} I_L + (1 + n_t) I_L \\ &= \frac{Y_m}{1 + n_t} V_L + \left(1 + n_t + \frac{Y_m Z_t}{1 + n_t} \right) I_L \end{aligned}$$



Step-up autotransformer ABCD (cont'd)

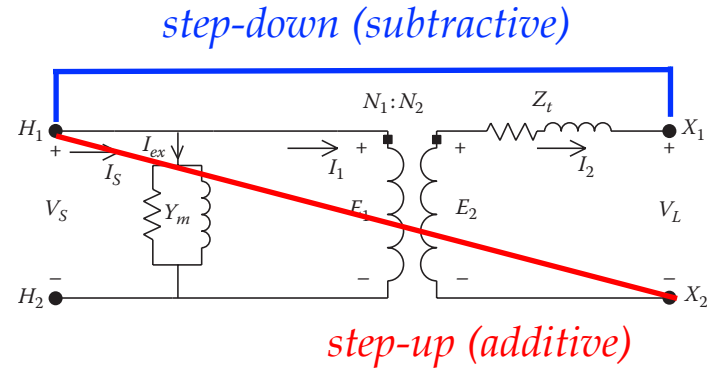


$$\begin{bmatrix} V_S \\ I_S \end{bmatrix} = \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} V_L \\ I_L \end{bmatrix} \quad \begin{aligned} a &= \frac{1}{1+n_t}; & b &= \frac{Z_t}{1+n_t} \\ c &= \frac{Y_m}{1+n_t}; & d &= 1+n_t + \frac{Y_m Z_t}{1+n_t} \end{aligned}$$

- Different from autotransformers, regulators have small ratios

Step-down autotransformer ABCD (cont'd)

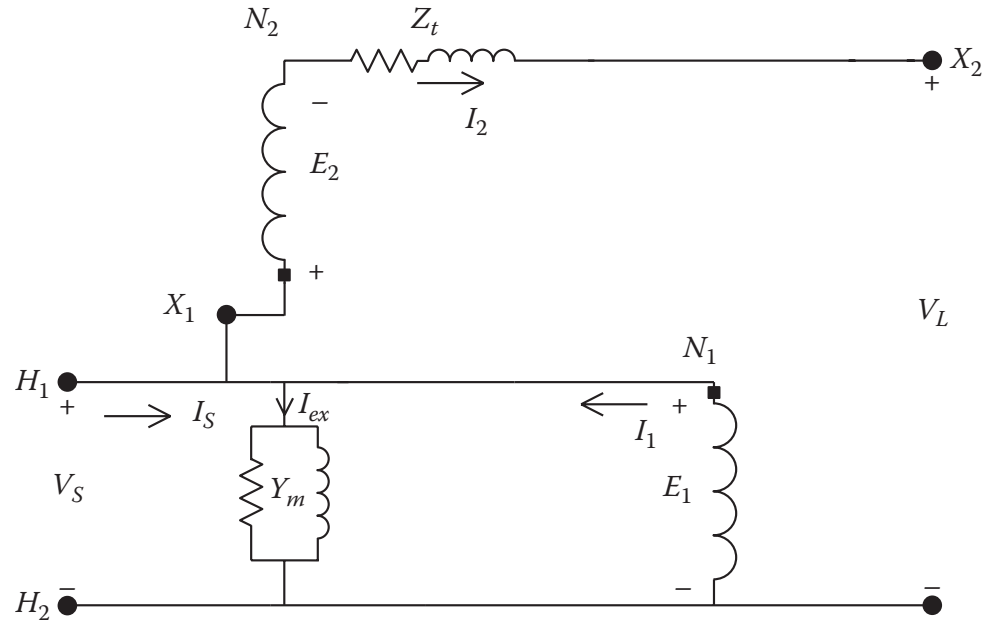
- Series winding voltage is subtracted from common winding voltage



- ABCD model can be obtained similarly using

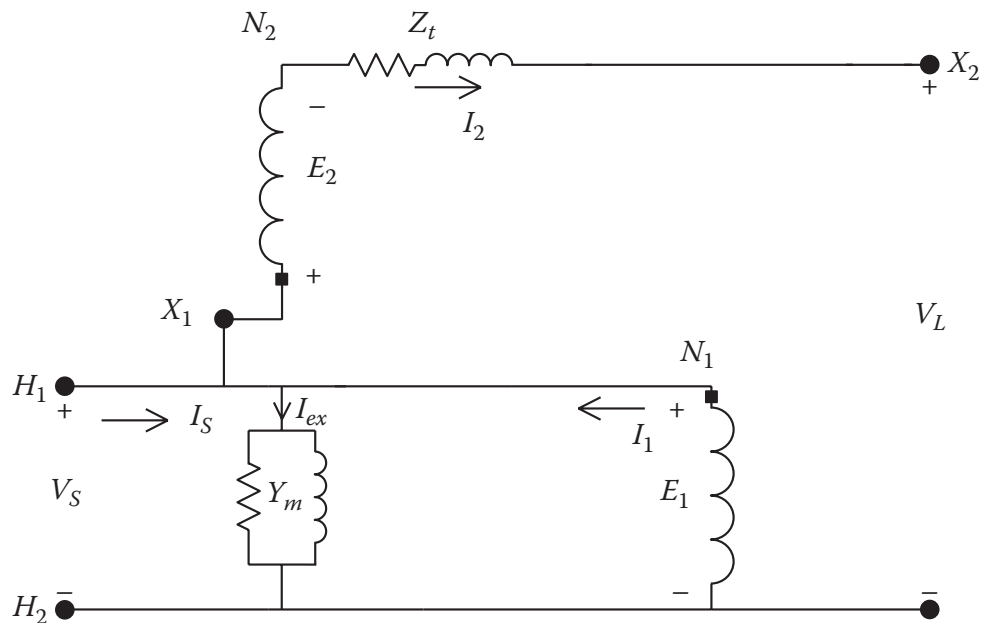
$$V_L = E_1 - E_2 - Z_t I_L$$

$$I_S = Y_m V_s - I_1 + I_2$$



- Current polarities have been reversed, but still satisfy dot convention

Step-down autotransformer ABCD (cont'd)



$$\begin{bmatrix} V_S \\ I_S \end{bmatrix} = \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} V_L \\ I_L \end{bmatrix} \quad \begin{aligned} a &= \frac{1}{1 - n_t}; & b &= \frac{Z_t}{1 - n_t} \\ c &= \frac{Y_m}{1 - n_t}; & d &= 1 - n_t + \frac{Y_m Z_t}{1 - n_t} \end{aligned}$$

- Receiving voltage in terms of its current and sending voltage

$$V_L = (1 \pm n_t)V_S - Z_t I_L$$

- Step-down model looks like step-up model with $n'_t = -n_t$

Rating advantage

- *Voltage ratings* $V_S = E_1$ and $V_L = E_1 \pm E_2 = \frac{E_2}{n_t} \pm E_2 = \frac{1 \pm n_t}{n_t} E_2$

- *Power ratings* for autotransformer and two-winding transformer

$$\begin{array}{l} S_A = V_S I_S = V_L I_L \\ S_W = E_1 I_1 = E_2 I_2 \end{array} \quad \longrightarrow \quad \frac{S_A}{S_W} = \frac{V_L I_L}{E_2 I_2} = \frac{1 \pm n_t}{n_t}$$

- Significant power rating advantage for small turns ratios

$$\text{If } V_L = (1 + n_t)V_S = 1.1 \cdot V_S \quad \Rightarrow \quad n_t = 0.1 \quad \Rightarrow \quad S_A = 11 \cdot S_W$$

$$\text{If } V_L = (1 + n_t)V_S = 10 \cdot V_S \quad \Rightarrow \quad n_t = 9 \quad \Rightarrow \quad S_A = 1.11 \cdot S_W$$

Per-unit impedances

- *Base impedances* significantly larger than those of two-winding transformers

load side

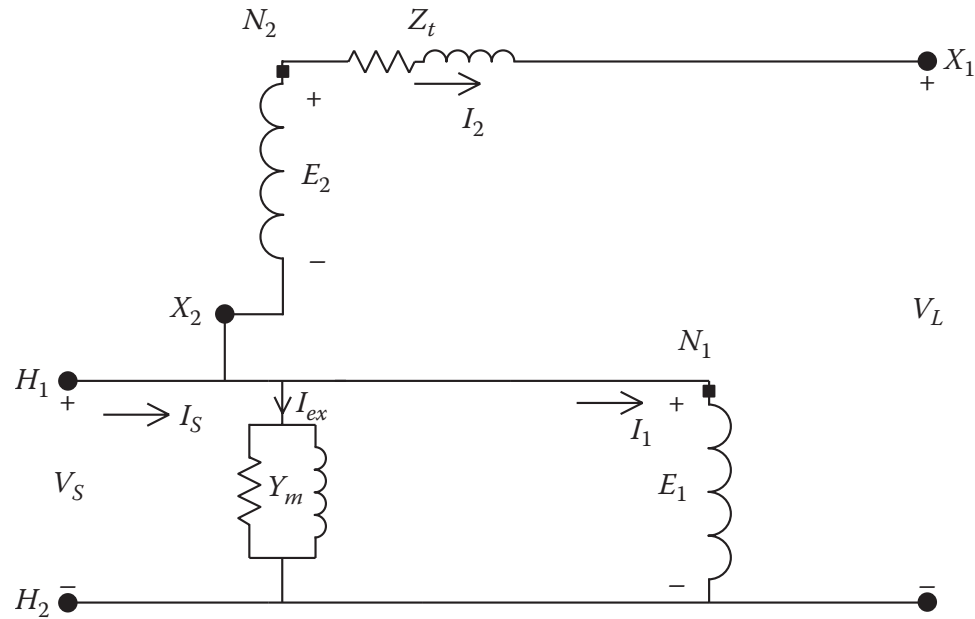
$$Z_{A,\text{base}} = \frac{V_L^2}{S_A} = \frac{(E_1 \pm E_2)^2}{S_A} = \left(\frac{1 \pm n_t}{n_t} \right)^2 \frac{E_2^2}{S_A} \quad \longrightarrow \quad \frac{Z_{A,\text{base}}}{Z_{W,\text{base}}} = \left(\frac{1 \pm n_t}{n_t} \right)^2 \frac{S_W}{S_A} = \frac{1 \pm n_t}{n_t}$$
$$Z_{W,\text{base}} = \frac{E_2^2}{S_W}$$

source side

$$Y_{A,\text{base}} = \frac{S_A}{V_S^2} = \left(\frac{1 \pm n_t}{n_t} \right)^2 \frac{S_A}{E_1^2} \quad \longrightarrow \quad \frac{Y_{A,\text{base}}}{Y_{W,\text{base}}} = \frac{1 \pm n_t}{n_t}$$
$$Y_{W,\text{base}} = \frac{S_W}{E_1^2}$$

