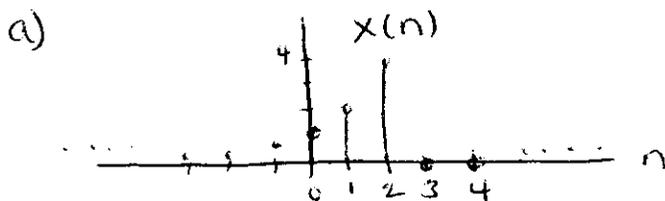


- You have 75 minutes to work the following five problems.
- Be sure to show all your work to obtain full credit.
- The exam is closed book and closed notes. However, you bring with you a sheet of formulas handwritten on both sides of one 8.5x11 in. sheet of paper
- Calculators are permitted.

1. (20 pts.) For the signal $x[n] = 2^n u[-n + 2]$, do the following:

- Sketch $x[n]$.
- State whether or not the signal has finite or infinite duration.
- State whether or not the signal is right-sided, left-sided, or two-sided.
- State whether or not the signal is causal or anticausal.
- Calculate the DTFT $X(e^{j\omega})$ of the signal $x[n]$.



- b) infinite duration
 c) Left-sided
 d) Neither
 e)

d)

$$\begin{aligned}
 X(\omega) &= \sum_{n=-\infty}^{\infty} 2^n e^{-j\omega n} = \sum_{n=-\infty}^0 2^n e^{-j\omega n} + 2^1 e^{-j\omega 1} + 2^2 e^{-j\omega 2} \\
 &= \sum_{m=0}^{\infty} 2^{-m} e^{+j\omega m} + 2e^{-j\omega} + 4e^{-j\omega 2} \\
 &= \sum_{m=0}^{\infty} \left(\frac{1}{2} e^{j\omega}\right)^m + 2e^{-j\omega} + 4e^{-j\omega 2} \\
 &= \frac{1}{1 - \frac{1}{2} e^{j\omega}} + 2e^{-j\omega} + 4e^{-j\omega 2} = \frac{4e^{-j\omega 2}}{1 - \frac{1}{2} e^{j\omega}}
 \end{aligned}$$



2. (20 pts.) Consider the system defined by the equation $y[n] = \frac{1}{3}(x[n] - x[n-3])$.
- State whether or not the system is linear, time-invariant, causal, and BIBO stable. (Note: no proofs are needed.)
 - Find the impulse response $h[n]$ for the system.
 - Find the frequency response $H(e^{j\omega})$ for the system, and determine simple expressions for the magnitude and phase of the frequency response.

a) Linear

Time Invariant

Causal

BIBO stable

$$b) h[n] = \frac{1}{3} (\delta[n] - \delta[n-3])$$

$$c) H(\omega) = \frac{1}{3} (1 - e^{-j\omega 3})$$

$$= \frac{1}{3} e^{-j\omega 3/2} (e^{j\omega 3/2} - e^{-j\omega 3/2}) \cdot \frac{2j}{2j}$$

$$= \frac{2j}{3} e^{-j\omega 3/2} \sin(3\omega/2)$$

$$= \frac{2}{3} e^{j(\frac{\pi}{2} - \frac{\omega 3}{2})} \sin(3\omega/2)$$

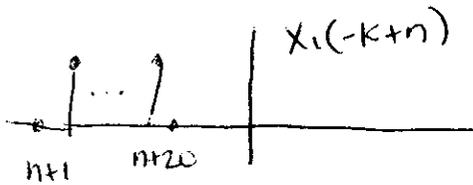
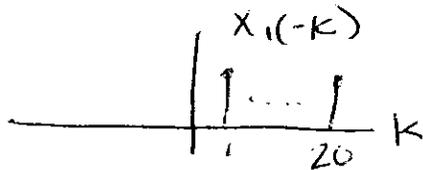
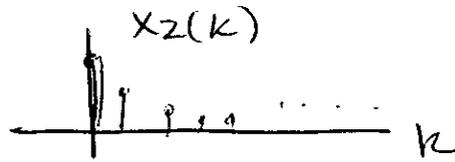
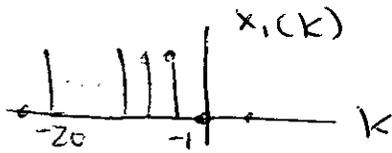
$$|H(\omega)| = \frac{2}{3} |\sin(3\omega/2)|$$

$$\angle H(\omega) = \begin{cases} \frac{\pi}{2} - \frac{\omega 3}{2} + \pi & \sin(3\omega/2) < 0 \\ \frac{\pi}{2} - \frac{3\omega}{2} & \sin(3\omega/2) \geq 0 \end{cases}$$



