ECE 438 Exam No. 2 Spring 2020

- You have 24 hours to work the following four problems.
- Be sure to show all your work to obtain full credit.
- The exam is open book and open notes.
- Please do **NOT** discuss the problems with anyone else.
- 1. (25 pts) Consider the causal DT system described by the difference equation

$$y[n] = x[n] + \frac{1}{2}y[n-1]$$

- a. (19) Use Z-transform techniques, including a partial fraction expansion, to find the response y[n] of this system to the input $x[n] = u[-n] + \left(\frac{1}{2}\right)^n u[n]$, where u[n] is the unit step function.
- b. (6) Carefully sketch by hand the input x[n] and the output y[n], as functions of the independent variable n. Be sure to dimension both axes.

2. (25 pts) Fast Fourier Transform Algorithm

- a. (3) Calculate the *approximate* number of complex operations (COs) required to compute an 18-point DFT by directly evaluating the 18-point DFT sum. Here a complex operation is taken to mean 1 complex addition and 1 complex multiplication.
- b. (12) Derive a complete set of equations to show how an 18-point Discrete Fourier Transform (DFT) can be calculated in terms of two 3-point DFTs and one 2-point DFT via decimation-in-time.
- c. (5) Based on your answer to part b) above, list the complete ordering of the 18-point input to your 18-point FFT algorithm.
- d. (5) Based on your answer to part b) above, calculate the *approximate* number of complex operations (COs) required to compute an 18-point DFT using your FFT algorithm.

3. (25 pts) Spectral analysis via the DFT

Consider the 10-point signal
$$x[n] = \cos\left(\frac{7\pi}{10}n\right)$$
, $n = 0,...,9$.

- a. (15) Determine an exact expression for the 10-point discrete Fourier transform (DFT) $X^{(10)}[k], k = 0,...,9$ in terms of the function $\operatorname{psinc}_N(\omega) = \frac{\sin(\omega N/2)}{\sin(\omega/2)}$.
- b. (7) Based on your answer to part a) above, carefully sketch *by hand* the 10-point DFT $X^{(10)}[k]$, k = 0,...,9. Be sure to dimension both axes. You may ignore the contribution of phase terms in your sketch.
- c. (3) Discuss whether or not *picket fence effect* or *leakage* are present for this example, and explain how they manifest themselves in the 10-point DFT that you sketched in your answer to part b) above.

4. (25 pts) Consider a random variable X with density function

$$f_X(x) = \begin{cases} |x|, & 0 \le |x| \le 1 \\ 0, & \text{else} \end{cases}.$$

- a. (3) Carefully sketch by hand the density function $f_X(x)$. Be sure to dimension both axes.
- b. (4) Find the mean and variance of X.
- c. (4) Suppose we generate a new random variable Y = Q(X) by quantizing X according to the following 3-level *uniform* quantizer:

$$Q(x) = \begin{cases} -\frac{2}{3}, & -1 \le x < -\frac{1}{3} \\ 0, & -\frac{1}{3} \le x < \frac{1}{3} \end{cases}.$$

$$\frac{2}{3}, & \frac{1}{3} \le x \le 1$$

Determine the *approximate* mean-squared quantization error $\varepsilon = E\left\{\left|Y - X\right|^2\right\}$ using the expression $\varepsilon_{\text{approx}} = \frac{\Delta^2}{12}$

- d. (12) Now calculate the *exact* mean-squared quantization error $\varepsilon_{\text{exact}} = E\{|Y X|^2\}$ for the 3-level *uniform* quantizer given above. Note that you can simplify your calculations by taking advantage of the symmetry of both $f_X(x)$ and Q(x).
- e. (2) Compare your answers to parts c) and d) above.

Problem 1

(25 pts) Consider the causal DT system described by the difference equation

$$y[n] = x[n] + \frac{1}{2}y[n-1]$$

a. (19) Use Z-transform techniques, including a partial fraction expansion, to find the response y[n] of this system to the input $x[n] = u[-n] + (\frac{1}{2})^n u[n]$, where u[n] is the unit step function.

