# SOLUTION

## NAME: 28 Oct. 2016 ECE 538 Digital Signal Processing I Exam 2 Fall 2016

## Cover Sheet

#### WRITE YOUR NAME ON THIS COVER SHEET

Test Duration: 60 minutes. Open Book but Closed Notes. One (both sides) handwritten 8.5 in x 11 in crib sheet allowed Calculators NOT allowed. All work should be done in the space provided.

There are three problems.

1. 50 points 2. 30 points 3. 20 points

Continuous-Time Fourier Transform (Hz):  $X(F) = \mathcal{F}\{x(t)\} = \int_{-\infty}^{\infty} x(t)e^{-j2\pi Ft}dt$ Continuous-Time Fourier Transform Pair (Hz):  $\mathcal{F}\left\{\frac{\sin(2\pi Wt)}{\pi t}\right\} = rect\left\{\frac{F}{2W}\right\}$  where rect(x) = 1 for |x| < 0.5 and rect(x) = 0 for |x| > 0.5. Continuous-Time Fourier Transform Property:  $\mathcal{F}\{x_1(t)x_2(t)\} = X_1(F) * X_2(F)$ , where \* denotes convolution, and  $\mathcal{F}\{x_i(t)\} = X_i(F), i = 1, 2$ . Relationship between DTFT and CTFT frequency variables in Hz:  $\omega = 2\pi \frac{F}{F_s}$ , where  $F_s = \frac{1}{T_s}$  is the sampling rate in Hz Problem 1. Consider the upsampler system below in Figure 1.

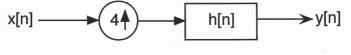
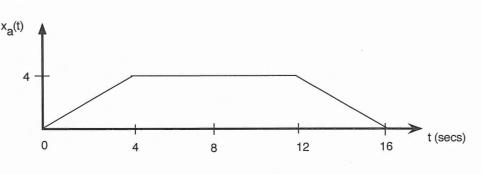


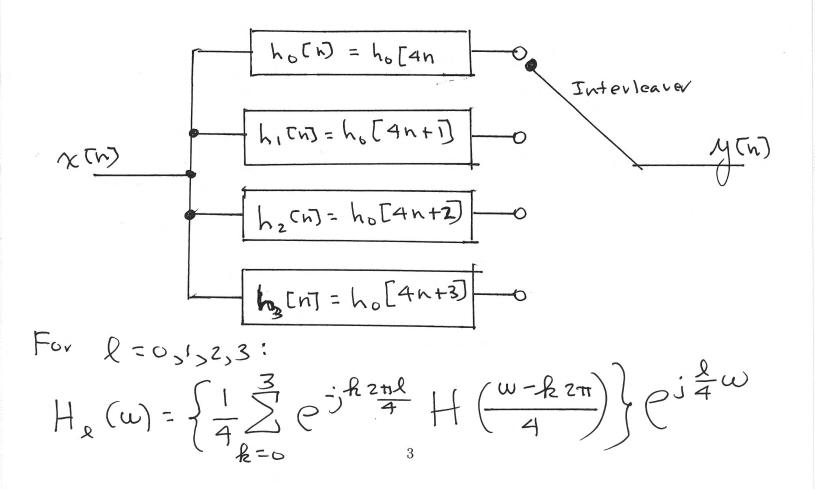
Figure 1.