- *Per-unit impedances* are significantly smaller than those of two-winding transformer
- Hence, autotransformers are oftentimes modeled as ideal (impedances ignored)

Approximate model of ideal transformer



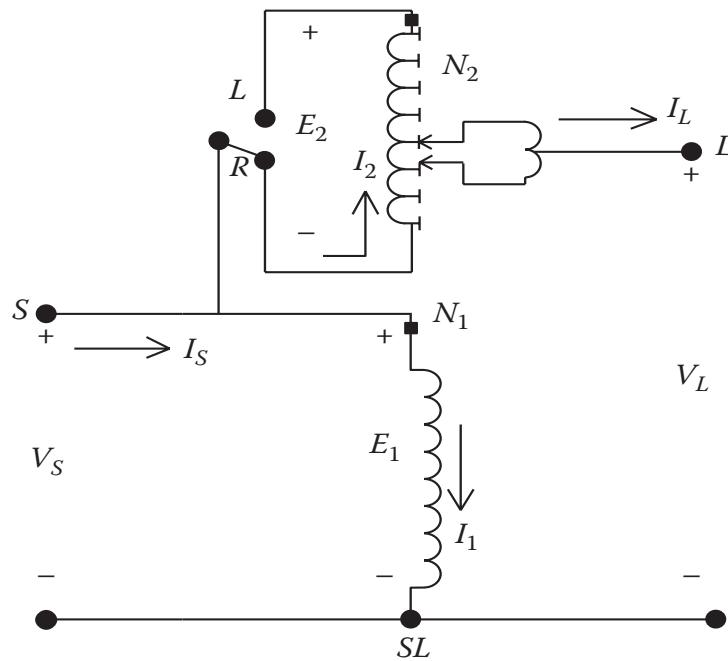
- Distribution auto-transformers can be approximated as *ideal ones* with little error

$$\begin{bmatrix} V_S \\ I_S \end{bmatrix} = \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} V_L \\ I_L \end{bmatrix} \quad \begin{array}{l} a = \frac{1}{1 \pm n_t}; \quad b \simeq 0 \\ c \simeq 0; \quad d \simeq 1 \pm n_t \end{array} \quad \begin{array}{l} + \text{ for raise} \\ - \text{ for lower} \end{array}$$

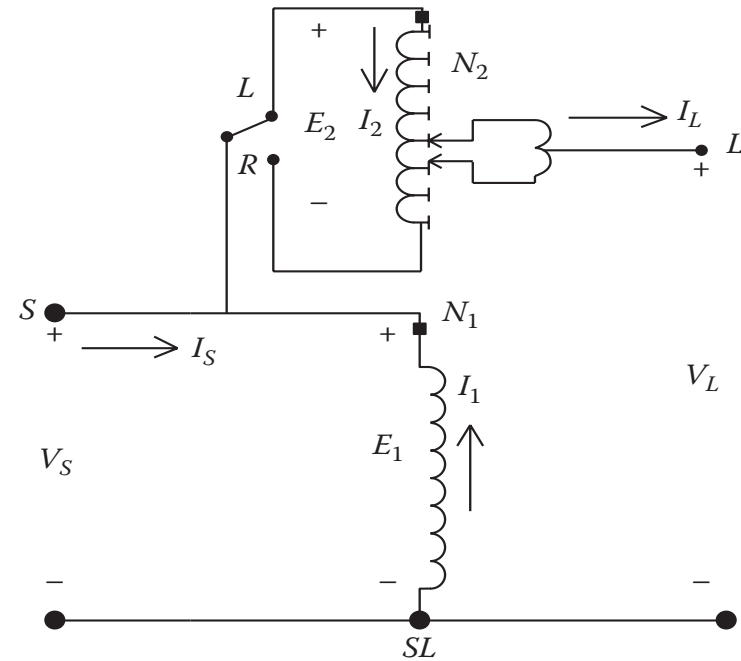
SVR of Type A

- Step-voltage regulator (SVR) = autotransformer + load tap changer (LTC) mechanism

Type A: load connected to series winding



raise position

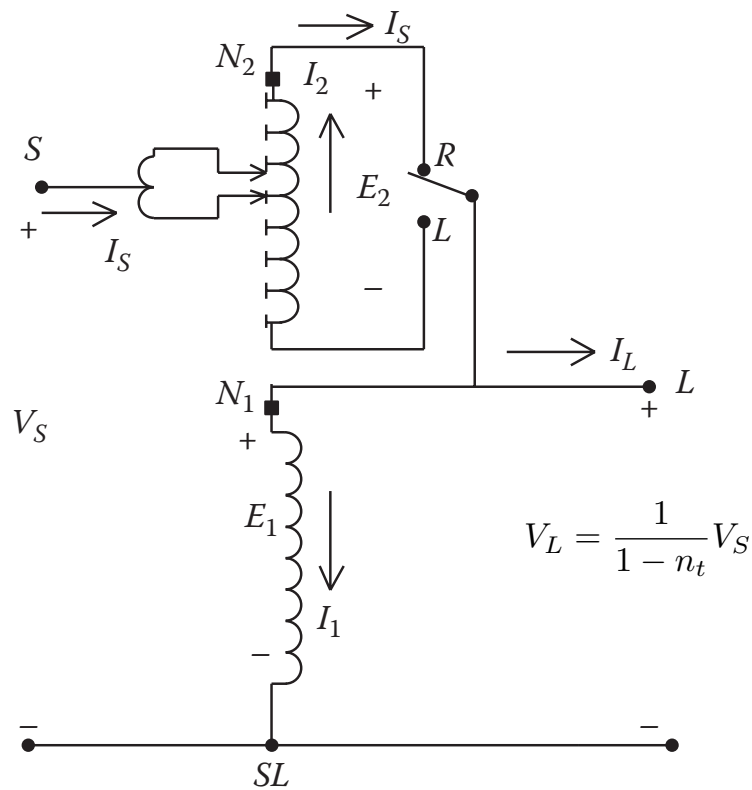


lower position

- Reversing switch changes directions of currents
- $\pm 10\%$ range in 32 steps of $20/32=0.625\%$ (approx. $0.75V$ for nominal $V_S=120V$)

SVR of Type B

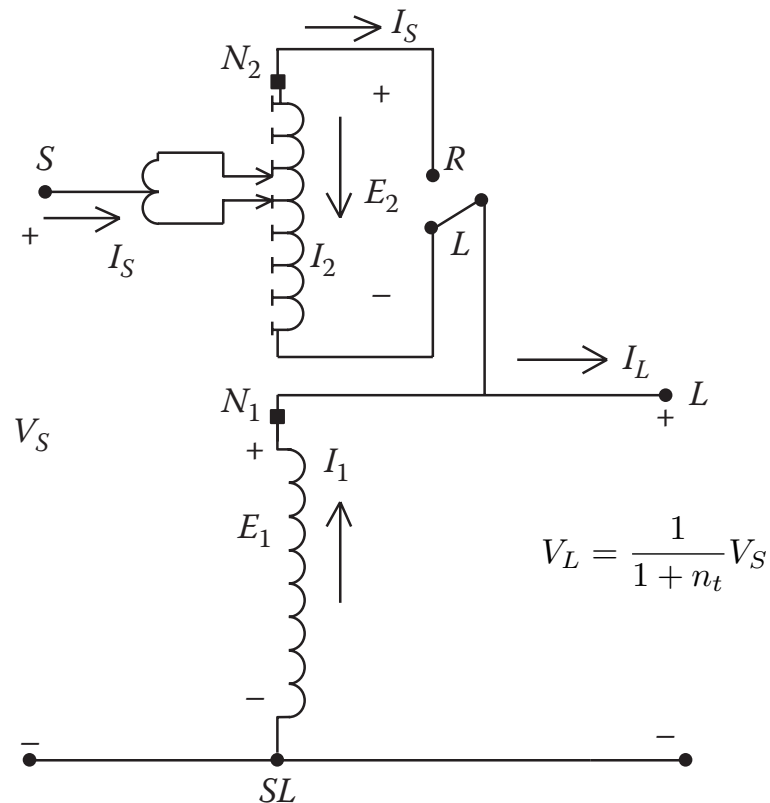
Type B: source connected to series winding (more popular)



raise position

$$V_S = E_1 - E_2 = (1 - n_t)E_1 = (1 - n_t)V_L$$

$$I_L = I_2 - I_1 = (1 - n_t)I_S$$



lower position

$$V_S = E_1 + E_2 = (1 + n_t)E_1 = (1 + n_t)V_L$$

$$I_L = I_2 + I_1 = (1 + n_t)I_S$$

Regulator model

- Neglect series impedances and shunt admittances

$$\begin{bmatrix} V_S \\ I_S \end{bmatrix} = \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} V_L \\ I_L \end{bmatrix}$$

- Type B (shown here)

$$a = a_R; \quad b = c = 0; \quad d = \frac{1}{a_R}; \quad a_R = 1 \mp \frac{N_2}{N_1}$$

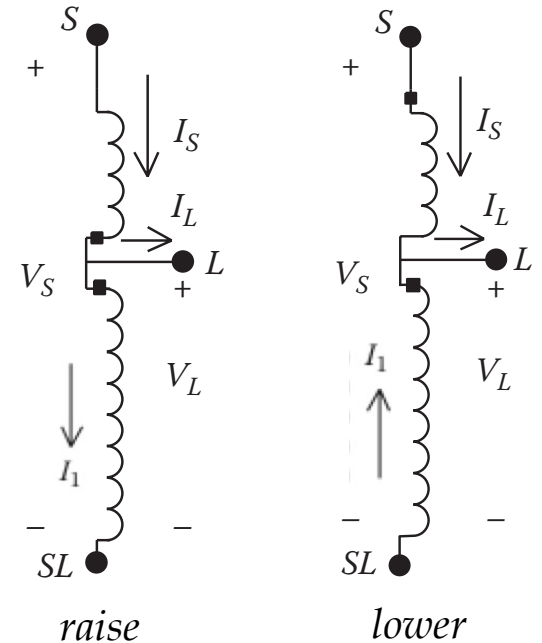
$$a_R = 1 \mp 0.00625 \cdot \text{Tap}$$