(6) Carefully sketch by hand the input x[n] and the output y[n], as functions of the independent variable n. Be sure to dimension both axes.

a.

$$y[n] = x[n] + \frac{1}{2}y[n-1]$$
 Using $x[n-n_0] \stackrel{\text{ZT}}{\longleftrightarrow} z^{-n_0}X(z)$

$$Y(z) = X(z) + \frac{1}{2}Y(z)z^{-1}$$
 H(z): 1 pt: Z transfrom 1 pt: correct H(z) 1 pt: ROC of H(z)

1 pt: correct H(z) expression

1 pt: ROC of H(z)

 $H(z) = \frac{Y(z)}{X(z)} = \frac{1}{1 - \frac{1}{2}z^{-1}}, \quad |z| > \frac{1}{2}$ since it is a causal system

Apply Z Transform to x[n], calculate Y(z), apply inverse Z Transform to Y(z)

Method 1:

$$x[n] = u[-n] + \left(\frac{1}{2}\right)^n u[n]$$

Using the following ZT pairs:

$$\begin{split} u[-n] &\longleftrightarrow \frac{1}{1-z}, |z| < 1 \\ a^n u[n] &\longleftrightarrow \frac{1}{1-az^{-1}}, |z| > |a| \end{split}$$

X(z): 3 pt: Z transform 1 pt: correct X(z) expression

1 pt: ROC of X(z)

$$x[n] \stackrel{\text{ZT}}{\longleftrightarrow} X(z) = \frac{1}{1-z} + \frac{1}{1-\frac{1}{2}z^{-1}}, \quad \frac{1}{2} < |z| < 1$$

$$\begin{split} \Rightarrow Y(z) &= H(z)X(z) \\ &= \frac{1}{(1 - \frac{1}{2}z^{-1})(1 - z)} + \frac{1}{(1 - \frac{1}{2}z^{-1})^2} \\ &= \frac{1}{(1 - \frac{1}{2}z^{-1})(1 - \frac{1}{z^{-1}})} + \frac{1}{(1 - \frac{1}{2}z^{-1})^2} \\ &= \frac{-z^{-1}}{(1 - \frac{1}{2}z^{-1})(1 - z^{-1})} + \frac{1}{(1 - \frac{1}{2}z^{-1})^2} \\ &= \frac{A}{1 - \frac{1}{2}z^{-1}} + \frac{B}{1 - z^{-1}} + \frac{(2z)(\frac{1}{2}z^{-1})}{(1 - \frac{1}{2}z^{-1})^2}, \quad \frac{1}{2} < |z| < 1 \end{split}$$

Solve for A and B: $A - Az^{-1} + B - \frac{1}{2}Bz^{-1} = -z^{-1} \Rightarrow \begin{cases} A + B = 0 \\ -A - \frac{1}{2}B = -1 \end{cases}$ and get A = 2, B = -2.

$$\begin{split} Y(z) &= \frac{2}{1 - \frac{1}{2}z^{-1}} - \frac{2}{1 - z^{-1}} + \frac{(2z)(\frac{1}{2}z^{-1})}{(1 - \frac{1}{2}z^{-1})^2} \\ &= \frac{2}{1 - \frac{1}{2}z^{-1}} - \frac{2}{1 - z^{-1}} + (2z)\frac{\frac{1}{2}z^{-1}}{(1 - \frac{1}{2}z^{-1})^2}, \qquad \frac{1}{2} < |z| < 1 \end{split}$$

Using the following ZT pairs:

$$a^{n}u[n] \longleftrightarrow \frac{1}{1-az^{-1}}, |z| > |a|$$

$$-a^{n}u[-n-1] \longleftrightarrow \frac{1}{1-az^{-1}}, |z| < |a|$$

$$na^{n}u[n] \longleftrightarrow \frac{az^{-1}}{(1-az^{-1})^{2}}, |z| > |a|$$

$$x[n-n_{0}] \longleftrightarrow z^{-n_{0}}X(z)$$

$$y[n]: 1 \text{ pt: } Y(z) = H(z)X(z)$$

$$2 \text{ pt: expansion of } Y(z)$$

$$2 \text{ pt: solving for A and B}$$

$$1 \text{ pt: ROC of } Y(z)$$

$$4 \text{ pt: Z transform}$$

$$1 \text{ pt: correct } y[n] \text{ expression}$$

$$\begin{split} Y(z) & \stackrel{\text{ZT}}{\longleftrightarrow} y[n] = 2 \left(\frac{1}{2}\right)^n u[n] + 2u[-n-1] + 2(n+1) \left(\frac{1}{2}\right)^{n+1} u[n+1] \\ &= 2 \left(\frac{1}{2}\right)^n u[n] + 2u[-n-1] + (n+1) \left(\frac{1}{2}\right)^n u[n+1], \quad \ \frac{1}{2} < |z| < 1 \end{split}$$

