

HW # 11.

Solutions

ECE 606, Fall 2012

$$\begin{aligned} \underline{a} \quad V_{fb} &= \phi_{ms} - \frac{Q_F}{C_{ox}} \\ &= -1 - \frac{8 \times 10^{-8} \text{ C/cm}^2}{2 \times 10^{-7} \text{ F/cm}^2} = \underline{\underline{-1.40 \text{ V}}} \end{aligned}$$

$$\underline{b} \quad \text{Electric field in oxide} \quad E_{ox} = -\frac{Q_{semi}}{t_{ox}}$$

$Q_{semi}$  = total charge density on semiconductor side of oxide

$$\therefore Q_{semi} = Q_F + Q_{IT} + Q_D + Q_N$$

$$Q_{IT} = 0 \text{ (stated)}$$

$$\text{At flat-band } Q_D = Q_N = 0$$

$$\therefore Q_{semi} = Q_F =$$

$$E_{ox} = \frac{-Q_F}{\epsilon_{ox}} = -2.318 \times 10^5 \text{ V/cm}$$

c To make  $\epsilon_{ox} \rightarrow 0$ , the total charge on the semiconductor side of oxide must be zero.

$$\text{In general } Q_{semi} = Q_N + Q_D + Q_F = 0$$

If we assume inversion.

$$Q_D = -q N_A x_d = -q N_A \sqrt{2 \epsilon_s \phi_s / q N_A}$$

$$\phi_s = 2\phi_F, \quad \phi_F = \frac{kT}{q} \ln\left(\frac{N_A}{n_i}\right) = 0.3578 \text{ V}$$

$$Q_D = -\sqrt{2q \epsilon_s N_A} (2\phi_F) = -4.890 \times 10^{-8} \text{ C/cm}^2$$

Since this charge is less than  $Q_F$ , the remainder of the semiconductor charge must be in the inversion layer

$$\therefore |Q_N| > 0 \quad \text{and surface potential } \phi_s = 2\phi_F = 2 \times 0.3578 \\ = \underline{\underline{0.7156 \text{ V}}}$$

17.2  
(a)

$$\phi_F = \frac{kT}{q} \ln(N_A/n_i) = 0.0259 \ln(10^{15}/10^{10}) = 0.298V$$

$$V_T = 2\phi_F + \frac{K_S x_0}{K_O} \sqrt{\frac{4qN_A}{K_S \epsilon_0} \phi_F} \quad \dots(17.1a)$$

$$= (2)(0.298) + \frac{(11.8)(5 \times 10^{-6})}{(3.9)} \left[ \frac{(4)(1.6 \times 10^{-19})(10^{15})(0.298)}{(11.8)(8.85 \times 10^{-14})} \right]^{1/2}$$

$$V_T = 0.800 V$$

(b) In the square-law theory

$$I_{Dsat} = \frac{Z \bar{\mu}_n C_o}{2L} (V_G - V_T)^2 \quad \dots(17.22)$$

$$C_o = \frac{K_O \epsilon_0}{x_0} = \frac{(3.9)(8.85 \times 10^{-14})}{(5 \times 10^{-6})} = 6.90 \times 10^{-8} \text{ F/cm}^2$$

$$I_{Dsat} = \frac{(5 \times 10^{-3})(800)(6.9 \times 10^{-8})(2 - 0.8)^2}{(2)(5 \times 10^{-4})} = 0.397 \text{ mA}$$

(c) In the bulk-charge theory we must first determine  $V_{Dsat}$  using Eq.(17.29). We know  $\phi_F$  and  $V_T$  from part (a), but must compute  $V_W$  before substituting into the  $V_{Dsat}$  expression.

$$W_T = \left[ \frac{2K_S \epsilon_0}{qN_A} (2\phi_F) \right]^{1/2} = \left[ \frac{(2)(11.8)(8.85 \times 10^{-14})(2)(0.298)}{(1.6 \times 10^{-19})(10^{15})} \right]^{1/2} = 0.882 \mu\text{m}$$

$$V_W = \frac{qN_A W_T}{C_o} = \frac{(1.6 \times 10^{-19})(10^{15})(8.82 \times 10^{-5})}{(6.90 \times 10^{-8})} = 0.205V$$

Noting that  $V_G - V_T = 1.20V$ , substituting into Eq.(17.29) then gives

$$V_{Dsat} = 1.20 - 0.205 \left\{ \left[ \frac{(1.20)}{(2)(0.298)} + \left( 1 + \frac{(0.205)}{(4)(0.298)} \right)^2 \right]^{1/2} - \left[ 1 + \frac{(0.205)}{(4)(0.298)} \right] \right\}$$

or

$$V_{Dsat} = 1.06V \quad \dots \text{smaller than } V_{Dsat} \text{ of square-law theory as expected}$$

Now

$$\frac{Z \bar{\mu}_n C_o}{L} = \frac{(5 \times 10^{-3})(800)(6.90 \times 10^{-8})}{(5 \times 10^{-4})} = 5.52 \times 10^{-4} \text{ amps/V}^2$$

Finally, substituting into Eq.(17.28) gives  $I_{Dsat}$  if  $V_D = V_{Dsat}$ . Thus

$$I_{Dsat} = (5.52 \times 10^{-4}) \left\{ (1.20)(1.06) - \frac{(1.06)^2}{2} - \frac{4}{3} (0.205)(0.298) \left[ \left( 1 + \frac{(1.06)}{(2)(0.298)} \right)^{3/2} - \left( 1 + \frac{(3)(1.06)}{(4)(0.298)} \right) \right] \right\}$$

$$I_{Dsat} = 0.349 \text{ mA} \quad \Leftarrow \text{bulk charge result (smaller than the square-law result as expected)}$$