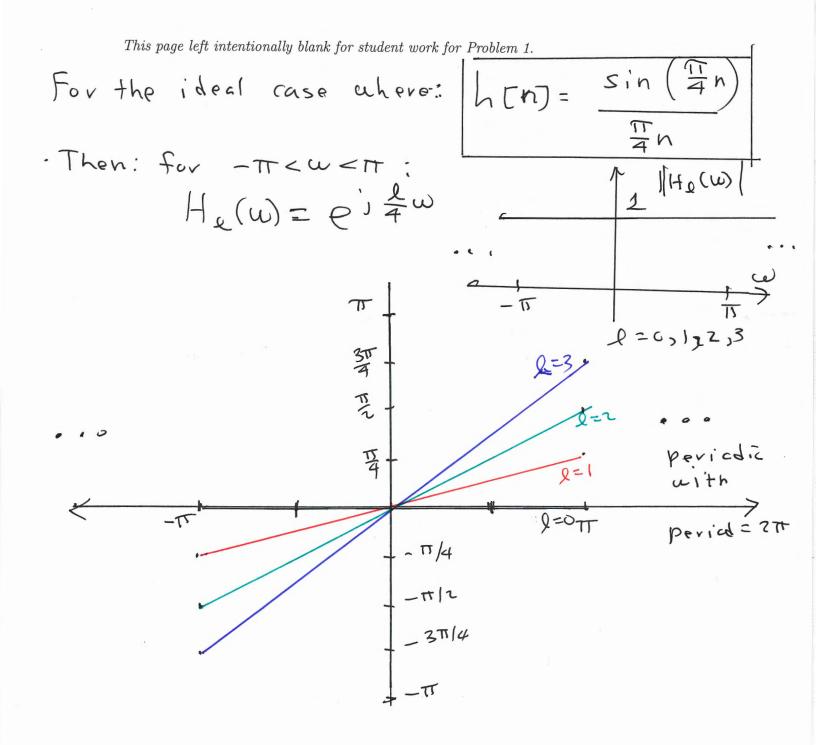
- (a) Draw block diagram of efficient implementation of the upsampler system in Fig. 1.
- (b) Your answer to part (a) should involve the polyphase components of h[n]:  $h_0[n] = h[4n]$ ,  $h_1[n] = h[4n + 1]$ ,  $h_2[n] = h[4n + 2]$ , and  $h_3[n] = h[4n + 3]$ . For the plots requested below, you can do all magnitude plots on one graph and you can do all phase plots on one graph, to save time and space.
  - (i) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , plot the magnitude of the DTFT of  $h_0[n] = h[4n], H_0(\omega)$ , over  $-\pi < \omega < \pi$ .
  - (ii) For the general case where h[n] is an arbitrary impulse response, express the DTFT of  $h_1[n] = h[4n + 1]$ , denoted  $H_1(\omega)$ , in terms of  $H(\omega)$ .
  - (iii) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , plot both the magnitude AND phase of the DTFT  $h_1[n] = h[4n+1]$ ,  $H_1(\omega)$ , over  $-\pi < \omega < \pi$ .
  - (iv) Express the DTFT of  $h_2[n] = h[4n+2]$ , denoted  $H_2(\omega)$ , in terms of  $H(\omega)$ .
  - (v) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , plot both the magnitude AND phase of the DTFT  $h_2[n] = h[4n+2], H_2(\omega)$ , over  $-\pi < \omega < \pi$ .
  - (vi) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , plot both the magnitude AND phase of the DTFT  $h_3[n] = h[4n+3]$ ,  $H_3(\omega)$ , over  $-\pi < \omega < \pi$ .
- (c) Consider that the input to the system in Figure 1 is a sampled version of the analog signal in Figure 2. For the remaining parts of this problem, the input signal is as defined below where  $x_a(t)$  is the analog signal in Figure 2. Assume that  $1/T_s = 1/2$  is above the Nyquist rate for this signal. That is, even though this signal is not strictly bandlimited, assume that aliasing effects are negligible.



$$x[n] = x_a(nT_s), \quad T_s = 2$$

- (i) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , determine the output y[n] of the system in Figure 1, when x[n] is input to the system. Write output in sequence form (indicating where n = 0 is) OR do stem plot.
- (ii) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , determine the output  $y_0[n] = x[n] * h_0[n]$ , when x[n] is input to the filter  $h_0[n] = h[4n]$ . Write output in sequence form (indicating where is n = 0 OR do stem plot.
- (iii) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , determine the output  $y_1[n] = x[n] * h_1[n]$ , when x[n] is input to the filter  $h_1[n] = h[4n+1]$ . Write output in sequence form (indicating where n = 0 is) OR do stem plot.
- (iv) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , determine the output  $y_2[n] = x[n] * h_2[n]$ , when x[n] is input to the filter  $h_2[n] = h[4n+2]$ . Write output in sequence form (indicating where n = 0 is) OR do stem plot.
- (v) For the ideal case where  $h[n] = 4 \frac{\sin(\frac{\pi}{4}n)}{\pi n}$ , determine the output  $y_3[n] = x[n]*h_3[n]$ , when x[n] is input to the filter  $h_3[n] = h[4n+3]$ . Write output in sequence form (indicating where n = 0 is) OR do stem plot.