Method 2:

$$x[n] = u[-n] + \left(\frac{1}{2}\right)^n u[n]$$
$$= \delta[n] + u[-n-1] + \left(\frac{1}{2}\right)^n u[n]$$

Using the following ZT pairs:

$$\delta[n] \longleftrightarrow 1, \text{all z}$$

$$a^n u[n] \longleftrightarrow \frac{1}{1 - az^{-1}}, |z| > |a|$$

$$-a^n u[-n - 1] \longleftrightarrow \frac{1}{1 - az^{-1}}, |z| < |a|$$

$$x[n] \stackrel{\text{ZT}}{\longleftrightarrow} X(z) = 1 - \frac{1}{1 - z^{-1}} + \frac{1}{1 - \frac{1}{2}z^{-1}}, \quad \frac{1}{2} < |z| < 1$$

$$\begin{split} &\Rightarrow Y(z) = H(z)X(z) \\ &= \frac{1}{1 - \frac{1}{2}z^{-1}} - \frac{1}{(1 - \frac{1}{2}z^{-1})(1 - z^{-1})} + \frac{1}{(1 - \frac{1}{2}z^{-1})^2} \\ &= \frac{2}{1 - \frac{1}{2}z^{-1}} - \left(\frac{A}{1 - \frac{1}{2}z^{-1}} + \frac{B}{1 - z^{-1}}\right) + \left(\frac{1}{(1 - \frac{1}{2}z^{-1})^2} - \frac{1}{1 - \frac{1}{2}z^{-1}}\right) \\ &= \frac{2}{1 - \frac{1}{2}z^{-1}} - \left(\frac{A}{1 - \frac{1}{2}z^{-1}} + \frac{B}{1 - z^{-1}}\right) \frac{\frac{1}{2}z^{-1}}{(1 - \frac{1}{2}z^{-1})^2}, \quad \frac{1}{2} < |z| < 1 \end{split}$$

Solve for A and B:
$$A - Az^{-1} + B - \frac{1}{2}Bz^{-1} = 1 \Rightarrow \begin{cases} A + B = 1 \\ -A - \frac{1}{2}B = 0 \end{cases}$$
 and get $A = -1, B = 2.$

$$\begin{split} Y(z) &= \frac{2}{1 - \frac{1}{2}z^{-1}} - \left(\frac{-1}{1 - \frac{1}{2}z^{-1}} + \frac{2}{1 - z^{-1}}\right) + \frac{\frac{1}{2}z^{-1}}{(1 - \frac{1}{2}z^{-1})^2} \\ &= \frac{3}{1 - \frac{1}{2}z^{-1}} - \frac{2}{1 - z^{-1}} + \frac{\frac{1}{2}z^{-1}}{(1 - \frac{1}{2}z^{-1})^2}, \quad \frac{1}{2} < |z| < 1 \end{split}$$

Using the following ZT pairs:

$$\begin{split} a^n u[n] &\longleftrightarrow \frac{1}{1-az^{-1}}, |z| > |a| \\ -a^n u[-n-1] &\longleftrightarrow \frac{1}{1-az^{-1}}, |z| < |a| \\ na^n u[n] &\longleftrightarrow \frac{az^{-1}}{(1-az^{-1})^2}, |z| > |a| \end{split}$$

$$\begin{split} Y(z) & \stackrel{\text{ZT}}{\longleftrightarrow} y[n] = 3 \left(\frac{1}{2}\right)^n u[n] + 2u[-n-1] + n \left(\frac{1}{2}\right)^n u[n] \\ &= (3+n) \left(\frac{1}{2}\right)^n u[n] + 2u[-n-1] \end{split}$$

b. Sketch:

$$x[n] = u[-n] + \left(\frac{1}{2}\right)^n u[n]$$

$$y[n] = (3+n)\left(\frac{1}{2}\right)^n u[n] + 2u[-n-1]$$

x[n]:

$$n < 0$$
: $x[n] = 1$

$$x[0] = 1 + 1 = 2$$

$$x[1] = \frac{1}{2}$$

$$x[2] = (\frac{1}{2})^2 = \frac{1}{4}$$

$$x[3] = (\frac{1}{2})^3 = \frac{1}{8}$$

$$x[0] = 1 + 1 = 2$$
 $x[1] = \frac{1}{2}$ $x[2] = (\frac{1}{2})^2 = \frac{1}{4}$ $x[3] = (\frac{1}{2})^3 = \frac{1}{8}$ $x[4] = (\frac{1}{2})^4 = \frac{1}{16}$

