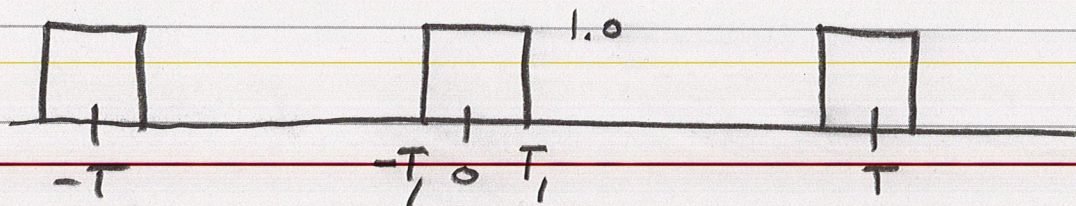


Continuous Time Fourier Transform (Week 9 - Monday)

1/7

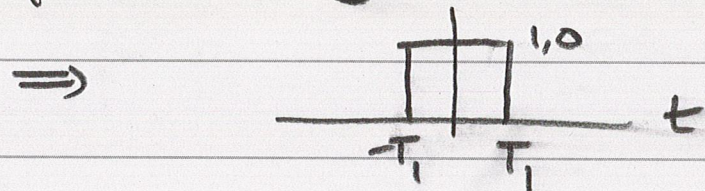
Approach: Start with F.S. and look at case: period $\rightarrow \infty$

Remember the example from before



Found the F.S. coeffs: $a_k = \frac{2 \sin(k\omega_0 T_1)}{k\omega_0 T}$ $k \in \mathbb{Z}$
 $\omega_0 = 2\pi/T$

- An aperiodic signal is the limit of a periodic signal as $T \rightarrow \infty$



this single pulse might be thought of as limit $T \rightarrow \infty$ of the pulse train

- F.S. coeffs are samples of

$$\left. \frac{2 \sin \omega T_1}{\omega} \right\} \text{ this is indep. of } T$$

- $a_k T = \left. \frac{2 \sin \omega T_1}{\omega} \right|_{\omega = k\omega_0}$

Two ways of looking at

$$T a_k = \frac{2 \sin(k \omega_0 T_1)}{k \omega_0}$$

as T grows

$$(\omega_0 = 2\pi/T)$$

Plotted as
function of
index k

(this is actually
a plot of a_k
(not $T a_k$)

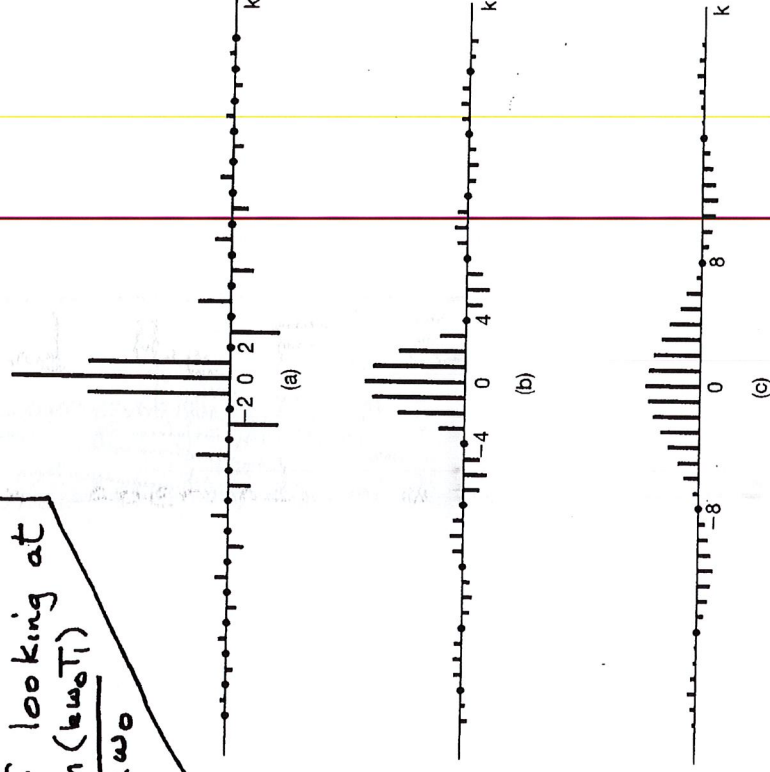


Figure 3.7 Plots of the scaled Fourier series coefficients $T a_k$ for the periodic square wave with T_1 fixed and for several values of T : (a) $T = 4 T_1$; (b) $T = 8 T_1$; (c) $T = 16 T_1$. The coefficients are regularly spaced samples of the envelope $(2 \sin \omega T_1)/\omega$, where the spacing between samples, $2\pi/T$, decreases as T increases.

