

**NAME:** **13 December 2016**  
**Digital Signal Processing I** **Final Exam** **Fall 2016**

## Cover Sheet

Test Duration: 120 minutes.

Open Book but Closed Notes. Three 8.5 x 11 crib sheets allowed

Calculators NOT allowed.

This test contains **five** problems.

All work should be done on the blank pages provided.

Your answer to each part of the exam should be clearly labeled.

Great having you in class this semester.

Have a great holiday break!

**Problem 1.** An OFDM signal is synthesized as a sum of four sinewaves of length 4 as

$$\tilde{x}[n] = \frac{1}{4} \sum_{k=0}^3 b_k e^{j2\pi \frac{k}{4}n} \{u[n] - u[n-4]\}$$

Each of the four values  $b_k$  is one of the eight values listed in the symbol alphabet below.

$$b_k \in \{-7, -5, -3, -1, 1, 3, 5, 7\} \quad k = 0, 1, 2, 3$$

We know in advance that the signal we transmit will be convolved with a filter of length  $L = 3$ . Thus, we add a cyclic prefix of length 3 which effectively creates a sum of sinewaves of length 7 as prescribed below:

$$x[n] = \frac{1}{4} \sum_{k=0}^3 b_k e^{j2\pi \frac{k}{4}n} \{u[n+3] - u[n-4]\}$$

The sequence of length 7 above is convolved with the filter below of length 3

$$h[n] = \{2, 4, 1\} = 2\delta[n] + 4\delta[n-1] + \delta[n-2]$$

This ultimately yields the following sequence of length  $7+3-1 = 9$

$$y[n] = x[n] * h[n] = \{2 + 2j, 6 + 4j, 7 - j, 13 - 4j, 19 + j, 10 + 4j, 7 - j, 5 - 4j, 1 - j\}$$

Your task is to determine the numerical values of each of the four symbols:  $b_0$ ,  $b_1$ ,  $b_2$ , and  $b_3$ . Explain and show all the steps in determining your answers. Please lay your work out nicely with logical ordering. Your work will be more important than your final answers.

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**Problem 2.** Problem 1 is modified as follows. The only thing that is changed is that the filter is changed to (only the  $n = 2$  value is changed)

$$h[n] = \{2, 4, 2\} = 2\delta[n] + 4\delta[n - 1] + 2\delta[n - 2]$$

Note that  $h[n]$  may also be expressed as two times the convolution of the sequence  $\{1, 1\}$  with itself as

$$h[n] = \{2, 4, 2\} = 2\{1, 1\} * \{1, 1\}$$

Everything else is the same. We send the same four symbols values  $b_k$  as in Problem 1. We add a cyclic prefix of length 3 which effectively creates a sum of sinewaves of length 7 as below, again, the same as in Problem 1.

$$x[n] = \frac{1}{4} \sum_{k=0}^3 b_k e^{j2\pi\frac{k}{4}n} \{u[n+3] - u[n-4]\}$$

The sequence of length 7 above is convolved with the filter below of length 3

$$h[n] = \{2, 4, 2\} = 2\delta[n] + 4\delta[n - 1] + 2\delta[n - 2]$$

This ultimately yields the following sequence of length  $7+3-1 = 9$

$$y[n] = x[n] * h[n] = \{2 + 2j, 6 + 4j, 8, 14 - 4j, 20, 14 + 4j, 8, 6 - 4j, 2 - 2j\}$$

You do NOT have to determine the values of  $b_k$ ,  $k = 0, 1, 2, 3$ . In fact, given this new filter and given ONLY the values of  $y[n]$  listed above, you will NOT be able to determine one of the values of  $b_k$ ,  $k = 0, 1, 2, 3$ .

**TASK:** Determine which one of the four values  $b_0$ ,  $b_1$ ,  $b_2$ , and  $b_3$  you are unable to determine, and explain why.

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**Problem 3.** [30 points] Consider a single-pole, analog all-pass filter with transfer function (Laplace Transform of impulse response)

$$H_a(s) = \frac{s + p_a^*}{s - p_a}$$

where the subscript  $a$  denotes analog. If one substitutes  $s = j\Omega$  to get the frequency response (Fourier Transform of impulse response)

$$H_a(\Omega) = \frac{j\Omega + p_a^*}{j\Omega - p_a}$$

it is easy to show that the magnitude  $|H_a(\Omega)| = 1$  for all  $\Omega$ .

(a) The transfer function for a digital filter is obtained via the bilinear transform

$$s = \frac{z - 1}{z + 1}$$

We discussed three properties of the bilinear transform in class. Which property of the bilinear transform guarantees us that an all-pass analog filter will be transformed into an all-pass digital filter?

- (b) Consider the case where the pole is located at  $p_a = -0.2 + 0.4j$ . Determine the values of the pole,  $p_d$ , and zero,  $z_d$ , of the digital filter obtained via the bilinear transform.
- (i) Draw a pole-zero diagram for the resulting digital filter.
  - (ii) Is the resulting digital filter stable? Explain your answer.
  - (iii) Is the resulting digital filter an all-pass filter? Explain your answer.
  - (iv) What digital frequency is the analog frequency  $\Omega = \sqrt{3}$  (in radians) mapped to?
  - (v) Determine and write the difference equation for the resulting digital filter.