- Type A (derived earlier)

$$a = \frac{1}{a_R}; \quad b = c = 0; \quad d = a_R; \quad a_R = 1 \pm \frac{N_2}{N_1}$$

- Example of raise position

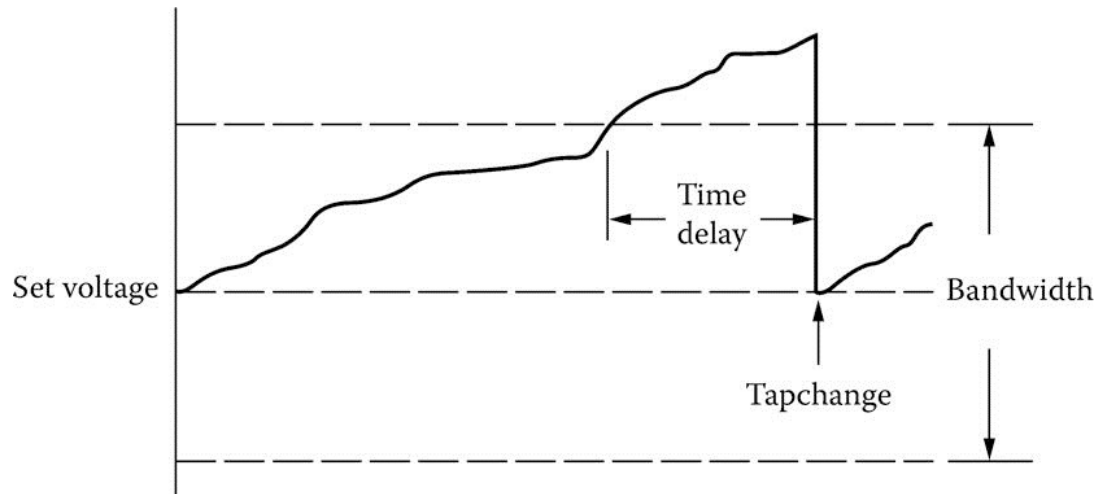
$$V_L = (1 + n_t)V_S \text{ (type A)} \quad \text{or} \quad V_L = \frac{1}{1 - n_t}V_S \text{ (type B)}$$



	Type A	Type B
Raise	+	-
Lower	-	+

Control settings

- *Set voltage* (e.g., 124 V)



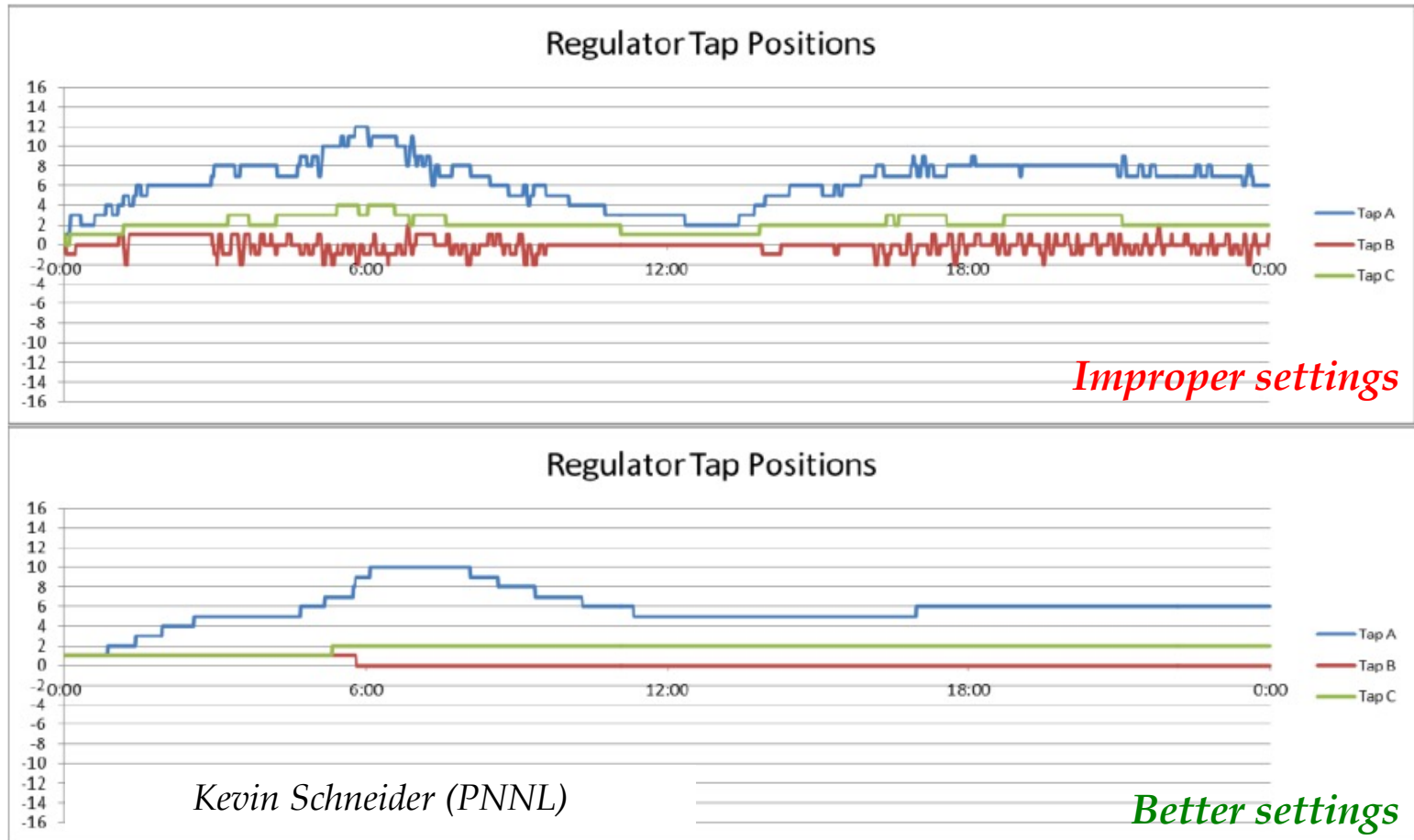
- *Bandwidth*: allowed deviation around set voltage level (e.g., $\pm 1\text{V}$); specified as full (two-sided) bandwidth

Deadband should be wider than the voltage change obtained by a single tap, otherwise tap may be flipping back-and-forth without ever falling in the deadband

- *Time delay* before taking action (to avoid fluctuations due to e.g., motor starting)

Tuning bandwidth and delay

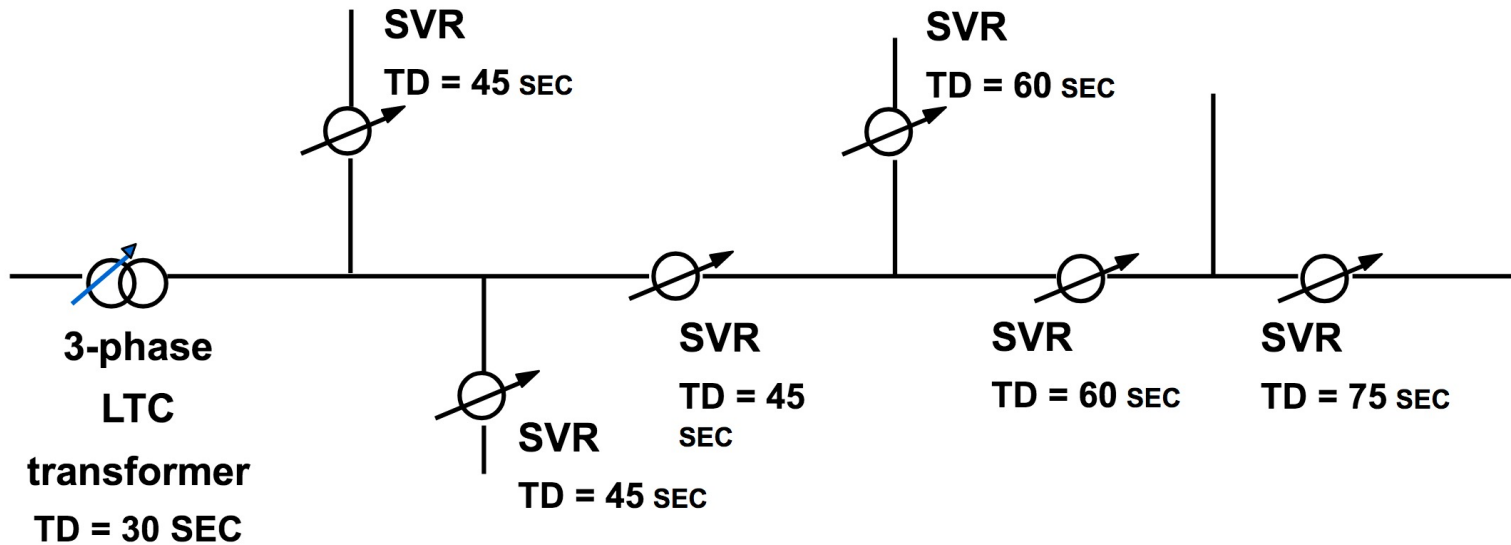
- Avoid excessive regulator operation and oscillations (reduces lifetime)



- Tuning is harder (impossible) with fluctuations in solar irradiance and PVs
- Smart inverters can handle fluctuations at faster timescale