Plotted as
samples of
envelope

$$T a_k \approx \frac{2 \sin \omega T_1}{\omega} \Big|_{\omega = k \omega_0}$$

sample spacing
is

$$\omega_0 = \frac{2\pi}{T}$$

i.e. samples
get closer
together as
 T increases.

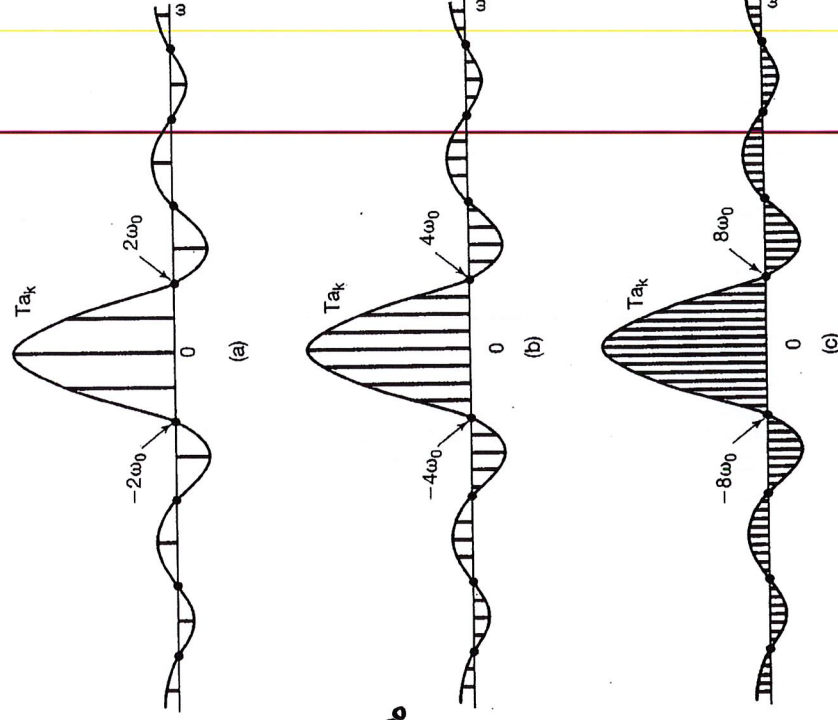


Figure 4.2 The Fourier series coefficients and their envelope for the periodic square wave in Figure 4.1 for several values of T (with T_1 fixed): (a) $T = 4 T_1$; (b) $T = 8 T_1$; (c) $T = 16 T_1$.

Look at the
matlab files
from 10/5 Lect.

Can use the prev. approx and idea of Riemann Sum as def. of an integral to derive form of F.T. $\longrightarrow$ see Text.

For our purposes: $x(t)$ $t \in \mathbb{R}$ aperiodic

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

this is the CTFT
also called the spectrum
of $x(t)$.

Think of this as a function of ω .

The inverse F.T.

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{+j\omega t} d\omega$$

$$x(t) \longleftrightarrow X(j\omega)$$

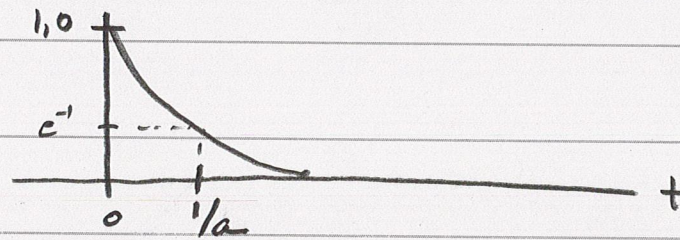
Freq variable is ω
The relationship between
radian freq and Hertzian
freq is

$$2\pi f = \omega$$

Example

$$x(t) = e^{-at} u(t) \quad a > 0$$

4/7



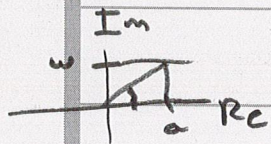
$$X(j\omega) = \int_{-\infty}^{\infty} e^{-at} u(t) e^{-j\omega t} dt = \int_0^{\infty} e^{-at} e^{-j\omega t} dt$$

