

- You have 50 minutes to work the following four problems.
- Be sure to show all your work to obtain full credit.
- The exam is closed book and closed notes.
- Calculators are permitted.

1. (25 pts.) Consider a causal linear, time-invariant system with transfer function

$$H(z) = \frac{1}{\left(1 + \frac{1}{2}z^{-1}\right)\left(1 - \frac{1}{2}z^{-1}\right)}.$$

- (3) Find the region of convergence for $H(z)$. Is this system BIBO stable?
- (13) Find the impulse response $h[n]$.
- (9) Use the graphical approach to find the magnitude and phase of the frequency response at the frequency $\omega = \pi/2$ radians/sample.

(a) given system is causal, poles located at $\frac{1}{2}$ and $-\frac{1}{2}$

$$\text{ROC: } |z| > \frac{1}{2}$$

Since the ROC includes the unit circle, system is BIBO stable.

$$b) H(z) = \frac{1}{\left(1 + \frac{1}{2}z^{-1}\right)\left(1 - \frac{1}{2}z^{-1}\right)} = \frac{A}{1 + \frac{1}{2}z^{-1}} + \frac{B}{1 - \frac{1}{2}z^{-1}}$$

$$A = H(z)\left(1 + \frac{1}{2}z^{-1}\right)\Big|_{z=-\frac{1}{2}} = \frac{1}{1 - \frac{1}{2}z^{-1}}\Big|_{z=-\frac{1}{2}} = \frac{1}{1 - \frac{1}{2}(-2)} = \frac{1}{2}$$

$$B = H(z)\left(1 - \frac{1}{2}z^{-1}\right)\Big|_{z=\frac{1}{2}} = \frac{1}{1 + \frac{1}{2}z^{-1}}\Big|_{z=\frac{1}{2}} = \frac{1}{2}$$

$$\Rightarrow H(z) = \frac{1}{2} \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{1}{2} \frac{1}{1 - \frac{1}{2}z^{-1}}$$

Since system is causal, $h[n]$ is right sided

$$h[n] = \frac{1}{2} \left(-\frac{1}{2}\right)^n u[n] + \frac{1}{2} \left(\frac{1}{2}\right)^n u[n]$$

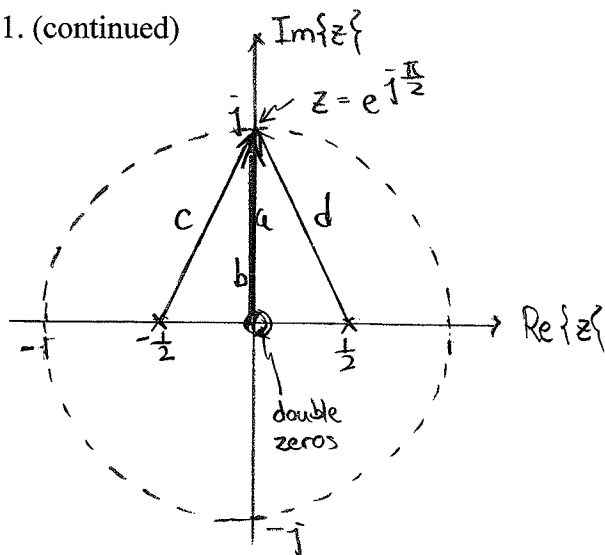
$$= \frac{1}{2} u[n] \left\{ \left(-\frac{1}{2}\right)^n + \left(\frac{1}{2}\right)^n \right\}$$

$$= \begin{cases} \frac{1}{2} u[n] \left\{ 2 \left(\frac{1}{2}\right)^n \right\} & \text{for } n \text{ even} \\ 0 & \text{for } n \text{ odd} \end{cases}$$

$$= \begin{cases} \left(\frac{1}{2}\right)^n u[n] & \text{for } n \text{ even} \\ 0 & \text{for } n \text{ odd} \end{cases}$$

1. (continued)

(c)



$$H(z) = \frac{z^2}{(z + \frac{1}{2})(z - \frac{1}{2})}$$

double zeros at $z = 0$ poles at $z = \pm \frac{1}{2}$

$$|H(e^{j\frac{\pi}{2}})| = \frac{|a||b|}{|c||d|}$$

$$|a| = |b| = 1$$

$$|c| = |d| = \sqrt{(\frac{1}{2} - 0)^2 + (0 - 1)^2} = \sqrt{\frac{5}{4}}$$

$$= \frac{1}{\sqrt{\frac{5}{4}} \sqrt{\frac{5}{4}}}$$

$$= \frac{4}{5} = 0.8$$

$$\angle H(e^{j\frac{\pi}{2}}) = \angle a + \angle b - \angle c - \angle d$$

$$\text{note: } \angle a = \angle b = \frac{\pi}{2}$$

$$\angle d = \pi - \angle c$$

$$\therefore \angle H(e^{j\frac{\pi}{2}}) = \frac{\pi}{2} + \frac{\pi}{2} - \cancel{\angle c} - \pi + \cancel{\angle c}$$

$$= \pi - \pi$$

$$= 0$$

2. (25 pts.) For this problem, you should use the formulas provided in the handout as much as possible.