Time delays

- Usually, in the range of 10-120 seconds



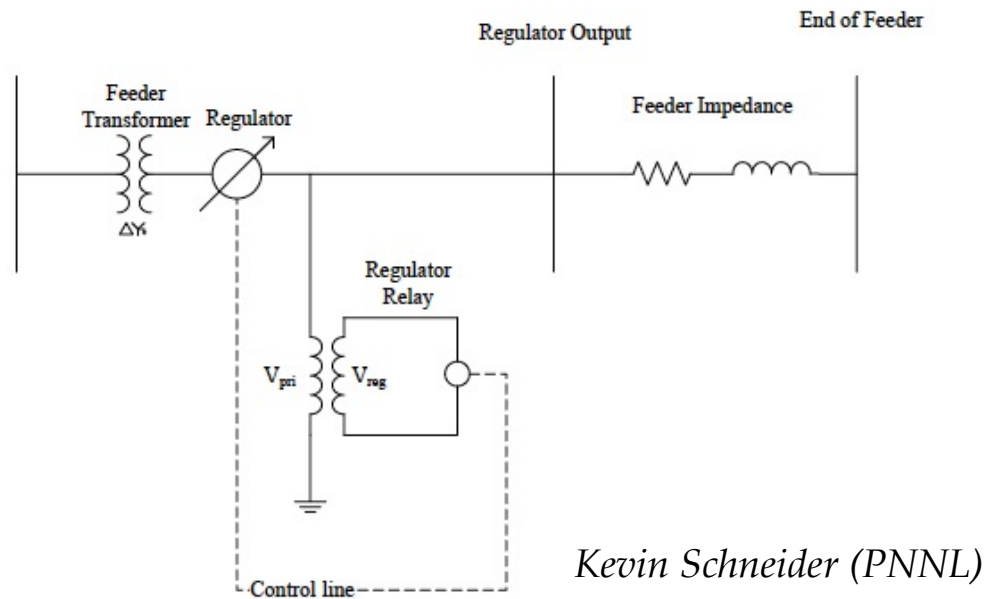
Eaton – Cooper Power Systems

- *Rule 1*: Each regulator in series downstream requires a longer time delay
- *Rule 2*: Minimum time delay from one SVR to the next in cascade is 15 seconds

Control of voltage regulators

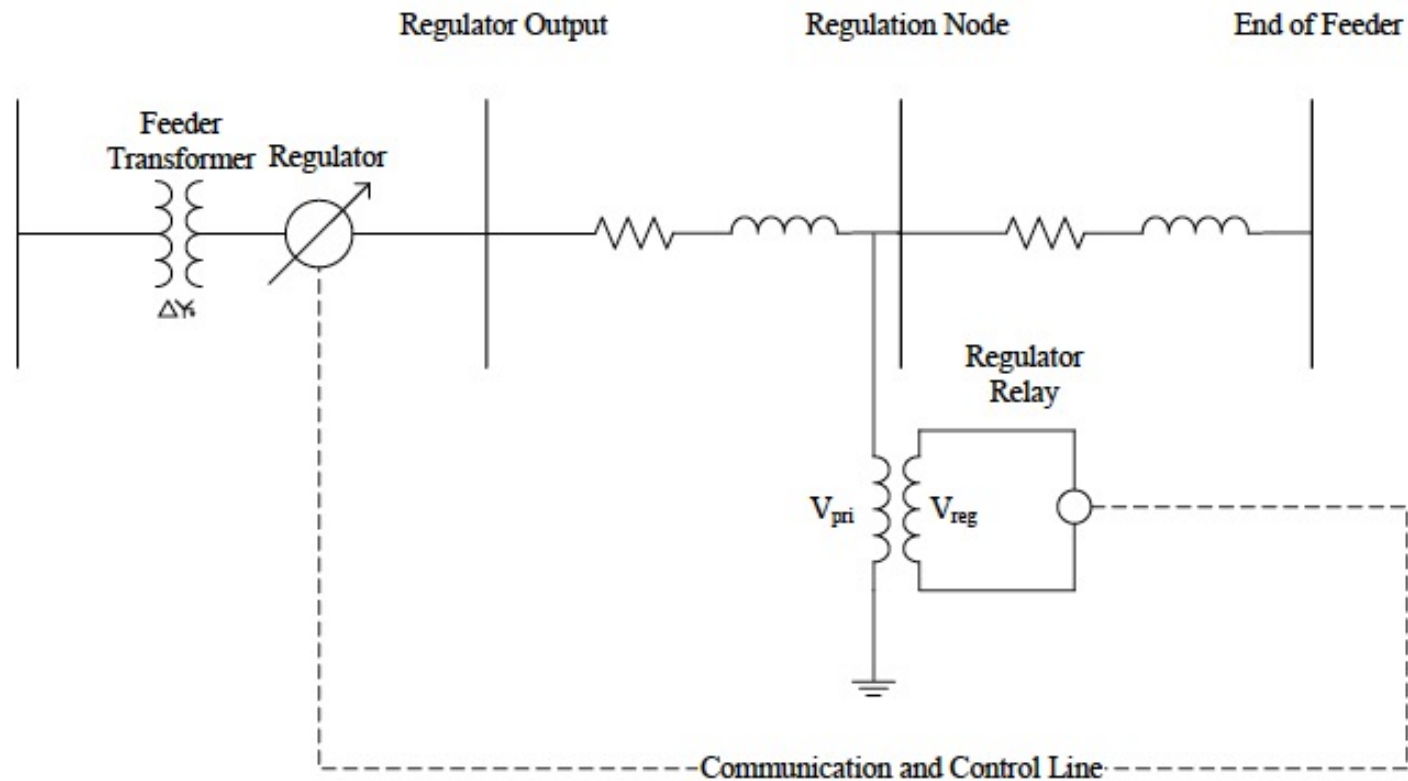
- Not all feeders have SVRs: no need for short and/or lightly-loaded feeders

1. Control based on *local voltage output*



Control of voltage regulators

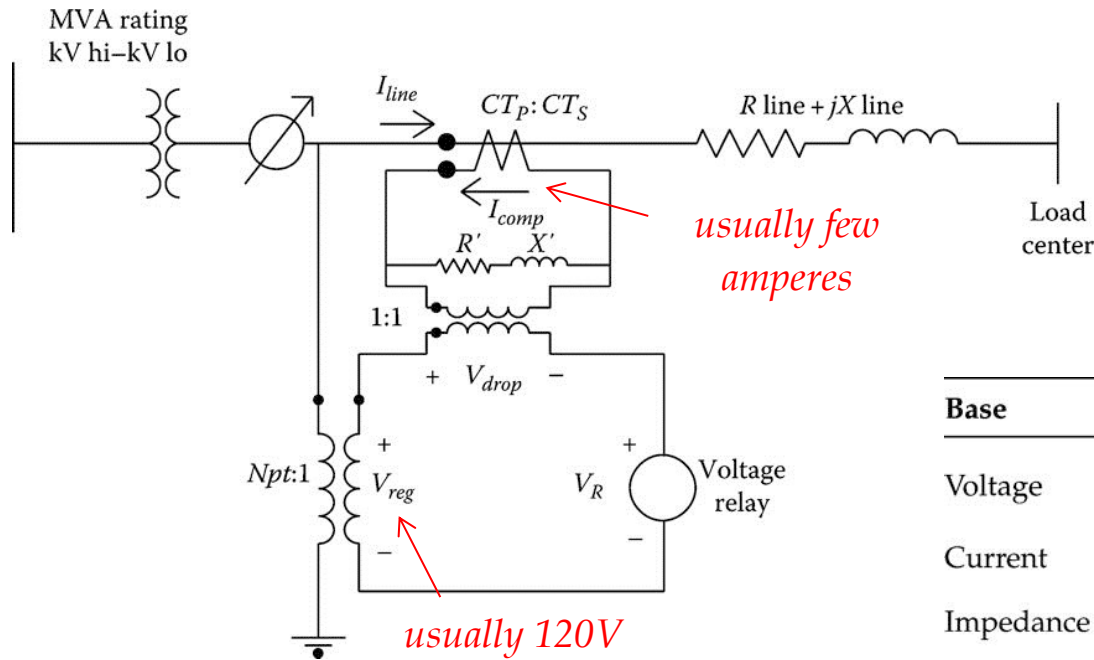
2. Control based on *remote voltage*



Kevin Schneider (PNNL)

Line drop compensator (LDC)

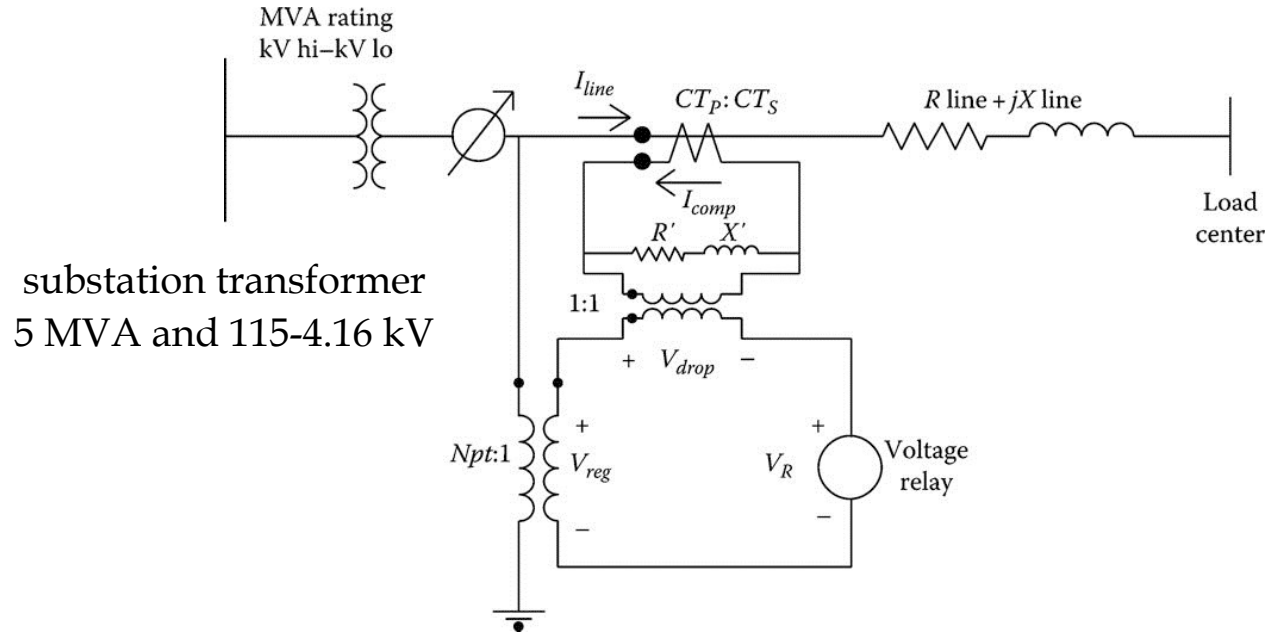
- Equivalent impedance is an analog (or digital) model of the actual line