y[n]:

$$n < 0$$
: $y[n] = 2$

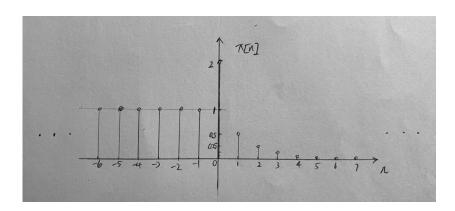
$$y[0] = 3 + 0 = 3$$

$$y[1] = (3+1)\frac{1}{2} = 2$$

$$y[0] = 3 + 0 = 3$$
 $y[1] = (3+1)\frac{1}{2} = 2$ $y[2] = (3+2)(\frac{1}{2})^2 = \frac{5}{4}$

$$y[3] = (3+3)(\frac{1}{2})^3 = \frac{3}{4}$$

$$y[3] = (3+3)(\frac{1}{2})^3 = \frac{3}{4}$$
 $y[4] = (3+4)(\frac{1}{2})^4 = \frac{7}{16}$



x[n]: 2 pt: correct values 1 pt: dimension both axes

y[n]

y[n]: 2 pt: correct values 1 pt: dimension both axes

Problem 2

(25 points) Fast Fourier Transform Algorithm

- (a) (3 points) Calculate the *approximate* number of complex operations (COs) required to compute an 18-point DFT by directly evaluating the 18-point DFT sum. Here a complex operation is taken to mean 1 complex addition and 1 complex multiplication.
- (b) **(12 points)** Derive a complete set of equations to show how an 18-point Discrete Fourier Transform (DFT) can be calculated in terms of three stages: two stages of 3-point DFTs and one stage of 2-point DFT via decimation-in-time.¹
- (c) **(5 points)** Based on your answer to part b) above, list the complete ordering of the 18-point input to your 18-point FFT algorithm.
- (d) **(5 points)** Based on your answer to part b) above, calculate the *approximate* number of complex operations (COs) required to compute an 18-point DFT using your FFT algorithm.
- (a) The DFT formula is

$$X[k] = \sum_{n=0}^{N-1} x[n]e^{-j\frac{2\pi kn}{N}}$$

To calculate each X_k , N multiplications and N-1 additions are required, so there are approximately N COs when N is large.

Therefore, there are approximately $N \times N = N^2 = 324$ COs.

Correct number of COs: 3 points

(b) Again, the DFT formula is

$$X[k] = \sum_{n=0}^{N-1} x[n]e^{-j\frac{2\pi kn}{N}}$$
$$= \sum_{n=0}^{17} x[n]e^{-j\frac{2\pi kn}{18}}$$

¹This part has been restated by Professor Allebach in Piazza.

(a) 4 points Stage III

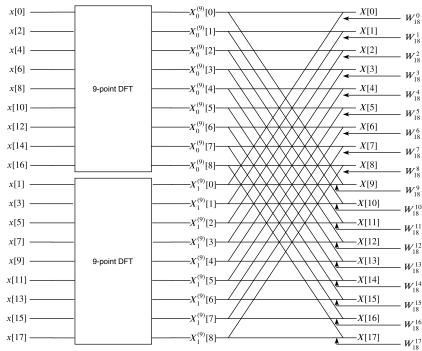
Let n = 2l + p, where l = 0, 1, ..., 8 and p = 0, 1,

$$\begin{split} X^{(18)}[k] &= \sum_{n=0}^{17} x[n] e^{-j\frac{2\pi kn}{18}} \\ &= \sum_{l=0}^{8} x[2l] e^{-j\frac{2\pi k2l}{18}} + \sum_{l=0}^{8} x[2l+1] e^{-j\frac{2\pi k(2l+1)}{18}} \\ &= \sum_{l=0}^{8} x[2l] e^{-j\frac{2\pi kl}{9}} + e^{-j\frac{2\pi k}{18}} \sum_{l=0}^{8} x[2l+1] e^{-j\frac{2\pi kl}{9}} \\ &= \sum_{l=0}^{8} x[2l] e^{-j\frac{2\pi kl}{9}} + W_{18}^{k} \sum_{l=0}^{8} x[2l+1] e^{-j\frac{2\pi kl}{9}} \end{split}$$

Let $x_p[l] = x[2l+p],$

$$X^{(18)}[k] = \sum_{l=0}^{8} x[2l]e^{-j\frac{2\pi kl}{9}} + W_{18}^{k} \sum_{l=0}^{8} x[2l+1]e^{-j\frac{2\pi kl}{9}}$$
$$= \sum_{l=0}^{8} x_{0}[l]e^{-j\frac{2\pi kl}{9}} + W_{18}^{k} \sum_{l=0}^{8} x_{1}[l]e^{-j\frac{2\pi kl}{9}}$$
$$= X_{0}^{(9)}[k] + W_{18}^{k} X_{1}^{(9)}[k],$$

where $W_N^k = e^{-j\frac{2\pi k}{N}}$.