Consider the 6-point signal $x[n] = \cos(\pi n / 3)$, $n = 0, \dots, 5$.

- a. (8) Find and sketch the 6-point DFT $X[k]$ of this signal.

Now consider the 6-point signal $y[n] = \cos(\pi n / 6)$, $n = 0, \dots, 5$.

- b. (15) Find and sketch the 6-point DFT $Y[k]$ of this signal.
c. (2) Explain the relationship between $X[k]$ and $Y[k]$.

(a) $x[n] = \cos(\pi n / 3)$ $n = 0, \dots, 5$

$$= \frac{1}{2} \left[e^{+j\frac{\pi n}{3}} + e^{-j\frac{\pi n}{3}} \right] \quad n = 0, \dots, 5$$

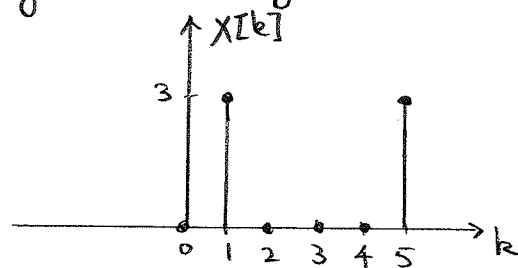
$$= \frac{1}{2} \left[e^{+j\frac{2\pi n}{6}} + e^{-j\frac{2\pi n}{6}} \right] \quad n = 0, \dots, 5$$

using transform relation:

$$\cos(2\pi k_0 n / N), 0 \leq k \leq N-1 \leftrightarrow \frac{N}{2} \{ \delta(k - k_0) + \delta(k - (N - k_0)) \} \quad 0 \leq k \leq N-1$$

for this case: $k_0 = 1$, $N = 6$

$$X[k] = 3 \{ \delta(k - 1) + \delta(k - 5) \}, \quad 0 \leq k \leq 5$$



(b) $y[n] = \cos(\pi n / 6)$, $n = 0, \dots, 5$

$$= \frac{1}{2} \left\{ e^{j\pi n / 6} + e^{-j\pi n / 6} \right\} \quad n = 0, \dots, 5$$

using transform relation:

$$e^{j\omega_0 n}, n = 0, \dots, N-1 \leftrightarrow \frac{\sin((\frac{2\pi k}{N} - \omega_0)(\frac{N}{2}))}{\sin((\frac{2\pi k}{N} - \omega_0)(\frac{1}{2}))} e^{-j(\frac{2\pi k}{N} - \omega_0)(\frac{N-1}{2})}$$

↓

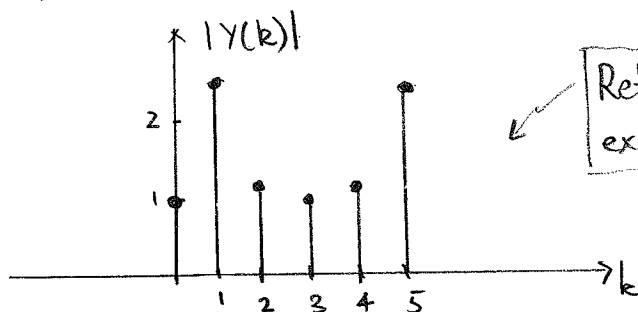
2. (continued)

(b) continued

for this case: $\omega_0 = \frac{\pi}{6}$, $N=6$

$$\begin{aligned}
 e^{-j\frac{\pi}{6}n}, n=0, \dots, N-1 &\leftrightarrow \frac{\sin\left(\left(\frac{2\pi k}{6} - \frac{\pi}{6}\right)3\right)}{\sin\left(\left(\frac{2\pi k}{6} - \frac{\pi}{6}\right)\frac{1}{2}\right)} e^{-j\left(\frac{2\pi k}{6} - \frac{\pi}{6}\right)\left(\frac{5}{2}\right)} \\
 &= \frac{\sin\left(\frac{2\pi}{6}\left(k - \frac{1}{2}\right)3\right)}{\sin\left(\frac{2\pi}{6}\left(k - \frac{1}{2}\right)\frac{1}{2}\right)} e^{-j\frac{2\pi}{6}\left(k - \frac{1}{2}\right)\left(\frac{5}{2}\right)} \\
 &= \frac{\sin\left(\pi\left(k - \frac{1}{2}\right)\right)}{\sin\left(\frac{\pi}{6}\left(k - \frac{1}{2}\right)\right)} e^{-j\frac{5\pi}{6}\left(k - \frac{1}{2}\right)}
 \end{aligned}$$

$$\therefore Y(k) = \frac{1}{2} \left[\frac{\sin\left(\pi\left(k - \frac{1}{2}\right)\right)}{\sin\left(\frac{\pi}{6}\left(k - \frac{1}{2}\right)\right)} e^{-j\frac{5\pi}{6}\left(k - \frac{1}{2}\right)} + \frac{\sin\left(\pi\left(k + \frac{1}{2}\right)\right)}{\sin\left(\frac{\pi}{6}\left(k + \frac{1}{2}\right)\right)} e^{-j\frac{5\pi}{6}\left(k + \frac{1}{2}\right)} \right]$$



