

EE538

DSP I

Module 25

Outline

- Extrapolation of the autocorrelation for a SoS signal
- Basics of Autoregressive (AR) spectral estimation

Test 3- Session 26

Same format as previous exams. , e.g. open book

Problems Topics:

1. Radix-2 FFT
2. DFT and time-domain aliasing
3. Windows and symmetric linear-phase FIR Filter

- problem with nonparametric spectral estimators:
inherently assume

$$r_{xx}[m] = 0 \text{ for } m > M$$

where: $M \leq N-1$

- in fact, though, for SoS (w/o noise)

$$r_{xx}[m] = \sum_{i=1}^P P_i e^{j\omega_i m}$$

- does not decay to zero as
 $m \rightarrow \infty$

• in general, sharp spectral peaks in $S_{xx}(\omega)$ imply $r_{xx}[m]$ takes a "long time" to decay to zero as $m \rightarrow \infty$

• interesting observation re: $r_{xx}[m]$ for SOS model \Rightarrow it satisfies recursive relationship

$$r_{xx}[m] = - \sum_{k=1}^P a_k r_{xx}[m-k] \quad \text{for } m \geq P$$

• where: $a_k, k=1, \dots, P$ are the coefficients of the polynomial

$$(z - e^{j\omega_1})(z - e^{j\omega_2}) \dots (z - e^{j\omega_P}) \\ = z^P + a_1 z^{P-1} + a_2 z^{P-2} + \dots + a_{P-1} z + a_P$$

$$1 + a_1 z^{-1} + a_2 z^{-2} + \dots + a_{P-1} z^{-(P-1)} + a_P z^{-P} = 0$$

for $z_i = e^{j\omega_i}$

$$\left\{ \begin{aligned} &1 + a_1 e^{j\omega_i(-1)} + a_2 e^{j\omega_i(-2)} \\ &+ \dots + a_{p-1} e^{j\omega_i(-(p-1))} + a_p e^{j\omega_i(-p)} \end{aligned} \right\} = 0$$

• multiply both sides by $P_i e^{j\omega_i m}$

• sum over all i (since $\sum 0 = 0$)

$$\begin{aligned} &\sum_{i=1}^p P_i e^{j\omega_i m} + a_1 \sum_{i=1}^p P_i e^{j\omega_i (m-1)} \\ &+ \dots + a_{p-1} \sum_{i=1}^p P_i e^{j\omega_i (m-(p-1))} + a_p \sum_{i=1}^p P_i e^{j\omega_i (m-p)} = 0 \end{aligned}$$

- ultimately:

$$r_{xx}[m] = - \sum_{k=1}^p a_k r_{xx}[m-k] \quad \text{for } m \geq p$$

- End of Proof

- Thus: if:

- if we estimate $r_{xx}[m]$ for
 $m = 0, 1, \dots, P$

- determine the coefficients
 $a_k, k = 1, \dots, P$

• we could extrapolate $r_{xx}[m]$ for all m , thereby averting a windowing effect

• ostensibly use recursion to determine $r_{xx}[m]$ for $m \geq P$ and

form
$$S_{xx}(\omega) = \sum_{m=-\infty}^{\infty} r_{xx}[m] e^{-j\omega m}$$

• however, in case of SoS, typically interested in frequencies, ω_i , which may be determined from roots of $z^p + a_1 z^{p-1} + \dots + a_{p-1} z + a_p$

• note: $r_{xx}[-m] = r_{xx}^*[m]$

• How do we estimate
 $a_k, k=1, \dots, P$ given

$r_{xx}[m], m=0, 1, \dots, P$

• $r_{xx}[m] = -\sum_{k=1}^P a_k r_{xx}[m-k]$

• write out P eqns. in

P unknowns $\Rightarrow m=1, 2, \dots, P$

$$\begin{bmatrix}
 r_{xx}[0] & r_{xx}[-1] & \dots & r_{xx}[-(P-1)] \\
 r_{xx}[1] & r_{xx}[0] & \dots & r_{xx}[-(P-2)] \\
 \vdots & \vdots & \dots & \vdots \\
 r_{xx}[P-1] & r_{xx}[P-2] & \dots & r_{xx}[0]
 \end{bmatrix}
 \begin{bmatrix}
 a_1 \\
 a_2 \\
 \vdots \\
 a_P
 \end{bmatrix}
 =
 \begin{bmatrix}
 r_{xx}[1] \\
 r_{xx}[2] \\
 \vdots \\
 r_{xx}[P]
 \end{bmatrix}$$

$$\underbrace{\begin{bmatrix} R \\ \vdots \\ R \end{bmatrix}}_{P \times P} \underbrace{\begin{bmatrix} a \\ \vdots \\ a \end{bmatrix}}_{P \times 1} = - \underbrace{\begin{bmatrix} r \\ \vdots \\ r \end{bmatrix}}_{P \times 1}$$

} see SoS extrop.m at web site

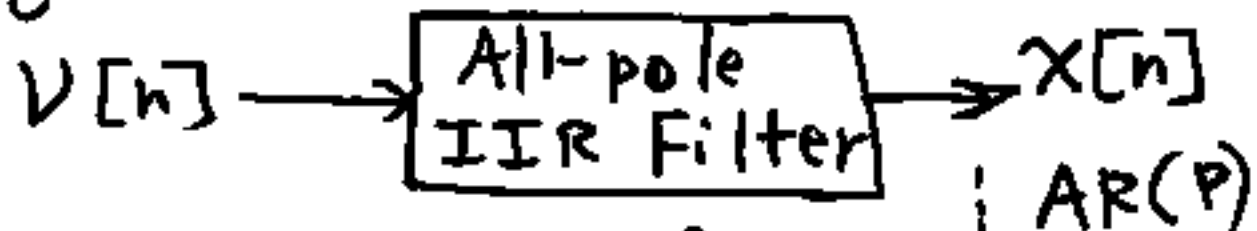
• these observations for the SoS model are the basis for the following spectral estimation techniques:

• Auto Regressive (AR) Method

• MUSIC (Multiple Signal)
- Sect. 12.5.3 Classification

• ESPRIT (Estimation of Signal Parameters by Rotational Invariance Techniques)
Sect. 12.5.4

- AR Spectral Estimation
- for sole purpose of matching the spectral density of a WSS random process, $x[n]$, we model $x[n]$ as having been generated as:



where: $r_{VV}[m] = \sigma_w^2 \delta[m]$ | AR(P) process

- $x[n] = -\sum_{k=1}^P a_k x[n-k] + v[n]$

- can show that $r_{xx}[m]$ satisfies

$$r_{xx}[m] = -\sum_{k=1}^P a_k r_{xx}[m-k] + \sigma_v^2 \delta[m]$$

• in freq. domain:

$$X(\omega) = H(\omega) V(\omega)$$

• where: $H(\omega) = \frac{1}{1 + \sum_{k=1}^p a_k e^{-j k \omega}}$

• from basis random process theory

$$S_{XX}(\omega) = |H(\omega)|^2 S_{VV}(\omega)$$

• where: $S_{yy}(\omega) = \text{DTFT}\{r_{yy}[m]\}$
 $= \text{DTFT}\{\sigma_w^2 \delta[m]\}$
 $= \sigma_w^2$ for all ω

• + thus:

$$S_{xx}(\omega) = \text{DTFT}\{r_{xx}[m]\}$$

$$= \frac{\sigma_w^2}{\left|1 + \sum_{k=1}^p a_k e^{-jk\omega}\right|^2}$$