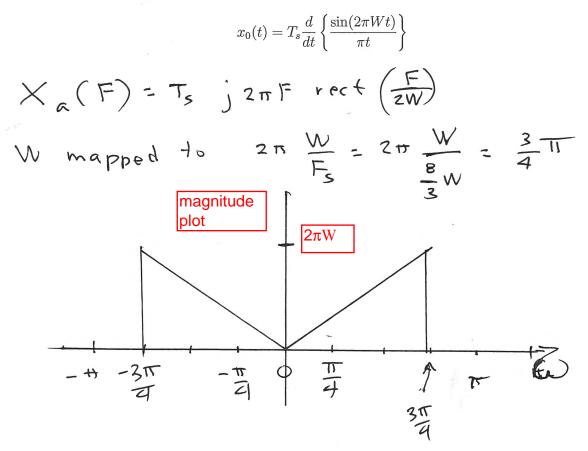


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2(c) X(n) = Xa(n2) => sample every 2 secs.  $\chi$  [n] = {0,2,4,4,4,4,4,2,0} (i) digital upsampling by a factor of 4:=> effectively => sample every  $\frac{z}{4} = \frac{1}{2}$  sec.  $M[n] = \int_{0}^{1} \int_{0}^{1} \int_{0}^{2} \int_{0}^{$ (ii) \$ (Th) = h TAN = Sin (TTN) = S[N] => samples you TTN = S[N] => samples you Yu(n) = X(n) = {(, 2, 4, 4, 4, 4, 4, 7, 0) black  $(iii) h_{2} \ln j = h (4n+1) = \chi_{a} \left( \frac{T_{a}}{4} + h T_{a} \right) = \chi_{a} \left( \frac{1}{2} + h z \right)$ = y [4n+i] ==  $red = \{\frac{1}{2}, \frac{5}{2}, 4, 4, 4, 7, \frac{7}{2}, \frac{3}{2}\}$  $(iv)h_{2}(n) = h[4n+2] = green = \{1,3,4,4,4,3,1\} = \chi_{a}(\frac{2T_{3}}{4}+nT_{3})$ => y[4n+2] = green =  $\{1,3,4,4,4,3,1\} = \chi_{a}(1+n2)$ (v) h<sub>3</sub>(n) = h[an+3] =7 y [4n+3] = blue =  $\{\frac{3}{2}, \frac{7}{2}, \frac{4}{2}, \frac{4}{2}, \frac{4}{2}, \frac{4}{2}, \frac{5}{2}, \frac{1}{2}\}$  $= \chi_{a} \left( \frac{3T_{s}}{4} + nT_{s} \right) = \chi_{a} \left( \frac{3}{2} + hz \right)$ 

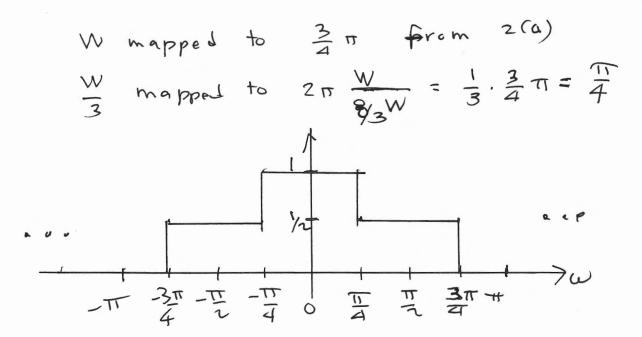
#### EE538 Digital Signal Processing I

Prob. 2(a) Consider the continuous-time signal  $x_0(t)$  below. A discrete-time signal is created by sampling  $x_0(t)$  according to  $x_0[n] = x_0(nT_s)$  with  $F_s = \frac{1}{T_s} = \frac{8}{3}W$ . Plot the magnitude of the DTFT of  $x_0[n]$ ,  $|X_0(\omega)|$ , over  $-\pi < \omega < \pi$ . Show all work.