The flow diagram is not required.

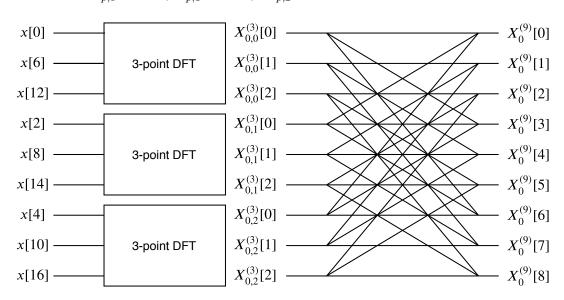
(b) 4 points Stage II

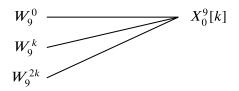
Let l = 3m + q, where m = 0, 1, 2 and q = 0, 1, 2.

$$\begin{split} X_p^{(9)}[k] &= \sum_{l=0}^9 x_p[l] e^{-j\frac{2\pi kl}{9}} \\ &= \sum_{m=0}^2 x_p[3m] e^{-j\frac{2\pi k3m}{9}} + \sum_{m=0}^2 x_p[3m+1] e^{-j\frac{2\pi k(3m+1)}{9}} + \sum_{m=0}^2 x_p[3m+2] e^{-j\frac{2\pi k(3m+2)}{9}} \\ &= \sum_{m=0}^2 x_p[3m] e^{-j\frac{2\pi km}{3}} + e^{-j\frac{2\pi k}{9}} \sum_{m=0}^2 x_p[3m+1] e^{-j\frac{2\pi km}{3}} + e^{-j\frac{2\pi 2k}{9}} \sum_{m=0}^2 x_p[3m+2] e^{-j\frac{2\pi km}{3}} \\ &= \sum_{m=0}^2 x_p[3m] e^{-j\frac{2\pi km}{3}} + W_9^k \sum_{m=0}^2 x_p[3m+1] e^{-j\frac{2\pi km}{3}} + W_9^{2k} \sum_{m=0}^2 x_p[3m+2] e^{-j\frac{2\pi km}{3}} \end{split}$$

Let $x_{p,q}[m] = x_p[3m + q],$

$$X_{p}^{(9)}[k] = \sum_{m=0}^{2} x_{p,0}[m]e^{-j\frac{2\pi km}{3}} + W_{9}^{k} \sum_{m=0}^{2} x_{p,1}[m]e^{-j\frac{2\pi km}{3}} + W_{9}^{2k} \sum_{m=0}^{2} x_{p,2}[m]e^{-j\frac{2\pi km}{3}}$$
$$= X_{p,0}^{(3)}[k] + W_{9}^{k} X_{p,1}^{(3)}[k] + W_{9}^{2k} X_{p,2}^{(3)}[k]$$





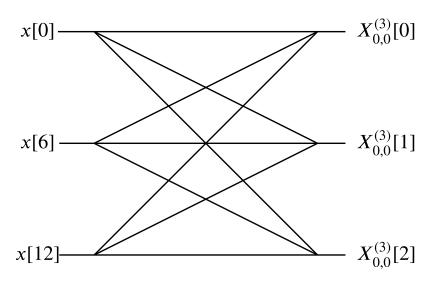
The flow diagram is not required.

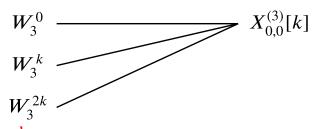
(c) 4 points Stage I

$$X_{p,q}^{(3)}[k] = \sum_{m=0}^{2} x_{p,q}[m]e^{-j\frac{2\pi km}{3}}$$

$$= x_{p,q}[0] + x_{p,q}[1]e^{-j\frac{2\pi k}{3}} + x_{p,q}[2]e^{-j\frac{2\pi 2k}{3}}$$

$$= x_{p,q}[0] + W_3^k x_{p,q}[1] + W_3^{2k} x_{p,q}[2]$$





The flow diagram is not required.

