

**NAME:** **29 Oct. 2012**  
**ECE 538 Digital Signal Processing I Exam 2 Fall 2012**

## Cover Sheet

**WRITE YOUR NAME ON THIS COVER SHEET**

Test Duration: 60 minutes.

Open Book but Closed Notes.

Calculators NOT allowed.

All work should be done in the space provided.

**Problem 1.** Consider a CT signal  $x_a(t)$  with bandwidth (maximum frequency)  $W$  in Hz. The sampling rate is chosen to be above the Nyquist rate at  $F_s = \frac{1}{T_s} = 3W$ .  $x_a(t)$  is reconstructed according to the formula below, where  $W_1 < W_2$ . Determine the respective values of  $W_1$  and  $W_2$ , both in terms of  $W$ , so that the formula below yields perfect reconstruction.

$$x_a(t) = \sum_{n=-\infty}^{\infty} x_a(nT_s)h(t-nT_s) \quad \text{where: } h(t) = T_s \frac{1}{2W_1} \frac{\sin(2\pi W_1 t)}{\pi t} \frac{\sin(2\pi W_2 t)}{\pi t} \quad \text{and } F_s = \frac{1}{T_s} = 3W$$

**Problem 2.** In class, we derived that the DTFT of  $y[n] = x[Ln]$  is  $Y(\omega) = \frac{1}{L} \sum_{k=0}^{L-1} X\left(\frac{\omega - k2\pi}{L}\right)$ .

Use this result in conjunction with the time-shift property of the DTFT to derive a similar general expression for the DTFT of  $h_\ell^-[n] = h[Ln - \ell]$ , denoted  $H_\ell^-(\omega)$  in terms of the DTFT of  $h[n]$ , denoted  $H(\omega)$ . Show your work directly below and put a box around your final answer.

**Problem 3.** Each of the four signals in this problem has the same bandwidth (that is, same maximum frequency) and is sampled at the same rate, the Nyquist rate (no aliasing.)

- (a) Consider the continuous-time signal  $x_0(t)$  below. A discrete-time signal is created by sampling  $x_0(t)$  according to  $x_0[n] = x_0(nT_s)$  for  $T_s = \frac{2\pi}{20}$ . Plot the magnitude of the DTFT of  $x_0[n]$ ,  $|X_0(\omega)|$ , over  $-\pi < \omega < \pi$ . Show all work.

$$x_0(t) = T_s \frac{1}{2} \left\{ \frac{\sin(10(t - \frac{\pi}{20}))}{\pi(t - \frac{\pi}{20})} + \frac{\sin(10(t + \frac{\pi}{20}))}{\pi(t + \frac{\pi}{20})} \right\}$$

- (b) Consider the continuous-time signal  $x_1(t)$  below. A discrete-time signal is created by sampling  $x_1(t)$  according to  $x_1[n] = x_1(nT_s)$  for  $T_s = \frac{2\pi}{20}$ . Plot the magnitude of the DTFT of  $x_1[n]$ ,  $|X_1(\omega)|$ , over  $-\pi < \omega < \pi$ . Show all work.

$$x_1(t) = T_s \frac{1}{2\pi} \left\{ \frac{\sin(2t)}{\pi t} \frac{\sin(8t)}{\pi t} \right\}$$

- (c) Consider the continuous-time signal  $x_2(t)$  below. A discrete-time signal is created by sampling  $x_2(t)$  according to  $x_2[n] = x_2(nT_s)$  for  $T_s = \frac{2\pi}{20}$ . Plot the magnitude of the DTFT of  $x_2[n]$ ,  $|X_2(\omega)|$ , over  $-\pi < \omega < \pi$ . Show all work.

$$x_2(t) = T_s \frac{\pi}{5} \left\{ \frac{\sin(5t)}{\pi t} \right\}^2$$

- (d) Consider the continuous-time signal  $x_3(t)$  below. A discrete-time signal is created by sampling  $x_3(t)$  according to  $x_3[n] = x_3(nT_s)$  for  $T_s = \frac{2\pi}{20}$ . Plot the magnitude of the DTFT of  $x_3[n]$ ,  $|X_3(\omega)|$ , over  $-\pi < \omega < \pi$ . Show all work.