2(b) Consider the continuous-time signal  $x_1(t)$  below. A discrete-time signal is created by sampling  $x_1(t)$  according to  $x_1[n] = x_1(nT_s)$  with  $F_s = \frac{1}{T_s} = \frac{8}{3}W$ . Plot the magnitude of the DTFT of  $x_1[n]$ ,  $|X_1(\omega)|$ , over  $-\pi < \omega < \pi$ . Show all work.

$$x_1(t) = T_s \frac{1}{2} \left\{ \frac{\sin(2\pi W t)}{\pi t} + \frac{\sin(2\pi \frac{W}{3} t)}{\pi t} \right\}$$



2(c) Consider the continuous-time signal  $x_2(t)$  below. A discrete-time signal is created by sampling  $x_2(t)$  according to  $x_2[n] = x_2(nT_s)$  with  $F_s = \frac{1}{T_s} = 4W$ . Plot the magnitude of the DTFT of  $x_2[n]$ ,  $|X_2(\omega)|$ , over  $-\pi < \omega < \pi$ . Show all work.

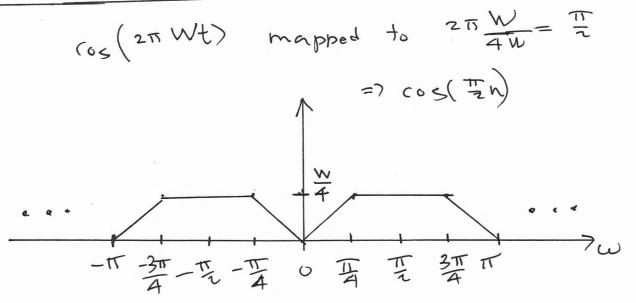
1-

$$x_{2}(t) = T_{s} \left\{ \frac{\sin(2\pi \frac{W}{4}t)}{\pi t} \frac{\sin(2\pi \frac{3W}{4}t)}{\pi t} \right\} \cos(2\pi W t)$$

$$\frac{3W}{4} - \frac{W}{4} = \frac{W}{2} \qquad \frac{W}{4} + \frac{3W}{4} = W$$

$$m \text{ apped to: (before multiply ine by rusine)}$$

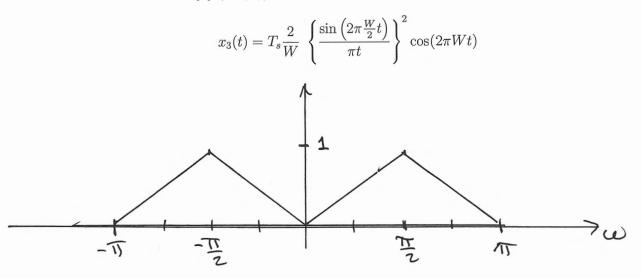
$$2\pi \frac{W}{2} = \frac{1}{4} \qquad 2\pi \frac{W}{4W} = \frac{1}{2}$$

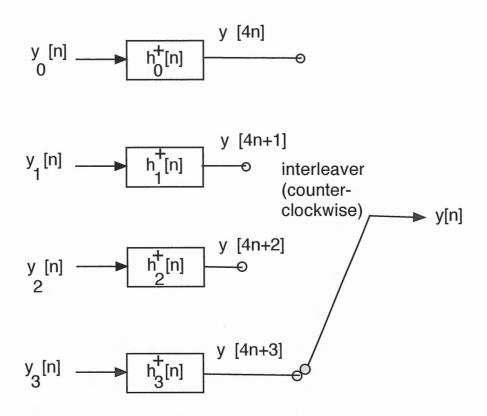


$$\chi[n] = \frac{1}{4W} \frac{\sin\left(\frac{\pi}{2}n\right)}{\frac{\pi}{4W}} \cdot \frac{\sin\left(\frac{3\pi}{2}n\right)}{\frac{\pi}{4W}} \cos\left(\frac{\pi}{2}n\right)$$

$$height = 4W \frac{1}{2} \cdot \frac{\pi}{4} \cdot \frac{1}{2\pi} = \frac{W}{4}$$

2(d) Consider the continuous-time signal  $x_3(t)$  below. A discrete-time signal is created by sampling  $x_3(t)$  according to  $x_3[n] = x_3(nT_s)$  with  $F_s = \frac{1}{T_s} = 4W$ . Plot the magnitude of the DTFT of  $x_3[n]$ ,  $|X_3(\omega)|$ , over  $-\pi < \omega < \pi$ . Show all work.