Base	Line Circuit	Compensator Circuit
Voltage	V_{LN}	$\frac{V_{LN}}{N_{PT}}$
Current	CT_P	CT_S
Impedance	$Z_{base_{line}} = \frac{V_{LN}}{CT_P}$	$Z_{base_{comp}} = \frac{V_{LN}}{N_{PT} \cdot CT_S}$

- LDC impedance $Z_{LDC} = Z_{line} \frac{Z_{LDC,base}}{Z_{line,base}} = Z_{line} \left(\frac{V_{LDC,base}}{V_{line,base}} \right) \left(\frac{I_{line,base}}{I_{LDC,base}} \right) = Z_{line} \frac{CT_P}{N_{PT} CT_S} \Omega$
- LDC impedance oftentimes expressed in Volts $Z_{LDC,V} = Z_{LDC} CT_S = Z_{line} \frac{CT_P}{N_{PT}} V$
(voltage drop under rated current)

Examples 7.4-7.6



1. Find PT/CT ratings if the LDC ratings are 120V and 5A

- LDC potential transformer

$$V_{LL,base} = 4.16\text{kV} \Rightarrow V_{LN,base} = 2,401.8\text{V} \simeq 2.4\text{kV} \Rightarrow N_{PT} = \frac{2,400\text{V}}{120\text{V}} = 20$$

- LDC current transformer

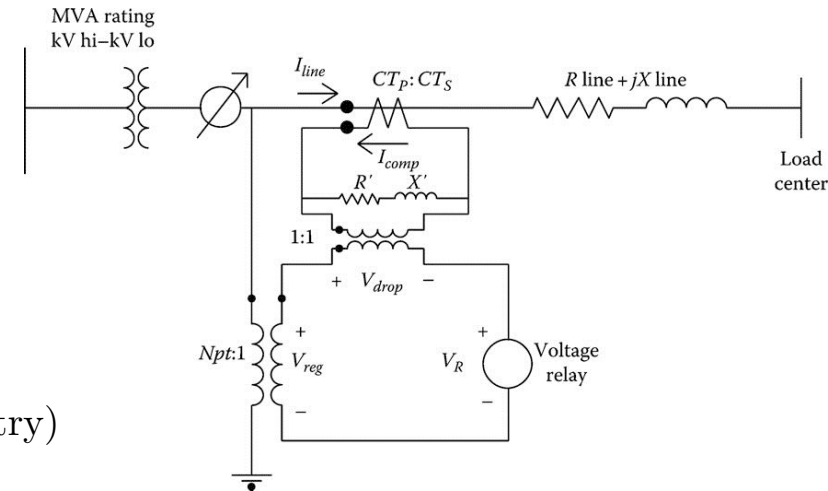
$$I_{base} = \frac{5\text{MVA}}{\sqrt{3} \cdot V_{LL,base}} = 693.9\text{A} \simeq 700\text{A} \Rightarrow CT = \frac{CT_P}{CT_S} = \frac{700\text{A}}{5\text{A}} = 140$$

Examples 7.4-7.6 (cont'd)

2. Find LDC setting in ohms and volts if the equivalent impedance between SVR and load center is $Z_{\text{line}} = 0.3 + j0.9 \Omega$

$$Z_{\text{LDC}} = Z_{\text{line}} \frac{CT_P}{N_{PT} CT_S} \Omega = 2.1 + j6.3 \Omega$$

$$Z_{\text{LDC},V} = (2.1 + j6.3) \Omega \cdot 5A = 10.5 + j31.5V \text{ (data entry)}$$

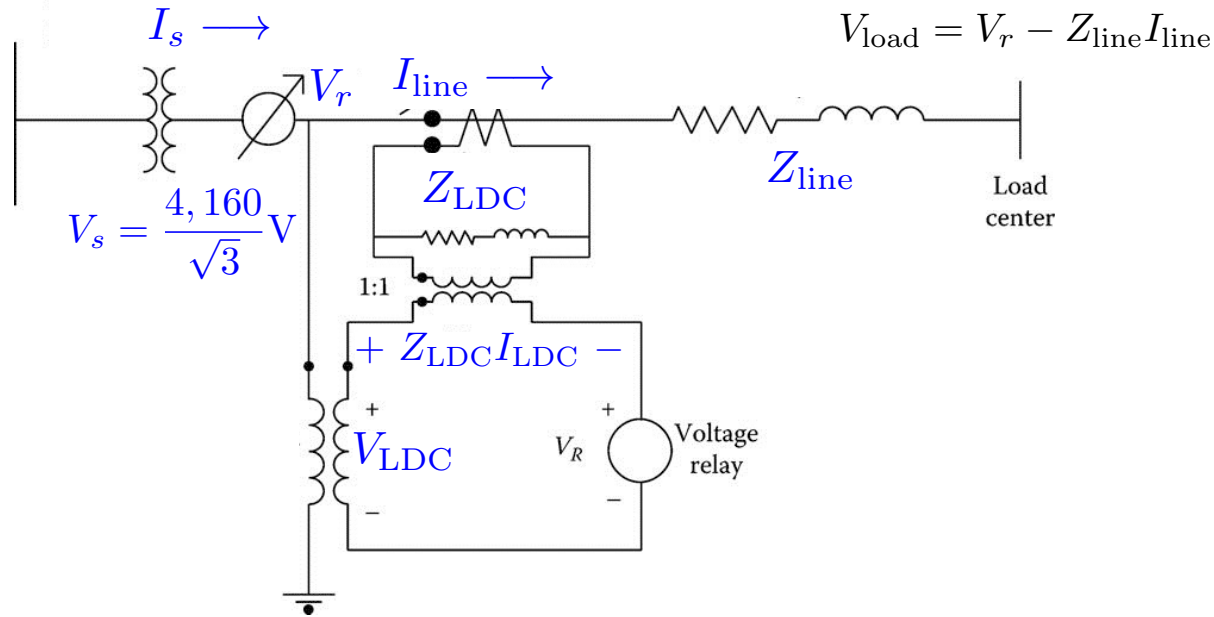


3. 2.5 MVA load is to be served at 4.16 kV with 0.9 PF lagging. Voltage level to be held within 2V around 120V. Determine tap position so relay voltage is within [119:121]V

- Line current (assuming constant-current load)

$$I_{\text{line}} = \frac{2.5\text{MVA}}{\sqrt{3} \cdot V_{\text{LL,base}}} = 346.97 \angle -25.84^\circ \Rightarrow I_{\text{LDC}} = 2.478 \angle -25.84^\circ \text{ A}$$

Examples 7.4-7.6 (cont'd)



- Relay voltage models load voltage $V_{\text{relay}} = V_{\text{LDC}} - Z_{\text{LDC}} I_{\text{LDC}} = \frac{V_r - Z_{\text{line}} I_{\text{line}}}{N_{\text{PT}}} = \frac{V_{\text{load}}}{N_{\text{PT}}}$
- Regulator output (type-B) $V_r = \frac{V_s}{a_R(\tau)}$ where $a_R(\tau) = 1 - 0.00625 \cdot \tau$
and $\tau \in \{-16, \dots, -1, 0, +1, \dots, +16\}$
- At neutral position $a_R(0) = 1$, relay voltage is $|V_{\text{relay}}| = 109.24 \text{ V}$
- Raise tap τ so that relay voltage goes above 119V

Examples 7.4-7.6 (cont'd)

- Relay voltage magnitude for different tap settings
- Working assumptions
 - constant-current load
 - VR started from neutral position
- Can you repeat the analysis for a constant-Z load?
- For specific loading, the SVR tap depends also on prior loading conditions
- ✓ e.g., if the SVR was at +16 before (due to heavy load earlier), then it will now stop at tap +14, not +13 simply because it started *'from above'*

neutral position

-16.0000	98.3929
-15.0000	99.0122
-14.0000	99.6388
-13.0000	100.2727
-12.0000	100.9139
-11.0000	101.5628
-10.0000	102.2193
-9.0000	102.8837
-8.0000	103.5560
-7.0000	104.2365
-6.0000	104.9252
-5.0000	105.6223
-4.0000	106.3279
-3.0000	107.0423
-2.0000	107.7656
-1.0000	108.4979

0	109.2394
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1.0000	109.9903
2.0000	110.7508
3.0000	111.5210
4.0000	112.3012
5.0000	113.0914
6.0000	113.8921
7.0000	114.7032
8.0000	115.5251
9.0000	116.3579
10.0000	117.2019
11.0000	118.0573
12.0000	118.9243

final position

13.0000	119.8031
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14.0000	120.6941
15.0000	121.5974
16.0000	122.5133