(d) Clearly here the device is biased below pinch-off. From Table 17.1 we note that both the square-law and bulk-charge theories reduce to the same result if  $V_D = 0$ .

$$g_d = \frac{Z \bar{\mu}_n C_o}{L} (V_G - V_T) = (5.52 \times 10^{-4})(2 - 0.8) = 0.662 \text{ mS}$$

(e) In the square-law theory,  $V_{Dsat} = V_G - V_T$ . Thus  $V_{Dsat} = 1.20V$  and  $V_D = 2V$ . Since  $V_D > V_{Dsat}$  the device is saturation (above-pinch-off) biased, and from Table 17.1

$$g_m = \frac{Z \bar{\mu}_n C_o}{L} (V_G - V_T) = 0.662 \text{ mS} \quad \dots \text{same as } g_d \text{ of part (d)}$$

(f) In part (c) we calculated the bulk-charge  $V_{Dsat} = 1.06V$ . Since  $V_D > V_{Dsat}$ , the device is above-pinch-off biased, and from Table 17.1

$$g_m = \frac{Z \bar{\mu}_n C_o}{L} V_{Dsat} = (5.52 \times 10^{-4})(1.06) = 0.585 \text{ mS}$$

(g) For the applied  $V_G = 2V$ ,  $V_{Dsat} = 1.20V$  in the square-law theory and  $V_{Dsat} = 1.06V$  in the bulk-charge theory. Since in either case  $V_D < V_{Dsat}$ , we can utilize the second form of Eq.(17.37).

$$f_{max} = \frac{\bar{\mu}_n V_D}{2\pi L^2} = \frac{(800)(1)}{(2\pi)(5 \times 10^{-4})^2} = 509 \text{ MHz}$$

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17.9

(a) From Fig. P17.9 we note in general that

$$V_G = V_D + V_B \quad \text{or} \quad V_D = V_G - V_B$$

In the square-law formulation  $V_{Dsat} = V_G - V_T$ . If  $V_B = V_T/2$ , then  $V_D = V_G - V_T/2 > V_{Dsat}$  and the MOSFET is *always biased into saturation*. Noting  $I_D = 0$  if  $V_G < V_T$  or  $V_D < V_T/2$ , and using Eq.(17.22), we conclude

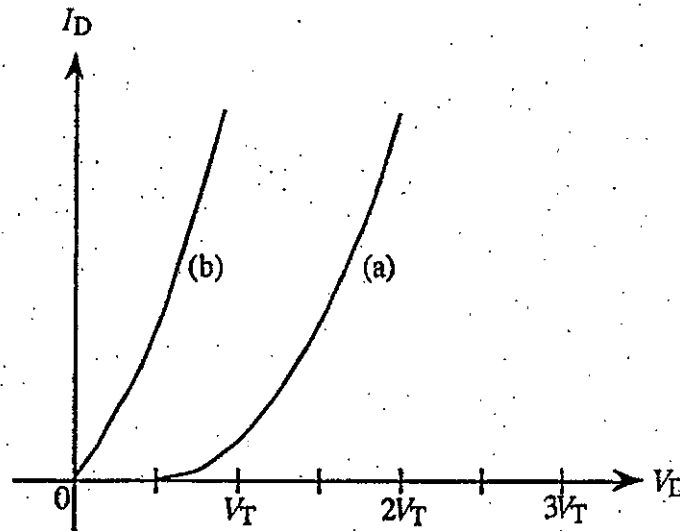
$$I_D = \frac{Z\bar{\mu}_n C_o}{2L} (V_G - V_T)^2 = \frac{Z\bar{\mu}_n C_o}{2L} (V_D - V_T/2)^2 \quad \dots V_D > V_T/2$$

and

$$I_D = 0 \quad \dots V_D < V_T/2$$

(b) If  $V_B = 2V_T$ , then  $V_D = V_G - 2V_T < V_G - V_T = V_{Dsat}$  and the MOSFET is always biased in the linear region of operation. The device turns on for  $V_G > V_T$  or  $V_D > -V_T$  and is therefore on for all  $V_D \geq 0$ . Using Eq.(17.17) we obtain

$$\begin{aligned} I_D &= \frac{Z\bar{\mu}_n C_o}{L} [(V_G - V_T)V_D - V_D^2/2] = \frac{Z\bar{\mu}_n C_o}{L} [(V_D + V_T)V_D - V_D^2/2] \\ &= \frac{Z\bar{\mu}_n C_o}{L} (V_D^2/2 + V_T V_D) = \frac{Z\bar{\mu}_n C_o}{2L} [(V_D + V_T)^2 - V_T^2] \quad \dots V_D \geq 0 \end{aligned}$$



Note that both curves have the same general shape; the part (b) curve is simply shifted to the left and displaced downward.

Q4 A smaller channel length is impacted by depletion region caused by drain-source bias. The channel length is shortened by the depletion region making it easier to turn-on the transistor. Use of shallow doping is one effective way to reduce this problem.

Q5 Sub-threshold swing can be calculated as follows.

$$\frac{10 I_D}{I_D} = \frac{e^{qV_{GS}'/\eta kT}}{e^{qV_{GS}/\eta kT}}$$

$$\Rightarrow 10 = e^{q\Delta V_{GS}/\eta kT}$$

$$\Rightarrow \Delta V_{GS} = \frac{\eta kT}{q} \ln 10 = \eta \times 60 \text{ mV/decade}$$

Since  $\eta \geq 1$ , the minimum value of sub-threshold swing must be at least 60 mV/decade.