$$x_3(t) = T_s \frac{1}{2j} \left\{ \frac{\sin(10(t - \frac{\pi}{10}))}{\pi(t - \frac{\pi}{10})} - \frac{\sin(10(t + \frac{\pi}{10}))}{\pi(t + \frac{\pi}{10})} \right\}$$

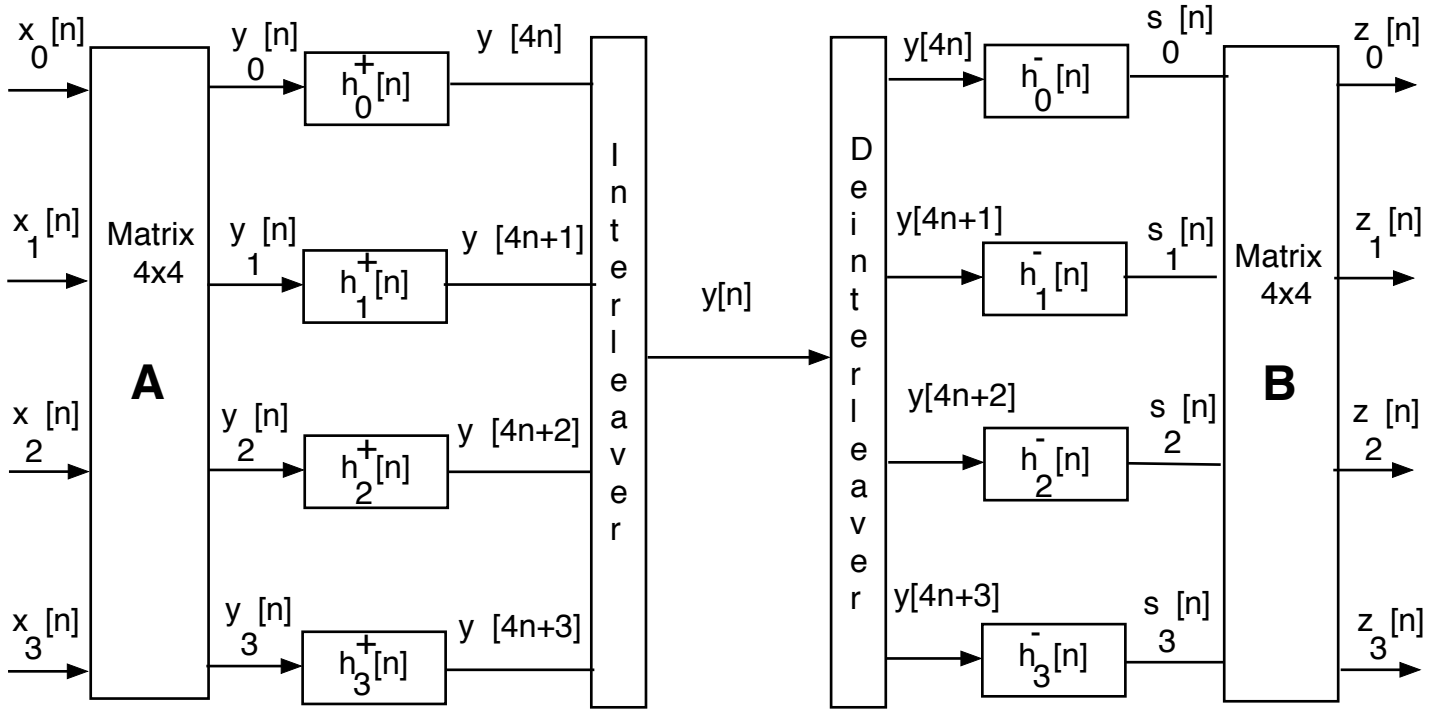


Figure 1.