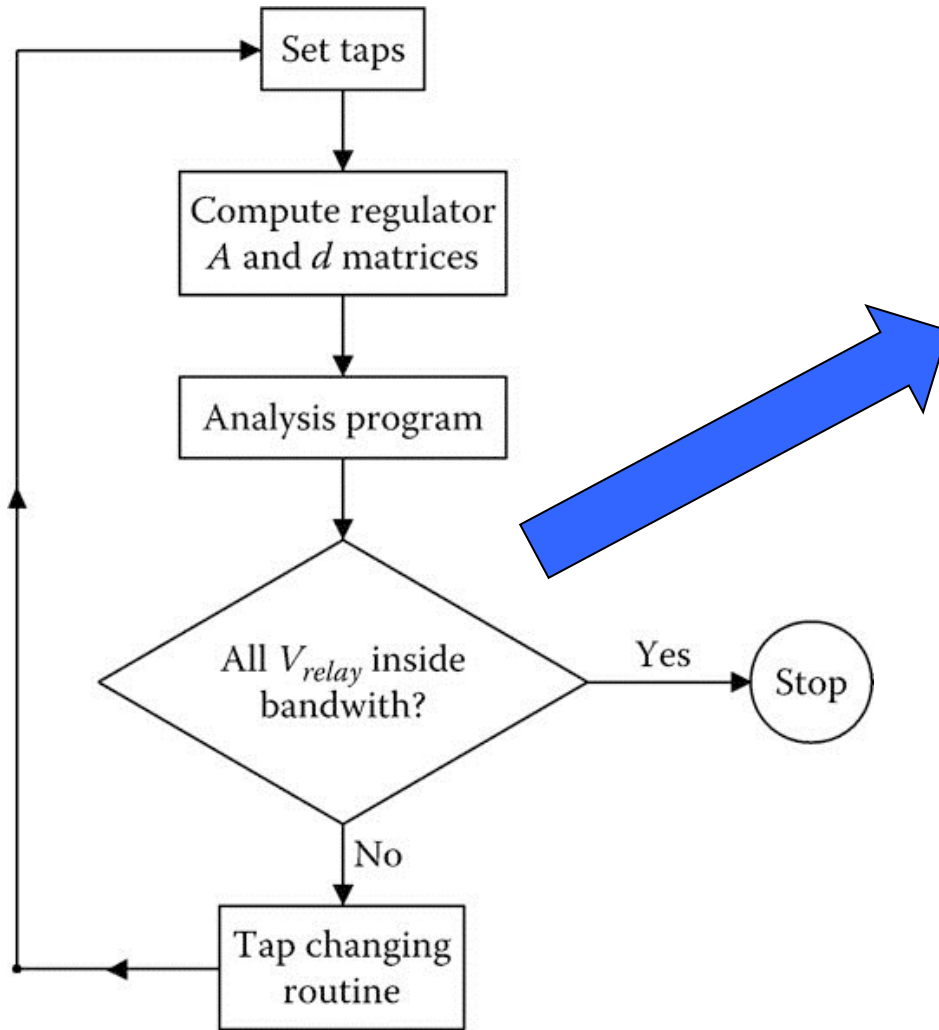
Examples 7.4-7.6 (textbook solution)

- Book's solution uses three approximations

$$\begin{aligned}
 V_{\text{relay}} &= \frac{V_s}{(1 - 0.00625 \cdot \tau) N_{\text{PT}}} - \frac{Z_{\text{line}} I_{\text{line}}}{N_{\text{PT}}} && \text{approx. 1: } \frac{1}{1 - 0.00625 \cdot \tau} \simeq 1 + 0.00625 \cdot \tau \\
 &\simeq (1 + 0.00625 \cdot \tau) \frac{V_s}{N_{\text{PT}}} - \frac{Z_{\text{line}} I_{\text{line}}}{N_{\text{PT}}} \\
 &= \left(\frac{V_s}{N_{\text{PT}}} - \frac{Z_{\text{line}} I_{\text{line}}}{N_{\text{PT}}} \right) + \left(0.00625 \frac{V_s}{N_{\text{PT}}} \right) \tau && \text{approx. 2: } 0.00625 \times (\sim 120) = 0.75 \\
 &\simeq \hat{V}_s + 0.75\tau \Rightarrow \\
 \tau &= \frac{V_{\text{relay}} - \hat{V}_s}{0.75} \simeq \frac{|V_{\text{relay}}| - |\hat{V}_s|}{0.75} = \frac{119 - 109.24}{0.75} = 13.02 && \text{approx. 3: small phase differences}
 \end{aligned}$$

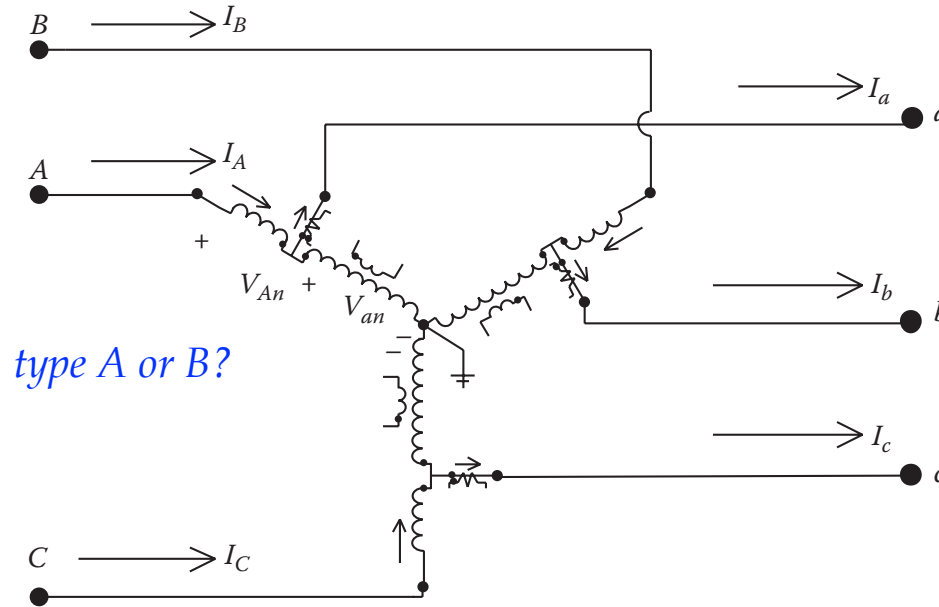
- Textbook also repeats analysis for constant power and finds final tap position to be +12

Finding taps using a PF solver



$$\begin{array}{l}
 Y := \text{for } i \in 1..3 \\
 \left| \begin{array}{l}
 V_{\text{reg}_i} \leftarrow \frac{VLN_{3i}}{N_{\text{pt}}} \\
 I_{\text{reg}_i} \leftarrow \frac{I_{\text{abc}_i}}{CT}
 \end{array} \right. \\
 V_{\text{relay}} \leftarrow V_{\text{reg}} - Z_{\text{comp}} \cdot I_{\text{reg}} \\
 \text{Tap}_1 \leftarrow \text{Tap}_1 + 1 \text{ if } |V_{\text{relay}_1}| < 120 \\
 \text{Tap}_2 \leftarrow \text{Tap}_2 + 1 \text{ if } |V_{\text{relay}_2}| < 120 \\
 \text{Tap}_3 \leftarrow \text{Tap}_3 + 1 \text{ if } |V_{\text{relay}_3}| < 120 \\
 \text{Out}_1 \leftarrow \text{Tap} \\
 \text{Out}
 \end{array}$$

Wye SVR



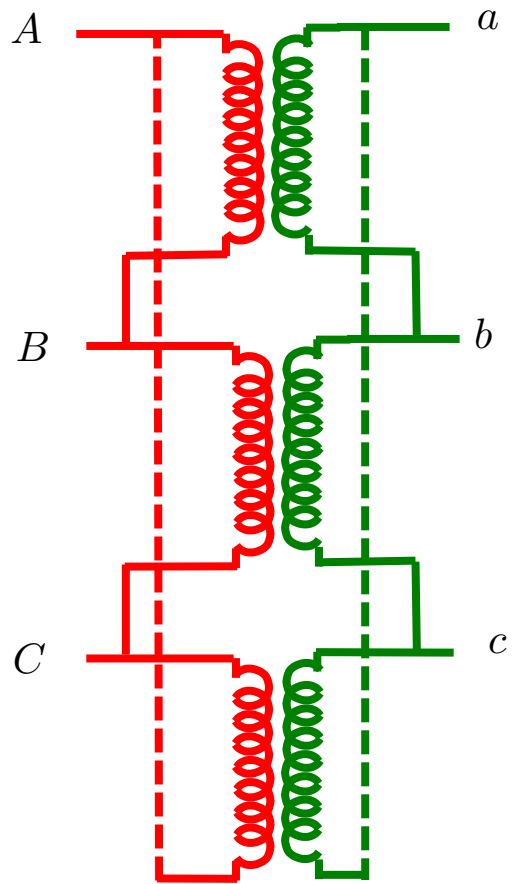
- Each regulator may have its own tap changer/LDC
- Advantage over a single three-phase SVRs that are 'gang'-operated
- Problem if LDCs share the same line impedance
- Generalized matrices

$$\mathbf{v}_\Phi = \mathbf{A}\mathbf{v}_\phi + \mathbf{B}\mathbf{i}_\ell$$

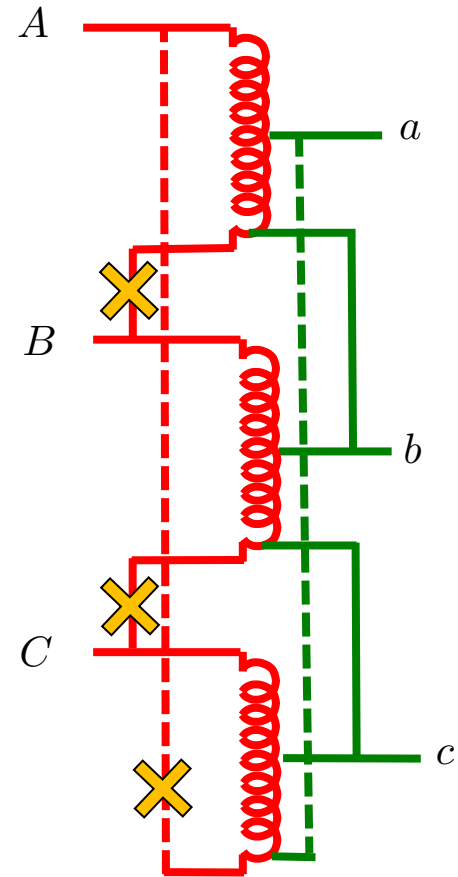
$$\mathbf{i}_L = \mathbf{C}\mathbf{v}_\phi + \mathbf{D}\mathbf{i}_\ell$$

$$\mathbf{A} = \begin{bmatrix} a_R^a & 0 & 0 \\ 0 & a_R^b & 0 \\ 0 & 0 & a_R^c \end{bmatrix}; \quad \mathbf{B} = \mathbf{C} = \mathbf{0}; \quad \mathbf{D} = \mathbf{A}^{-1}$$
- Model holds for raise/lower position; extends trivially to open Wye connection