$$= \int_0^{\infty} e^{-(a+j\omega)t} dt = \left. \frac{1}{-(a+j\omega)} e^{-(a+j\omega)t} \right|_{t=0}^{\infty}$$

$$= \frac{1}{a+j\omega}$$

$\Rightarrow$ If want a plot, then two ways to do so

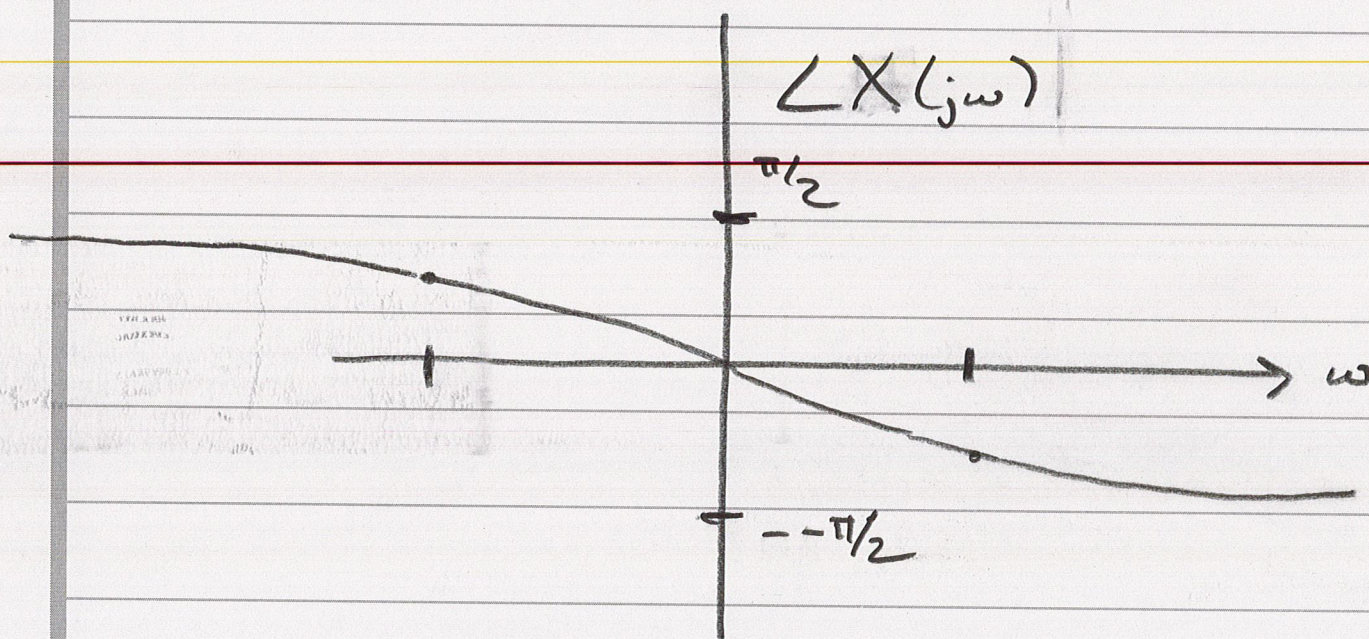
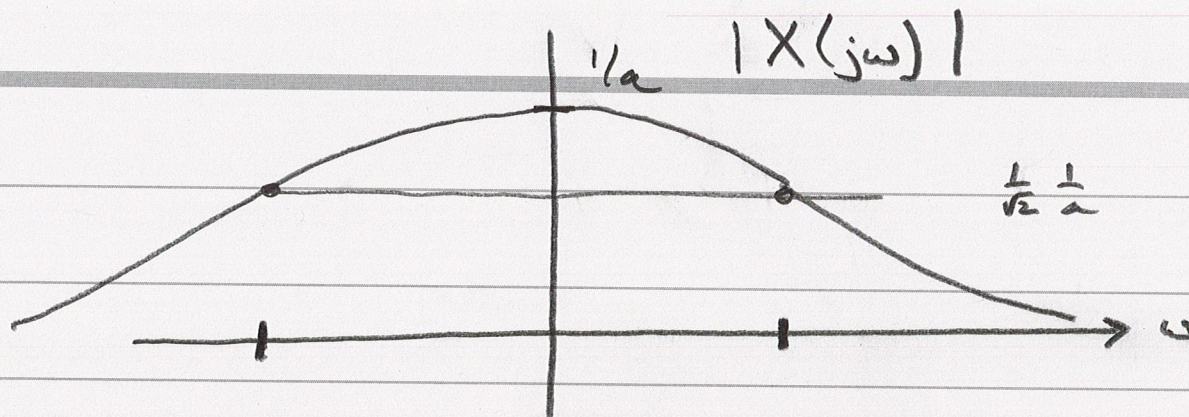
- ① plot real, imag vs. ω separately
- ② plot $|X(j\omega)|$, $\angle X(j\omega)$ vs. ω sep.



