# EE538 Digital Signal Processing I Session 13 Exam 1 

## Cover Sheet

Test Duration: 50 minutes.<br>Coverage: Sessions 1-10.<br>Open Book but Closed Notes.<br>Calculators not allowed.<br>This test contains three problems.<br>All work should be done in the blue books provided.<br>You must show all work for each problem to receive full credit.<br>Do not return this test sheet, just return the blue books.

Prob. No. Topic(s)

1. DT Correlation
2. LTI Systems (Causal):

Properties, Frequency Response
Interconnection, Pole-Zero Diagram
3. Digital Upsampling

## Points

30
30

40

Problem 1. [30 points]
Consider a very simplistic CDMA system with only two users assigned the following length 4 orthogonal complex-value codes, respectively, which are two rows of a 4-point DFT matrix:

$$
\begin{array}{ll}
\text { User 1's code: } & c_{1}[n]=\{1,-j,-1, j\} \\
\text { User 2's code: } & c_{2}[n]=\{1, j,-1,-j\}
\end{array}
$$

Consider transmitting a block of two QAM symbols for each of the two users,
User 1's two info. symbols: $b_{1}[n]=\left\{b_{1}[0], b_{1}[1]\right\}$
User 2's two info. symbols: $\quad b_{2}[n]=\left\{b_{2}[0], b_{2}[1],\right\}$
where $b_{k}[n]$ can take on one of four different complex values (for each value of $k$ and $n$ ):

1. $1+\mathrm{j}$ representing the bit pair $\{1,1\}$;
2. $-1+\mathrm{j}$ representing the bit pair $\{0,1\}$;
3. $-1-j$ representing the bit pair $\{0,0\}$;
4. 1-j representing the bit pair $\{1,0\}$;

The transmitted code-division multiplexed block may be mathematically expressed as

$$
x[n]=\sum_{k=1}^{2} \sum_{m=0}^{1} b_{k}[m] c_{k}[n-4 m], \quad n=0,1, \ldots, 7
$$

Given that the received block has the following numerical values

$$
x[n]=\{\underbrace{2 j}_{\uparrow},-2 j,-2 j, 2 j,-2 j, 2 j, 2 j,-2 j\}
$$

where the first entry above is the value of $x[0]$, determine the numerical values of $b_{1}[n], n=$ 0,1 and $b_{2}[n], n=0,1$. Your answer should consist of 4 numerical values all together. Show all work in arriving at your answer. Note that code 1 is orthogonal to code 2 and that the two users are synchronized.

Problem 2. [30 points]
Consider the causal, second-order difference equation below where it is noted that $.9025=$ $.95^{2}$.

$$
y[n]=-0.95 y[n-1]-0.9025 y[n-2]+x[n]+x[n-1]+x[n-2]
$$

Consider implementing this second-order difference equation as two first-order systems in parallel (one pole each) as shown in the diagram.

(a) Determine and write the first-order difference equation for each of the two first-order systems in parallel. The upper first-order system has impulse response $h_{1}[n]$ and is described by the difference equation

$$
y_{1}[n]=a_{1}^{(1)} y_{1}[n-1]+b_{0}^{(1)} x[n]+b_{1}^{(1)} x[n-1]
$$

The lower first-order system has impulse response $h_{2}[n]$ and is described by the difference equation

$$
y_{2}[n]=a_{1}^{(2)} y_{2}[n-1]+b_{0}^{(2)} x[n]+b_{1}^{(2)} x[n-1]
$$

Determine the numerical values of $a_{1}^{(i)}, b_{0}^{(i)}$, and $b_{1}^{(i)}, i=1,2-$ six values total.
(b) For EACH of the two first-order systems, $i=1,2$, do the following:
(i) Plot the pole-zero diagram.
(ii) State and plot the region of convergence for $H_{i}(z)$.
(iii) Determine the DTFT of $h_{i}[n]$ and plot the magnitude $\left|H_{i}(\omega)\right|$ over the interval $-\pi<\omega<\pi$ showing as much detail as possible. In particular, explicitly point out if there are any values of $\omega$ for which $\left|H_{i}(\omega)\right|$ is exactly zero.

Problem 3. [40 points]

$\left.\xrightarrow{\mathrm{Xa}_{\mathrm{a}}^{(\mathrm{t})}} \begin{array}{c}\text { Ideal A/D} \\ \mathrm{F}_{\mathrm{s}}=4 \mathrm{~W}\end{array}\right) \xrightarrow{\mathrm{x}[\mathrm{n}]} \xrightarrow{\mathrm{w}[\mathrm{n}]} \begin{aligned} & \text { Lowpass Filter } \mathrm{h}_{\mathrm{LP}}[\mathrm{n}] \\ & \omega_{\mathrm{p}}=\frac{\pi}{4} \quad \omega_{\mathrm{S}}=\frac{3 \pi}{4} \\ & \text { gain }=2\end{aligned} \quad y[\mathrm{n}]$
Figure 1.
The analog signal $x_{a}(t)$ with CTFT $X_{a}(F)$ shown above is input to the system above, where $x[n]=x_{a}\left(n / F_{s}\right)$ with $F_{s}=4 W$, and

$$
h_{L P}[n]=\frac{\sin \left(\frac{\pi}{2} n\right)}{\frac{\pi}{2} n} \frac{\cos \left(\frac{\pi}{4} n\right)}{1-\frac{n^{2}}{4}}, \quad-\infty<n<\infty
$$

such that $H_{L P}(\omega)=2$ for $|\omega| \leq \frac{\pi}{4}, H_{L P}(\omega)=0$ for $\frac{3 \pi}{4} \leq|\omega| \leq \pi$, and $H_{L P}(\omega)$ has a cosine roll-off from 1 at $\omega_{p}=\frac{\pi}{4}$ to 0 at $\omega_{s}=\frac{3 \pi}{4}$. Finally, the zero inserts may be mathematically described as

$$
w[n]= \begin{cases}x\left(\frac{n}{2}\right), & n \text { even } \\ 0, & n \text { odd }\end{cases}
$$

(a) Plot the magnitude of the DTFT of the output $y[n], Y(\omega)$, over $-\pi<\omega<\pi$.
(b) Given that

$$
x[n]=\frac{\sin \left(\frac{\pi}{2} n\right)}{\pi n} \quad-\infty<n<\infty
$$

provide an analytical expression for $y[n]$ for $-\infty<n<\infty$ (similar to the expression for either $x[n]$ or $h_{L P}[n]$ above, for example.)

## THIS PROBLEM IS CONTINUED ON THE NEXT PAGE.

(c) The up-sampling by a factor of 2 in Figure 1 can be efficiently done via the block diagram in Figure 2 below.
(i) Provide an analytical expression for $h_{0}[n]=h_{L P}[2 n]$ for $-\infty<n<\infty$. Simplify as much as possible.
(ii) Plot the magnitude of the DTFT of $h_{0}[n],\left|H_{0}(\omega)\right|$, over $-\pi<\omega<\pi$.
(iii) Provide an analytical expression for the output $y_{0}[n]$ for $-\infty<n<\infty$.
(iv) Plot the magnitude of the DTFT of $h_{1}[n],\left|H_{1}(\omega)\right|$, over $-\pi<\omega<\pi$.
(v) Plot the phase of the DTFT of $h_{1}[n], \angle H_{1}(\omega)$, over $-\pi<\omega<\pi$.
(vi) Provide an analytical expression for the output $y_{1}[n]$ for $-\infty<n<\infty$.


Figure 2.