Delta-connected regulators



3 single-phase ordinary transformers connected in delta



3 single-phase voltage regulators (type B) connected in delta

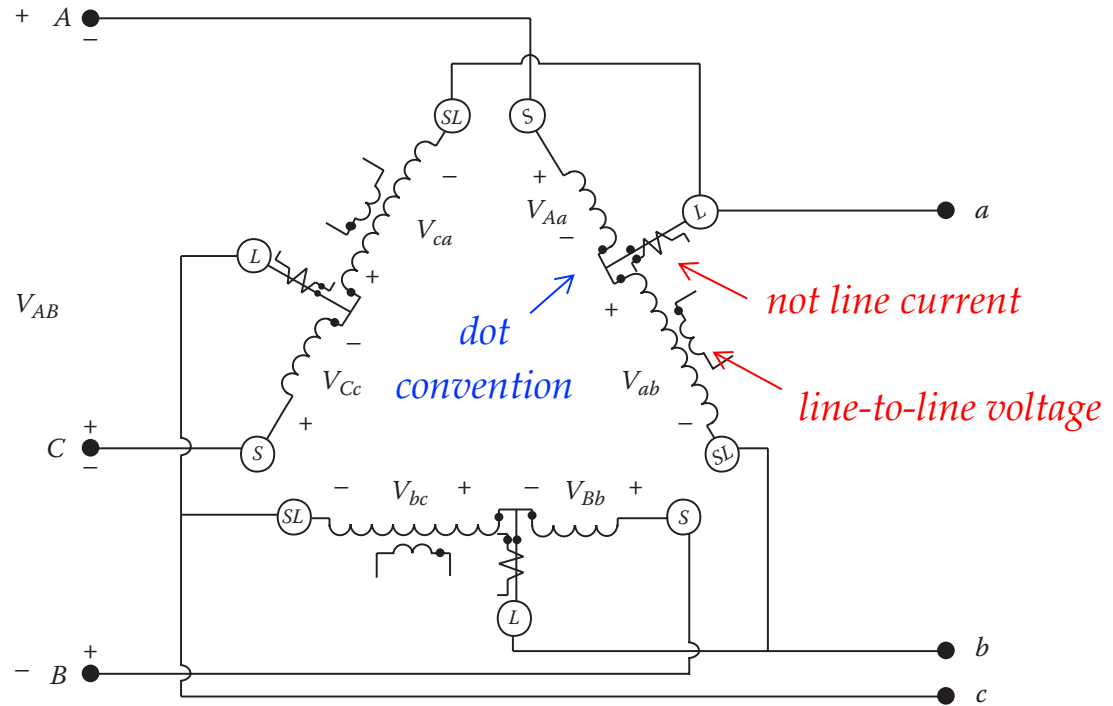
- *E.g.*, point b cannot be connected at both ends of the series winding of second VR

Delta-connected regulators (voltages)

- Used in three-wire delta feeders

- Voltage transformations

$$\begin{aligned}
 V_{AB} &= V_{Aa} + V_{ab} - V_{Bb} \\
 &= -\frac{N_2^{ab}}{N_1^{ab}} V_{ab} + V_{ab} - \left(-\frac{N_2^{bc}}{N_1^{bc}} V_{bc} \right) \\
 &= \left(1 - \frac{N_2^{ab}}{N_1^{ab}} \right) V_{ab} + \frac{N_2^{bc}}{N_1^{bc}} V_{bc} \\
 &= a_R^{ab} V_{ab} + (1 - a_R^{bc}) V_{bc}
 \end{aligned}$$



recall that $a_R^{ab} = 1 \mp \frac{N_2^{ab}}{N_1^{ab}}$

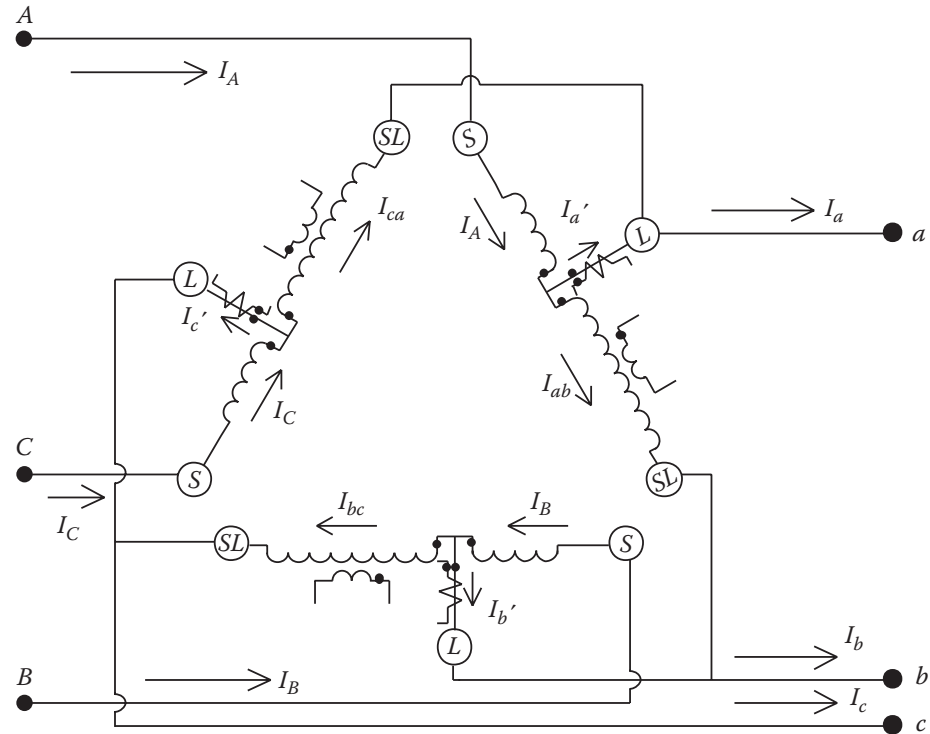
- Repeat for other LL loops to get AB model

$$\mathbf{v}_{\Phi,LL} = \begin{bmatrix} a_R^{ab} & 1 - a_R^{bc} & 0 \\ 0 & a_R^{bc} & 1 - a_R^{ca} \\ 1 - a_R^{ab} & 0 & a_R^{ca} \end{bmatrix} \mathbf{v}_{\phi,LL}$$

Delta-connected regulators (currents)

- Current transformations

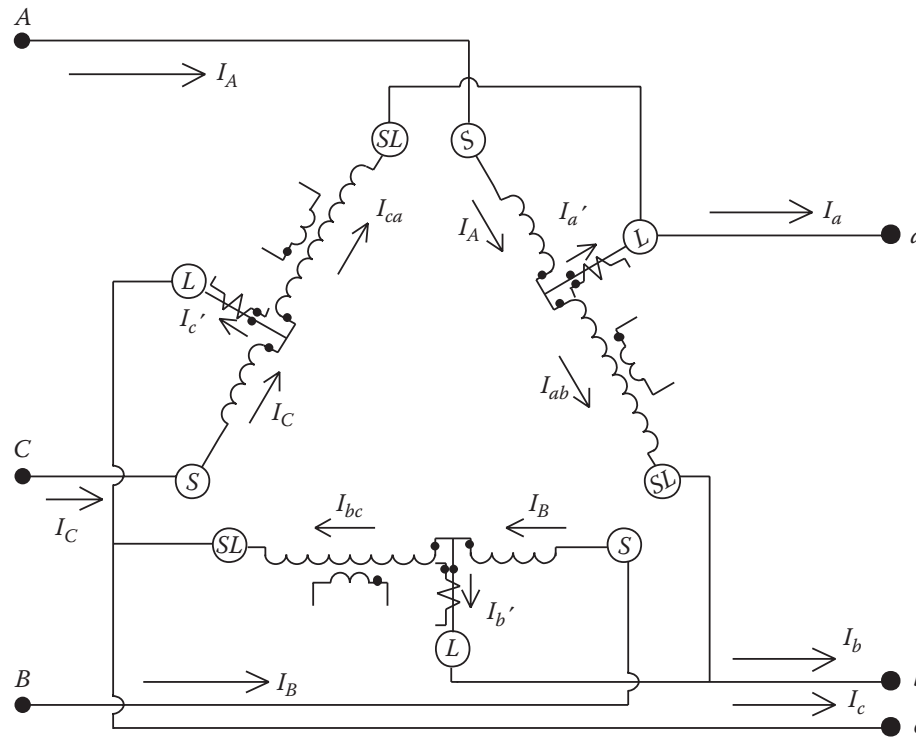
$$\begin{aligned}
 I_a &= I_{ca} + I_A - I_{ab} \\
 &= \frac{N_2^{ca}}{N_1^{ca}} I_C + I_A - \frac{N_2^{ab}}{N_1^{ab}} I_A \\
 &= \left(1 - \frac{N_2^{ab}}{N_1^{ab}}\right) I_A + \frac{N_2^{ca}}{N_1^{ca}} I_C \\
 &= a_R^{ab} I_A + (1 - a_R^{ca}) I_C
 \end{aligned}$$



- Repeat for other load nodes to get the CD *after inversion*

$$\mathbf{i}_\ell = \begin{bmatrix} a_R^{ab} & 0 & 1 - a_R^{ca} \\ 1 - a_R^{ab} & a_R^{bc} & 0 \\ 0 & 1 - a_R^{bc} & a_R^{ca} \end{bmatrix} \mathbf{i}_L$$

Closed delta SVR (ABCD)



- Generalized matrices $\mathbf{v}_{\Phi,LL} = \mathbf{A}\mathbf{v}_{\phi,LL} + \mathbf{B}\mathbf{i}_{\ell}$
 $\mathbf{i}_L = \mathbf{C}\mathbf{v}_{\phi,LL} + \mathbf{D}\mathbf{i}_{\ell}$

$$\mathbf{A} = \begin{bmatrix} a_R^{ab} & 1 - a_R^{bc} & 0 \\ 0 & a_R^{bc} & 1 - a_R^{ca} \\ 1 - a_R^{ab} & 0 & a_R^{ca} \end{bmatrix}; \mathbf{B} = \mathbf{C} = \mathbf{0}; \mathbf{D} = \begin{bmatrix} a_R^{ab} & 0 & 1 - a_R^{ca} \\ 1 - a_R^{ab} & a_R^{bc} & 0 \\ 0 & 1 - a_R^{bc} & a_R^{ca} \end{bmatrix}^{-1}$$