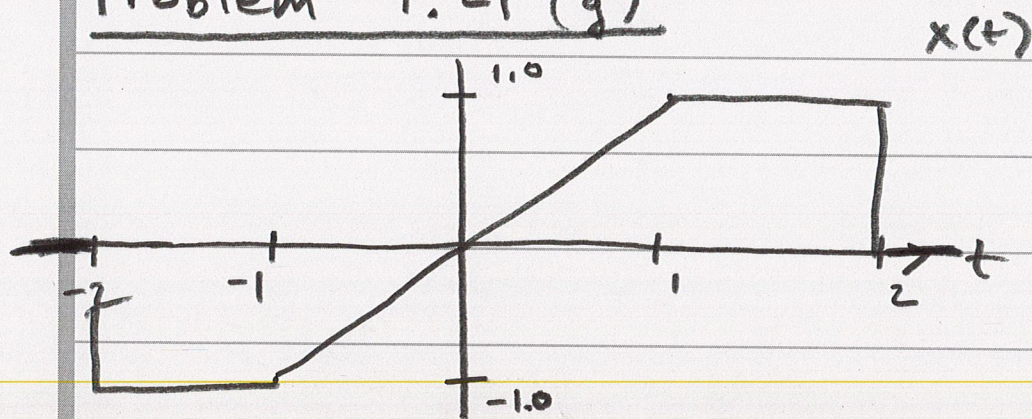
$$|X(j\omega)| = \frac{1}{\sqrt{a^2 + \omega^2}}$$

$$\angle X(j\omega) = -\tan^{-1}(\omega/a)$$

$$\omega = a \quad \frac{1}{\sqrt{2}a} = \frac{1}{\sqrt{2}} \frac{1}{a}$$



Problem 4.21 (g)



$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt = \underbrace{-\int_{-2}^{-1} e^{-j\omega t} dt}_{\frac{1}{j\omega} [e^{j\omega} - e^{j2\omega}]} + \underbrace{\int_{-1}^1 t e^{-j\omega t} dt}_{-1} + \underbrace{\int_1^2 e^{-j\omega t} dt}_{\frac{1}{j\omega} [e^{-j2\omega} - e^{-j\omega}]}$$

use integration by parts.

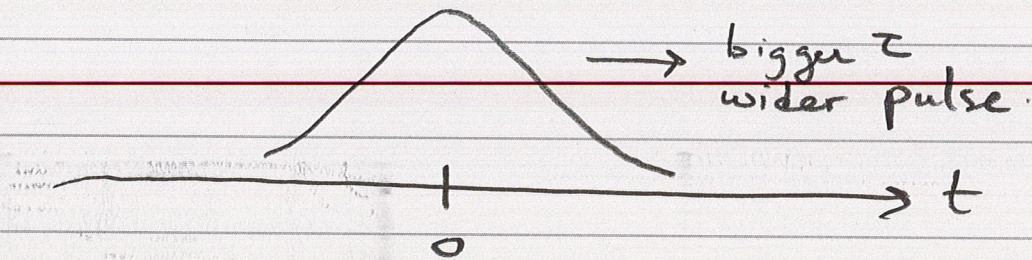
$$= \frac{1}{-j\omega} [e^{-j\omega} + e^{j\omega}] + \left(\frac{1}{\omega^2}\right) [e^{-j\omega} - e^{+j\omega}]$$

If combine + simplify :

$$X(j\omega) = \frac{j^2}{\omega} \left[\cos(2\omega) - \frac{\sin \omega}{\omega} \right]$$

Think about the Gaussian Pulse

$$x(t) = \frac{1}{\sqrt{2\pi}\tau} e^{-t^2/2\tau^2} \quad t \in \mathbb{R}$$



Compute $X(j\omega)$??

$$X(j\omega) = \int_{-\infty}^{\infty} \frac{1}{\sqrt{2\pi}\tau} e^{-t^2/2\tau^2} e^{-j\omega t} dt$$

use "completion of the square"