(c) Since

$$x_p[l] = x[2l+p] \tag{1}$$

$$x_{p,q}[m] = x_p[3m+q]$$
 (2)

substitute (1) into (2), we get

$$x_{p,q}[m] = x[6m + 2q + p],$$

where m = 0, 1, 2, q = 0, 1, 2, p = 0, 1.

p	q	m	$x_{p,q}[m]$	6m + 2q + p	x[6m + 2q + p]
0	0	0	$x_{0,0}[0]$	0	x[0]
0	0	1	$x_{0,0}[1]$	6	x[6]
0	0	2	$x_{0,0}[2]$	12	x[12]
0	1	0	$x_{0,1}[0]$	2	x[2]
0	1	1	$x_{0,1}[1]$	8	x[8]
0	1	2	$x_{0,1}[2]$	14	x[14]
0	2	0	$x_{0,2}[0]$	4	x[4]
0	2	1	$x_{0,2}[1]$	10	x[10]
0	2	2	$x_{0,2}[2]$	16	x[16]
1	0	0	$x_{1,0}[0]$	1	x[1]
1	0	1	$x_{1,0}[1]$	7	x[7]
1	0	2	$x_{1,0}[2]$	13	x[13]
1	1	0	$x_{1,1}[0]$	3	x[3]
1	1	1	$x_{1,1}[1]$	9	x[9]
1	1	2	$x_{1,1}[2]$	15	x[15]
1	2	0	$x_{1,2}[0]$	5	x[5]
1	2	1	$x_{1,2}[1]$	11	x[11]
1	2	2	$x_{1,2}[2]$	17	x[17]

Table 1: Input Ordering is shown in the 6th column from top to bottom

Correct ordering: 5 points

(d) In the Stage 3, there are 18 multiplications and 18 additions. So there are 18 COs.

Correct number of COs in Stage 3: 2 points

In the Stage 2, there are 2 multiplications and 2 additions for each of the point in a 9-point DFT, so there are $2 \times 9 \times 2 = 36$ COs.

Correct number of COs in Stage 2: 2 points

In the Stage 1, there are 2 multiplications and 2 additions for each of the point in a 3-point DFT, so there are $2 \times 3 \times 6 = 36$ COs.

Correct number of COs in Stage 1: 1 points

Totally, there are approximately 18 + 36 + 36 = 90 COs.

Consider the 10-point signal $x[n] = \cos\left(\frac{7\pi}{10}n\right)$, n = 0, ..., 9.

a. (15) Determine an exact expression for the 10-point discrete Fourier transform (DFT) $X^{(10)}[k]$, k = 0, ..., 9 in terms of the function $psinc_N(\omega) = \frac{\sin(\frac{\omega N}{2})}{\sin(\frac{\omega}{2})}$.

$$\cos\left(\frac{7\pi}{10}n\right) = \frac{1}{2} \left(e^{j\frac{7\pi}{10}n} + e^{-j\frac{7\pi}{10}n}\right)$$

$$w[n] = u[n] - u[n - 10] \stackrel{DTFT}{\Longleftrightarrow} psinc_{10}(\omega)e^{-j\omega\frac{9}{2}} = \frac{\sin(\frac{10\omega}{2})}{\sin(\frac{\omega}{2})}e^{-j\omega\frac{9}{2}}$$

$$x[n] = w[n] \times \frac{1}{2} \left(e^{j\frac{7\pi}{10}n} + e^{-j\frac{7\pi}{10}n}\right) = \frac{1}{2} \left(w[n]e^{j\frac{7\pi}{10}n} + w[n]e^{-j\frac{7\pi}{10}n}\right)$$

By the modulation property of the DTFT, we get:

$$X(\omega) = \frac{1}{2} \left(p sinc_{10} \left(\omega - \frac{7\pi}{10} \right) e^{-j(\omega - \frac{7\pi}{10})\frac{9}{2}} + p sinc_{10} \left(\omega + \frac{7\pi}{10} \right) e^{-j(\omega + \frac{7\pi}{10})\frac{9}{2}} \right) e^{-j(\omega + \frac{7\pi}{10})\frac{9}{2}}$$

We get the 10-point DFT when we take samples of the DTFT where $\omega = \frac{2\pi}{10}k$, k = 0, ..., 9.