- Harder to control due to coupling in voltages and currents

Open delta SVR

- Regulate voltage in 3-wire delta system with only two regulators
- Voltage transformations

$$V_{AB} = V_{Aa} + V_{ab}$$

$$= \left(1 - \frac{N_2^{ab}}{N_1^{ab}}\right) V_{ab} = a_R^{ab} V_{ab}$$

$$V_{BC} = a_R^{bc} V_{bc}$$

$$V_{CA} = -(V_{AB} + V_{BC})$$

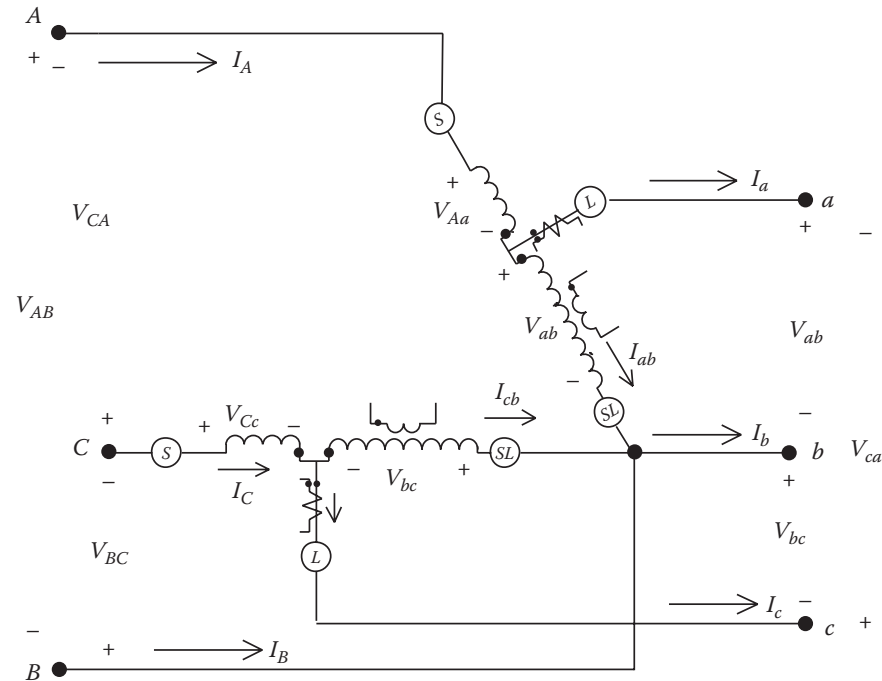
$$= -a_R^{ab} V_{ab} - a_R^{bc} V_{bc}$$

- Current transformations

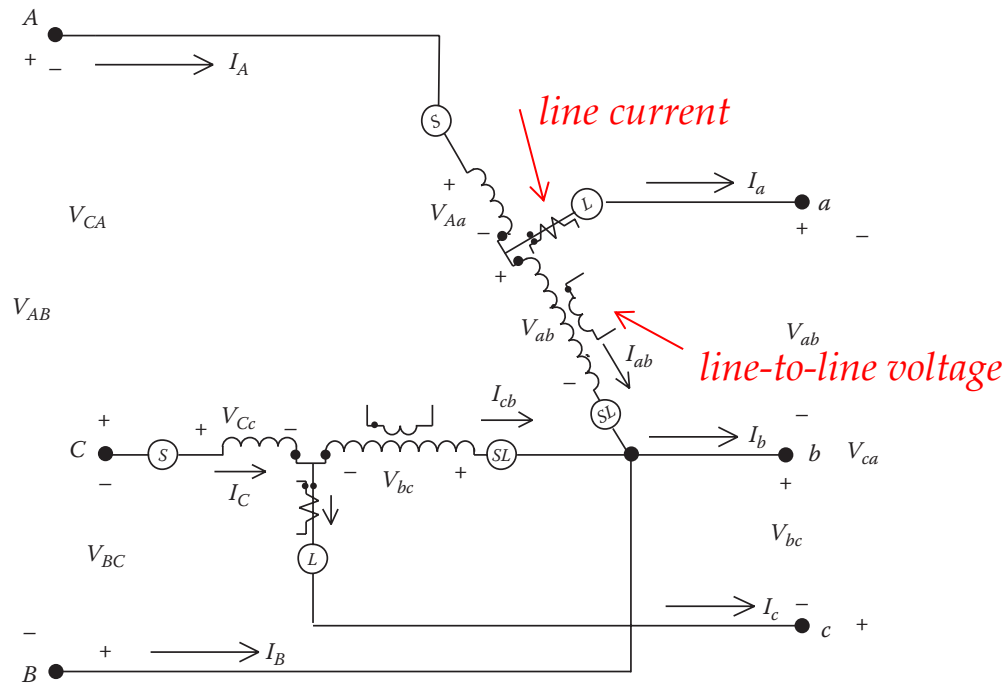
$$I_A = I_a + I_{ab} = I_a + \frac{N_2^{ab}}{N_1^{ab}} I_a \Rightarrow I_A = \frac{1}{a_R^{ab}} I_a$$

similarly: $I_C = \frac{1}{a_R^{cb}} I_c$

$$I_B = -(I_A + I_B)$$



Open delta SVR (cont'd)



- Generalized matrices $\mathbf{v}_{\Phi,LL} = \mathbf{A}\mathbf{v}_{\phi,LL} + \mathbf{B}\mathbf{i}_{\ell}$
 $\mathbf{i}_L = \mathbf{C}\mathbf{v}_{\phi,LL} + \mathbf{D}\mathbf{i}_{\ell}$

$$\mathbf{A} = \begin{bmatrix} a_R^{ab} & 0 & 0 \\ 0 & a_R^{cb} & 0 \\ -a_R^{ab} & -a_R^{cb} & 0 \end{bmatrix}; \mathbf{B} = \mathbf{C} = \mathbf{0}; \mathbf{D} = \begin{bmatrix} 1/a_R^{ab} & 0 & 0 \\ -1/a_R^{ab} & 0 & -1/a_R^{cb} \\ 0 & 0 & 1/a_R^{cb} \end{bmatrix}$$

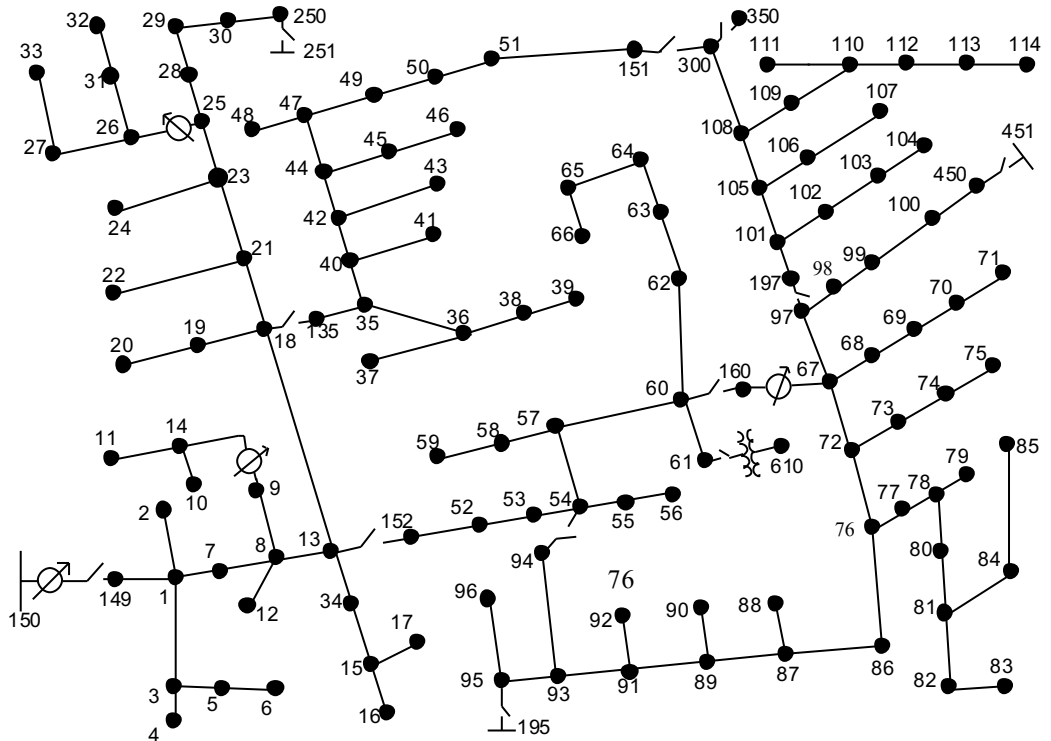
Examples from IEEE 123- bus feeder

Regulator ID:	3	
Line Segment:	25 - 26	
Location:	25	
Phases:	A-C	
Connection:	2-Ph,L-G	
Monitoring Phase:	A & C	
Bandwidth:	1	
PT Ratio:	20	
Primary CT Rating:	50	
Compensator:	Ph-A	Ph-C
R - Setting:	0.4	0.4
X - Setting:	0.4	0.4
Voltage Level:	120	120

Regulator ID:	2
Line Segment:	9 - 14
Location:	9
Phases:	A
Connection:	1-Ph, L-G
Monitoring Phase:	A
Bandwidth:	2.0 volts
PT Ratio:	20
Primary CT Rating:	50
Compensator:	Ph-A
R - Setting:	0.4
X - Setting:	0.4
Voltage Level:	120

Regulator ID:	1
Line Segment:	150 - 149
Location:	150
Phases:	A-B-C
Connection:	3-Ph, Wye
Monitoring Phase:	A
Bandwidth:	2.0 volts
PT Ratio:	20
Primary CT Rating:	700
Compensator:	Ph-A
R - Setting:	3
X - Setting:	7.5
Voltage Level:	120

Regulator ID:	4		
Line Segment:	160 - 67		
Location:	160		
Phases:	A-B-C		
Connection:	3-Ph, LG		
Monitoring Phase:	A-B-C		
Bandwidth:	2		
PT Ratio:	20		
Primary CT Rating:	300		
Compensator:	Ph-A	Ph-B	Ph-C
R - Setting:	0.6	1.4	0.2
X - Setting:	1.3	2.6	1.4
Voltage Level:	124	124	124



Summary

- Voltage regulation is challenging in distribution systems
- Voltage control mechanisms
 - OLTC
 - in-line voltage regulators
 - capacitors
 - (smart inverters)
- Control of SVR can be performed based on
 - local readings
 - remote readings
 - LDC circuit
 - centrally computed voltage setpoints
- Derived ABCD models for single- and three-phase SVRs