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**Problem 4.** For all parts of this problem,  $x[n]$  is the fine-length sinewave of length  $L = 8$  with frequency  $\omega_o = \pi/2$  defined below, and  $h[n]$  is a causal filter of length  $M = 4$  which may be expressed in sequence form as  $h[n] = \{1, 2, -2, -1\}$ .

$$x[n] = \cos\left(\frac{\pi}{2}n\right) \{u[n] - u[n-8]\} \qquad h[n] = \{1, 2, -2, -1\}$$

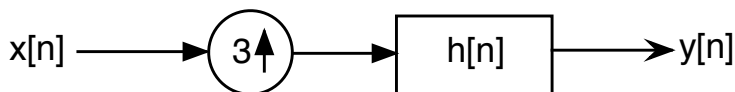
- (a) Compute the linear convolution of  $x[n]$  and  $h[n]$ . Indicate which points are the transient points (partial overlap) at the beginning and end, and also which points are “pure” sinewave (full overlap.)
- (b) With  $X_N(k)$  computed as the 8-pt DFT of  $x[n]$  and  $H_N(k)$  computed as the 8-pt DFT of  $h[n]$ , the product  $Y_N(k) = X_N(k)H_N(k)$  is formed. Determine the  $N = 8$  values of the 8-pt Inverse DFT of  $Y_N(k) = X_N(k)H_N(k)$ .
- (c) Using your answer to (a), explain your answer to (b) by mathematically illustrating the time-domain aliasing effect.
- (d) The product sequence  $Y_N(k) = X_N(k)H_N(k)$ , formed as directly above with  $N = 8$ , is used in Eqn (1). Write a closed-form expression for the reconstructed spectrum  $Y_r(\omega)$ , computed according to Eqn (1) below:

$$Y_r(\omega) = \sum_{k=0}^{N-1} Y_N(k) \frac{\sin\left[\frac{N}{2}\left(\omega - \frac{2\pi k}{N}\right)\right]}{N \sin\left[\frac{1}{2}\left(\omega - \frac{2\pi k}{N}\right)\right]} e^{-j\frac{N-1}{2}\left(\omega - \frac{2\pi k}{N}\right)} \quad (1)$$

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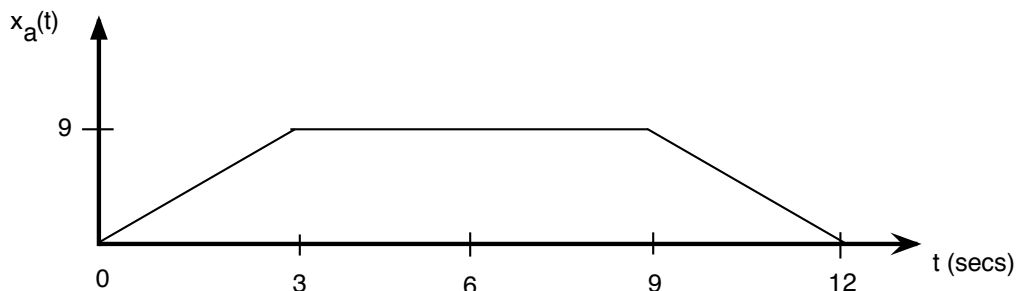
**Problem 5.** Consider the upsampler system below in Figure 1.



**Figure 1.**

- (a) Draw block diagram of efficient implementation of the upsampler system in Fig. 1.
- (b) Your answer to part (a) should involve the polyphase components of  $h[n]$ :  $h_0[n] = h[3n]$ ,  $h_1[n] = h[3n + 1]$ , and  $h_2[n] = h[3n + 2]$  and the DTFT of  $h[n]$ , denoted  $H(\omega)$ .
- (c) Consider that the input to the system in Figure 1 is a sampled version of the analog signal in Figure 2. For the remaining parts of this problem, the input signal is as defined below where  $x_a(t)$  is the analog signal in Figure 2. Assume that  $1/T_s = 1$  is above the Nyquist rate for this signal. That is, even though this signal is not strictly bandlimited, assume that aliasing effects are negligible.

$$x[n] = x_a(nT_s), \quad T_s = 1$$



**Figure 2.**

- (i) For the ideal case where  $h[n] = 3 \frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , determine the output  $y[n]$  of the system in Figure 1, when  $x[n]$  is input to the system. Write output in sequence form (indicating where  $n = 0$  is) OR do stem plot.
- (ii) For the ideal case where  $h[n] = 3 \frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , determine the output  $y_0[n] = x[n] * h_0[n]$ , when  $x[n]$  is input to the filter  $h_0[n] = h[3n]$ . Write output in sequence form (indicating where is  $n = 0$  OR do stem plot.
- (iii) For the ideal case where  $h[n] = 3 \frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , determine the output  $y_1[n] = x[n] * h_1[n]$ , when  $x[n]$  is input to the filter  $h_1[n] = h[3n + 1]$ . Write output in sequence form (indicating where  $n = 0$  is) OR do stem plot.
- (iv) For the ideal case where  $h[n] = 3 \frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , determine the output  $y_2[n] = x[n] * h_2[n]$ , when  $x[n]$  is input to the filter  $h_2[n] = h[3n + 2]$ . Write output in sequence form (indicating where  $n = 0$  is) OR do stem plot.



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