3. (20 pts.) Find the convolution of the following two signals:

$$x_1[n] = u[n+20] - u[n], \quad x_2[n] = 2^{-n}u[n].$$

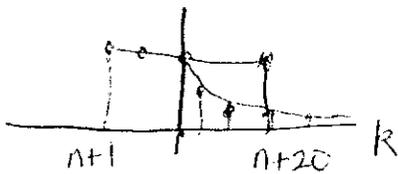


Case 1 No overlap



$$\begin{aligned} n+20 &< 0 \\ n &< -20 \quad y(n) = 0 \end{aligned}$$

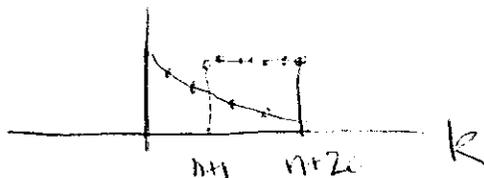
Case 2 $x_1(-k+n)$ entering $x_2(k)$



$$\begin{aligned} n+20 &\geq 0 \quad \text{and} \quad n+1 < 0 \\ n &\geq -20 \quad \quad \quad n < -1 \\ -20 &\leq n < -1 \end{aligned}$$

$$y(n) = \sum_{k=0}^{n+20} 2^{-k} = \frac{1 - (\frac{1}{2})^{n+21}}{1 - \frac{1}{2}} = 2 \left(1 - (\frac{1}{2})^{n+21} \right)$$

Case 3 $x_1(-k+n)$ and $x_2(k)$ fully overlap



$$\begin{aligned} n+1 &\geq 0 \quad \text{and} \quad n+20 \geq 0 \\ n &\geq -1 \end{aligned}$$

3. (continued)

$$y(n) = \sum_{n+1}^{n+20} 2^{-k} \quad \text{let } m = k - (n+1)$$

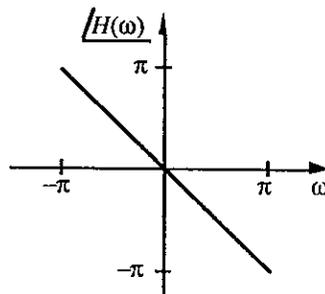
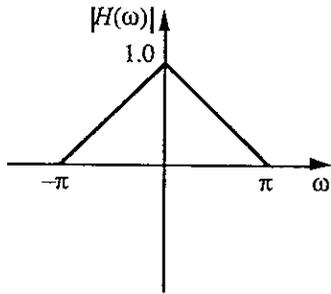
$$= \sum_{m=0}^{19} 2^{-(m+n+1)} = 2^{-n} 2^{-1} \sum_{m=0}^{19} 2^{-m}$$

$$= 2^{-n} 2^{-1} \left(\frac{1 - 2^{-20}}{1 - \frac{1}{2}} \right) = \frac{2^{-n} 2^{-1} (1 - (\frac{1}{2})^{20})}{2^{-1}}$$

$$= 2^{-n} (1 - (\frac{1}{2})^{20})$$

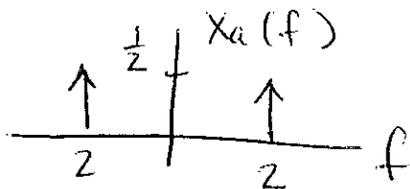
$$y(n) = \begin{cases} 0 & n < -20 \\ 2(1 - (\frac{1}{2})^{n+21}) & -20 \leq n < -1 \\ 2^{-n} (1 - (\frac{1}{2})^{20}) & n \geq -1 \end{cases}$$

4. (20 pts.) The CT signal $x_a(t) = \cos[2\pi(2000)t]$ is sampled with an ideal sampler at a rate $f_s = 8$ kHz. to yield the DT signal $x_d[n]$. The signal $x_d[n]$ is filtered with a digital filter with frequency response $H(e^{j\omega})$ given below to yield the DT signal $y_d[n]$. This signal is then converted to the CT signal $y_a(t)$ using an ideal D/A convertor with a cutoff filter at 4 kHz. Find $y_a(t)$.



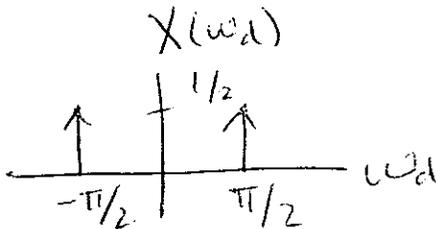
$$x_a(t) = \cos(2\pi(2000)t)$$

$$X_a(f) = \frac{1}{2} (\delta(f-2000) + \delta(f+2000))$$



A/D $X(\omega) = X_a(f) \Big|_{\omega_d = \frac{2\pi f_a}{f_s} = 2\text{kHz}}$ $= \frac{1}{2} (\delta(\omega - \frac{\pi}{2}) + \delta(\omega + \frac{\pi}{2}))$

$f_s \leftarrow 8\text{kHz}$

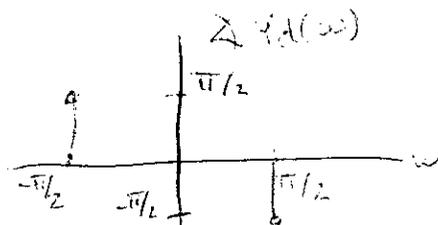
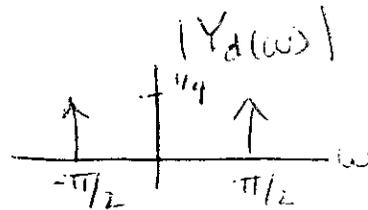