$$X^{(10)}[k] = \frac{1}{2} \left(psinc_{10} \left(\frac{2\pi}{10} k - \frac{7\pi}{10} \right) e^{-j(\frac{2\pi}{10} k - \frac{7\pi}{10})\frac{9}{2}} + \ psinc_{10} \left(\frac{2\pi}{10} k + \frac{7\pi}{10} \right) e^{-j(\frac{2\pi}{10} k + \frac{7\pi}{10})\frac{9}{2}} \right), k = 0, \dots, 9$$

Since $\frac{7\pi}{10} = \frac{2\pi(3.5)}{10}$, we can rewrite this as:

$$X^{(10)}[k] = \frac{1}{2} \left(psinc_{10} \left(\frac{2\pi}{10} (k - 3.5) \right) e^{-j\frac{2\pi}{10} (k - 3.5)\frac{9}{2}} + psinc_{10} \left(\frac{2\pi}{10} (k + 3.5) \right) e^{-j\frac{2\pi}{10} (k + 3.5)\frac{9}{2}} \right),$$

$$k = 0, \dots, 9$$

We normally don't like seeing $k + k_0$ in a DFT, and since $psinc_{10}(\cdot)$ is periodic in relation to k with period 10, we can rewrite it as:

$$psinc_{10}\left(\frac{2\pi}{10}(k+3.5)\right) = psinc_{10}\left(\frac{2\pi}{10}(k-10+3.5)\right) = psinc_{10}\left(\frac{2\pi}{10}(k-6.5)\right)$$

So,

$$X^{(10)}[k] = \frac{1}{2} \left(psinc_{10} \left(\frac{2\pi}{10} (k - 3.5) \right) e^{-j\frac{2\pi}{10} (k - 3.5)\frac{9}{2}} + psinc_{10} \left(\frac{2\pi}{10} (k - 6.5) \right) e^{-j\frac{2\pi}{10} (k - 6.5)\frac{9}{2}} \right),$$

$$k = 0, \dots, 9$$

3 pts. – Correct DTFT of w[n]

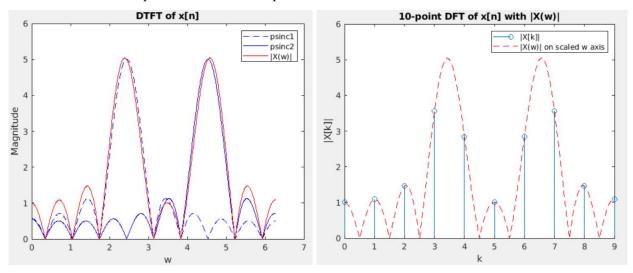
3 pts. – Correct DTFT of x[n]

4 pts. – Correct DFT frequency shift k₀ (2 pts. each)

5 pts. – Correct expression of DFT of x[n]

b. (7) Based on your answer to part a) above, carefully sketch *by hand* the 10-point DFT $X^{(10)}[k], k = 0, ..., 9$. Be sure to dimension both axes. You may ignore the contribution of phase terms in your sketch.

Note: This solution keeps the contribution of phase terms.



4 pts. Correct DFT sketch

2 pts. Dimension axes (1 pt. each)

1 pt. Label axes

c. (3) Discuss whether or not *picket fence effect* or *leakage* are present for this example, and explain how they manifest themselves in the 10-point DFT that you sketched in your answer to part b) above.

Leakage is caused by the truncation of a signal, since the signal $\cos\left(\frac{7\pi}{10}n\right)$ is truncated to only be on the interval $n=0,\ldots,9$, this example does experience leakage. The leakage is manifest in the DFT by the $psinc_{10}(\cdot)$.

The picket fence effect is caused by sampling in the frequency domain. For a sinusoidal signal, it appears if the frequency of the sinusoid, ω_0 , is not an integer multiple of $\frac{2\pi}{N}$. In this case, N=10, and the frequency of the sinusoid is $\omega_0 = \frac{7\pi}{10} = \frac{2\pi(3.5)}{10}$. The picket fence effect is manifest in this DFT by the samples being offset from the center of the main lobe of each $psinc_{10}(\cdot)$.