**Problem 4.** This problem is about digital subbanding of the four DT signals  $x_i[n]$ ,  $i = 1, 2, 3, 4$  from Problem 3. Digital subbanding of these four signals is effected in the efficient way via filter bank in Figure 1. All of the quantities in Figure 1 are defined below.

Relative to Figure 1, the respective impulse responses of the polyphase component filters are defined in terms of the lowpass filter with the ideal impulse response below.

$$h_{LP}[n] = 4 \frac{\sin\left(\frac{\pi}{4}n\right)}{\pi n} \quad (1)$$

The polyphase component filters on the left side of Figure 1 are defined as

$$h_\ell^+[n] = h_{LP}[4n + \ell], \quad \ell = 0, 1, 2, 3. \quad (2)$$

The respective signals at the inputs to these filters are formed from the input signals (from Problem 3) via the matrix transformation below

$$\begin{bmatrix} y_0[n] \\ y_1[n] \\ y_2[n] \\ y_3[n] \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix} \begin{bmatrix} x_0[n] \\ x_1[n] \\ x_2[n] \\ x_3[n] \end{bmatrix} \Rightarrow \mathbf{A} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix} \quad (3)$$

The polyphase component filters on the right side of Figure 1 are defined as

$$h_\ell^-[n] = h_{LP}[4n - \ell], \quad \ell = 0, 1, 2, 3. \quad (4)$$

The final output signals (on the far right side of Figure 1) are formed from linear combinations of the outputs of these filters via the matrix transformation below.

$$\begin{bmatrix} z_0[n] \\ z_1[n] \\ z_2[n] \\ z_3[n] \end{bmatrix} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} s_0[n] \\ s_1[n] \\ s_2[n] \\ s_3[n] \end{bmatrix} \Rightarrow \mathbf{B} = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \quad (5)$$



Problem 4 part (a). Show all work in the space below. For all parts of this problem,

$$h_{LP}[n] = 4 \frac{\sin\left(\frac{\pi}{4}n\right)}{\pi n}.$$

- (a) (i) Determine and write a simplified expression for the DTFT,  $H_2^+(\omega)$ , of  $h_2^+[n] = h_{LP}[4n + 2]$  that holds for  $-\pi < \omega < \pi$ . Simplify as much as possible.
- (ii) Plot the magnitude of  $H_2^+(\omega)$  over  $-\pi < \omega < \pi$ .
- (iii) Plot the phase of  $H_2^+(\omega)$  over  $-\pi < \omega < \pi$ .

- (b) Express the output of the filter  $h_2^+[n] = h_{LP}[4n + 2]$  in terms of sampled and time-shifted versions of the original analog input signals  $x_0(t)$ ,  $x_1(t)$ ,  $x_2(t)$ , and  $x_3(t)$ .

You don't need to write out the expressions for  $x_0(t)$ ,  $x_1(t)$ ,  $x_2(t)$ , and  $x_3(t)$ . Also, just carry  $T_s$  along as a variable, rather than having to write its specific numerical value of  $T_s = \frac{2\pi}{20}$ . You don't have to do a lot of work here; just briefly explain your answer.

- (c) Plot the magnitude of the DTFT,  $Y(\omega)$ , of the interleaved signal  $y[n]$ . Carefully label and graph the plot, clearly demarcating the subbands and showing what's in each subband.

(d) Determine the convolution of  $h_2^+[n] = h_{LP}[4n + 2]$  and  $h_2^-[n] = h_{LP}[4n - 2]$ , where

$$h_{LP}[n] = 4 \frac{\sin\left(\frac{\pi}{4}n\right)}{\pi n}.$$

$$g_\ell[n] = h_2^+[n] * h_2^-[n] = h_{LP}[4n + 2] * h_{LP}[4n - 2] = ?$$

This is most easily determined via frequency domain analysis. You must show and explain your work. Do a stem plot of  $g_\ell[n] = h_2^+[n] * h_2^-[n]$ .

(e) Express the output of the filter  $z_2[n]$  in terms of  $x_0[n]$ ,  $x_1[n]$ ,  $x_2[n]$ , and  $x_3[n]$ .