$H(\omega)$

$$|H(\omega)| \Big|_{\omega = \pm \frac{\pi}{2}} = \frac{1}{2}$$

$$\angle H(\omega) \Big|_{\omega = \frac{\pi}{2}} = -\frac{\pi}{2}$$

$$\angle H(\omega) \Big|_{\omega = -\frac{\pi}{2}} = +\frac{\pi}{2}$$



4. (continued)

$$\begin{aligned}
 Y_d(\omega) &= \frac{1}{4} \left[\delta\left(\omega - \frac{\pi}{2}\right) e^{-j\pi/2} + \delta\left(\omega + \frac{\pi}{2}\right) e^{j\pi/2} \right] \\
 &= \frac{1}{4} e^{-j\pi/2} \left(\delta\left(\omega - \frac{\pi}{2}\right) - \delta\left(\omega + \frac{\pi}{2}\right) \right) \\
 &= \frac{-j}{4} \left(\delta\left(\omega - \frac{\pi}{2}\right) - \delta\left(\omega + \frac{\pi}{2}\right) \right) \\
 &= \frac{1}{2} \cdot \frac{1}{2j} \left(\delta\left(\omega - \frac{\pi}{2}\right) - \delta\left(\omega + \frac{\pi}{2}\right) \right)
 \end{aligned}$$

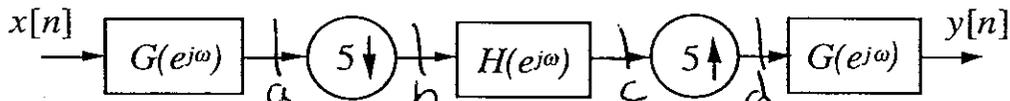
• D/A $f = \frac{\omega f_s}{2\pi}$

$$Y_a(f) = \frac{1}{2} \cdot \frac{1}{2j} \left(\delta(\omega - 2000) - \delta(\omega + 2000) \right)$$

$f_c = 4 \text{ kHz} \Rightarrow$ Nothing above $\pm 2 \text{ kHz}$, so
filter has no effect

• $y(t) = \frac{1}{2} \sin(2\pi(2000)t)$ //

5. (20 pts.) Consider the system shown below.



where the filter frequency responses of the two filters are given by

$$G(e^{j\omega}) = \begin{cases} 1, & |\omega| < \pi/5 \\ 0, & \text{else} \end{cases} \quad H(e^{j\omega}) = e^{-j\omega/2} \cos(\omega/2)$$

Find a simple expression for the frequency response of the overall system.

$$Y_a(\omega) = \begin{cases} X(\omega) & |\omega| < \pi/5 \\ 0 & \text{else} \end{cases}$$

$$Y_b(\omega) = \frac{1}{5} \sum_{k=0}^4 Y_a\left(\frac{\omega + 2\pi k}{5}\right)$$

$$\begin{aligned} Y_c(\omega) &= \left(\frac{1}{5} \sum_{k=0}^4 Y_a\left(\frac{\omega + 2\pi k}{5}\right) \right) H(\omega) \\ &= \frac{1}{5} \sum_{k=0}^4 Y_a\left(\frac{\omega + 2\pi k}{5}\right) e^{-j\omega/2} \cos(\omega/2) \end{aligned}$$

$$\begin{aligned} Y_d(\omega) &= Y_c(5\omega) \\ &= \frac{1}{5} \sum_{k=0}^4 Y_a\left(\frac{5\omega + 2\pi k}{5}\right) e^{-j5\omega/2} \cos(5\omega/2) \end{aligned}$$

$$\begin{aligned} Y(\omega) &= \begin{cases} Y_d(\omega) & |\omega| < \pi/5 \\ 0 & \text{else} \end{cases} \\ &= \begin{cases} \frac{1}{5} e^{-j5\omega/2} \cos\left(\frac{5\omega}{2}\right) X(\omega) & |\omega| < \pi/5 \\ 0 & \text{else} \end{cases} \end{aligned}$$

5. (continued)

$$H(\omega) = \frac{Y(\omega)}{X(\omega)} = \begin{cases} \frac{1}{5} e^{-j\frac{5\omega}{2}} \cos\left(\frac{5\omega}{2}\right) & |\omega| < \frac{\pi}{5} \\ 0 & \text{else} \end{cases}$$

$$H(\omega) = \frac{1}{5} e^{-j5\omega/2} \cos(5\omega/2) \underbrace{\text{rect}\left(\frac{5\omega}{2\pi}\right)}_{\substack{\downarrow \\ \text{rect}\left(\frac{\omega-0}{2\pi/5}\right)}} //$$