### Figure 3.

**Problem 2.** This problem is about digital subbanding of the four DT signals  $x_i[n]$ , i = 0, 1, 2, 3 from Problem 2. Digital subbanding of these four signals is effected in an efficient way via filter bank in Figure 3. All of the quantities in Figure 1 are defined below: the respective impulse responses of the polyphase component filters are defined in terms of the ideal lowpass filter impulse response below.

$$h_{LP}[n] = 4 \frac{\sin\left(\frac{\pi}{4}n\right)}{\pi n} \tag{1}$$

The polyphase component filters on the left side of Figure 3 are defined as

$$h_{\ell}^{+}[n] = h_{LP}[4n + \ell], \quad \ell = 0, 1, 2, 3.$$
 (2)

The respective signals at the inputs to these filters are formed from the input signals as (from Problem 3) as described below. There is only ONE part to this problem: plot the magnitude of the DTFT  $Y(\omega)$  of the interleaved signal y[n].

$$y_{0}[n] = x_{0}[n] + x_{1}[n]\cos\left(\frac{\pi}{2}0\right) - \hat{x}_{1}[n]\sin\left(\frac{\pi}{2}0\right) + x_{2}[n]\cos\left(\frac{\pi}{2}0\right) + \hat{x}_{2}[n]\sin\left(\frac{\pi}{2}0\right) + x_{3}[n]\cos(\pi 0)$$

$$y_{1}[n] = x_{0}[n] + x_{1}[n]\cos\left(\frac{\pi}{2}1\right) - \hat{x}_{1}[n]\sin\left(\frac{\pi}{2}1\right) + x_{2}[n]\cos\left(\frac{\pi}{2}1\right) + \hat{x}_{2}[n]\sin\left(\frac{\pi}{2}1\right) + x_{3}[n]\cos(\pi 1)$$

$$y_{2}[n] = x_{0}[n] + x_{1}[n]\cos\left(\frac{\pi}{2}2\right) - \hat{x}_{1}[n]\sin\left(\frac{\pi}{2}2\right) + x_{2}[n]\cos\left(\frac{\pi}{2}2\right) + \hat{x}_{2}[n]\sin\left(\frac{\pi}{2}2\right) + x_{3}[n]\cos(\pi 2)$$

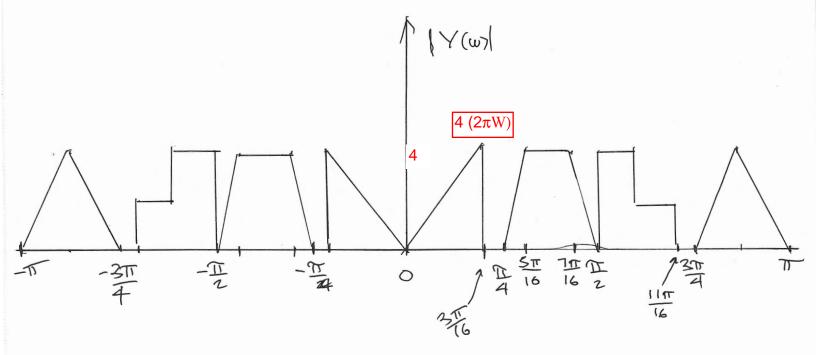
$$y_{3}[n] = x_{0}[n] + x_{1}[n]\cos\left(\frac{\pi}{2}3\right) - \hat{x}_{1}[n]\sin\left(\frac{\pi}{2}3\right) + x_{2}[n]\cos\left(\frac{\pi}{2}3\right) + \hat{x}_{2}[n]\sin\left(\frac{\pi}{2}3\right) + x_{3}[n]\cos(\pi 3)$$

$$(3)$$

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DTFT of each signal is compressed by a factor of F  
Signal Ø is centered at 
$$w=0$$
  
Signal 3 is centered at  $w=T$  (and  $w=-T$ )  
Signal 1's upper side band is in  $\frac{T}{2} < w < \frac{3T}{4}$   
mirror image in  $-\frac{3T}{4} < w < -\frac{T}{2}$   
Signal 2's lower sideband is in  $\frac{T}{4} < w < \frac{T}{2}$   
mirror image in  $-\frac{T}{4} < w < \frac{T}{4}$ 



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