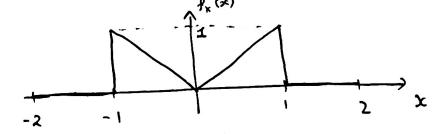
1 pt. – Saying leakage is present

1 pt. – Saying picket fence effect is present

0.5 pt. – Explain how leakage is manifest

0.5 pt. – Explain how picket fence effect is manifest

$$\int_{X} (x) = \begin{cases} |x| & 0 \le |x| \le 1 \\ 0 & \text{else} \end{cases}$$



b.
$$E[X] = \int_{-\infty}^{\infty} x \, f_{x^2x^3} dx = \int_{-1}^{1} x \, 1x 1 \, dx = \int_{-1}^{\infty} x \cdot (-x) \, dx + \int_{0}^{1} x \cdot x \, dx$$

$$= \int_{-\infty}^{1} -x^2 \, dx + \int_{0}^{1} x^2 \, dx = -(0 - \frac{1}{3}) + (\frac{1}{3} - 0) = 0$$

$$E[x^{2}] = \int_{-\infty}^{\infty} x^{2} f_{x}(x) dx = \int_{-1}^{\infty} x^{2} \cdot (-x) dx + \int_{0}^{1} x^{2} \cdot x dx$$
$$= -\frac{x^{4}}{4} \int_{0}^{\infty} = \frac{1}{4} + \frac{1}{4} = \frac{1}{2}$$

$$f_{x} = E[x] = 0$$
 $f_{x}^{2} = E[x^{2}] - (E[x])^{2} = \frac{1}{2}$

c.
$$Q(x) = \begin{cases} -\frac{2}{3} & 4 \le 3 \le \frac{1}{3} \\ 0 & -\frac{1}{3} \le x \le \frac{1}{3} \\ \frac{2}{3} & \frac{1}{3} \le x \le 1 \end{cases}$$

$$\Delta = \frac{2}{3}$$
 $\mathcal{E}_{approx} = \frac{\Delta^2}{12} = \frac{4}{9.12} = \frac{1}{27}$

$$\begin{aligned}
\delta \cdot & \mathcal{E}_{e_{Xa_{x}A}} = \mathcal{E}\left[|Y-X|^{2} \right] = \int_{-\infty}^{\infty} \left(Q_{(x)-x} \right)^{2} \int_{X_{x}} (x) dx \\
&= \int_{-1}^{\frac{1}{3}} \left(\frac{2}{3} - x \right)^{2} |x| dx + \int_{-1}^{\infty} \left(\nabla - x \right)^{2} |x| dx + \int_{-1}^{\infty} \left(\frac{2}{3} - x \right)^{2} |x| dx \\
&= \int_{-1}^{\frac{1}{3}} \left(\frac{2}{3} - x \right)^{2} |x| dx + \int_{-1}^{\frac{1}{3}} \left(\nabla - x \right)^{2} |x| dx + \int_{-1}^{\infty} \left(\frac{2}{3} - x \right)^{2} |x| dx \\
&= \int_{-1}^{\frac{1}{3}} \left(\frac{2}{3} - x \right)^{2} (-x) dx = \int_{-1}^{-\frac{1}{3}} \cdot \frac{q}{3} x - \frac{q}{3} x^{2} - x^{3} dx = \int_{-\frac{1}{3}}^{-\frac{q}{3}} x^{2} - \frac{q}{3} x^{3} + \frac{q}{4} \right) \Big]_{-1}^{\frac{1}{3}} \\
&= -\frac{2}{3} \cdot \frac{1}{3} + \frac{q}{4} \cdot \frac{1}{2} - \frac{1}{4} \cdot \frac{1}{81} + \frac{2}{3} - \frac{q}{3} + \frac{1}{4} = \frac{q}{243} \\
&= \frac{2}{3} - \frac{q}{3} + \frac{1}{4} - \frac{2}{3} \cdot \frac{1}{3} + \frac{q}{3} \cdot \frac{1}{23} - \frac{1}{4} \cdot \frac{1}{81} = \frac{q}{243} \\
&= \frac{1}{3} - \frac{q}{3} + \frac{1}{4} - \frac{2}{3} \cdot \frac{1}{3} + \frac{q}{3} \cdot \frac{1}{23} - \frac{1}{4} \cdot \frac{1}{81} = \frac{q}{243} \\
&= \frac{1}{3} - \frac{q}{3} + \frac{1}{4} - \frac{1}{324} = \frac{1}{162} \\
&= \frac{1}{324} + \frac{1}{324} = \frac{1}{162} \\
&= \frac{1}{324} + \frac{1}{324} = \frac{1}{162} = \frac{19}{486} \end{aligned}$$

e.
$$\mathcal{E}_{e_{x_{a_{c}}}} = \frac{19}{486}$$

$$\mathcal{E}_{a_{porox}} = \frac{1}{2\lambda} = \frac{18}{486}$$

Approximate and exact quantization errors are very close.