Refer to next page for explanation of plot

- (c) Leakage occurs for $Y(k)$, whereas no leakage occurs for $X(k)$. Both $Y(k)$ and $X(k)$ are DFT's of 6-pt windowed cosines.

2.(b) Clarification

let $y[n] = \cos(\omega_0 n)$, $n=0, \dots, N-1$

$$Y(k) = \frac{1}{2} \left\{ \underbrace{\frac{\sin\left(\left(\frac{2\pi k}{N} - \omega_0\right)\frac{N}{2}\right)}{\sin\left(\left(\frac{2\pi k}{N} - \omega_0\right)\frac{1}{2}\right)} e^{-j\left(\frac{2\pi k}{N} - \omega_0\right)\frac{N-1}{2}}}_{\text{psinc}_1} + \underbrace{\frac{\sin\left(\left(\frac{2\pi k}{N} + \omega_0\right)\frac{N}{2}\right)}{\sin\left(\left(\frac{2\pi k}{N} + \omega_0\right)\frac{1}{2}\right)} e^{j\left(\frac{2\pi k}{N} + \omega_0\right)\frac{N-1}{2}}}_{\text{psinc}_2} \right\}$$

For N large, psinc_1 and psinc_2 has minimal overlap
 $\Rightarrow$ can approximately sketch as one psinc located at ω_0
 and another located at $-\omega_0$.

But

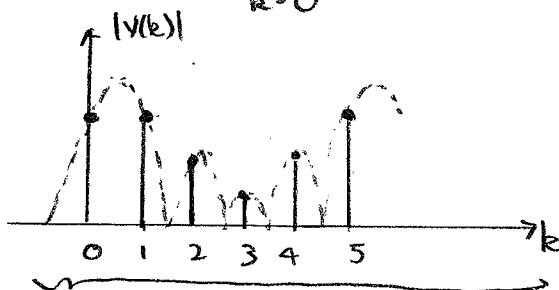
For N small (such as the case here with $N=6$), psinc_1 and psinc_2 has significant overlap. The values for $k=0, \dots, 5$ are listed below for psinc_1 and psinc_2 , $\omega_0 = \pi/6$.

k	$\frac{1}{2}\text{psinc}_1(k)$	$\frac{1}{2}\text{psinc}_2(k)$	$\frac{1}{2}\{\text{psinc}_1 + \text{psinc}_2\} = Y(k)$
0	$0.5 + 1.867i$	$0.5 - 1.867i$	1
1	$0.5 - 1.867i$	$0.5 - 0.5i$	$1 - 2.366i$
2	$0.5 - 0.5i$	$0.5 - 0.134i$	$1 - 0.634i$
3	$0.5 - 0.134i$	$0.5 + 0.134i$	1
4	$0.5 + 0.134i$	$0.5 + 0.5i$	$1 + 0.634i$
5	$0.5 + 0.5i$	$0.5 + 1.866i$	$1 + 2.366i$

note the significant overlap
 at each k , also note
 that the values are complex

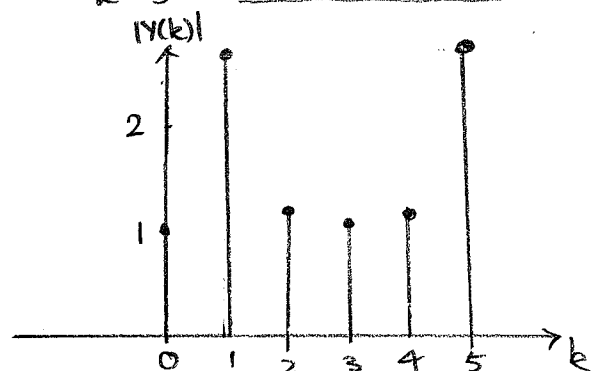
$$\therefore |Y(k)| = [1 \quad 2.569 \quad 1.184 \quad 1 \quad 1.184 \quad 2.569]$$

$\uparrow$ $k=0$ $\uparrow$ $k=5$



this is not a very accurate
 graph due to overlap
 between the two psinc's
 for N small

$\Rightarrow$



This is an accurate graph.

Key point
 $|Y(k)| = \frac{1}{2}|\text{psinc}_1 + \text{psinc}_2|$
 $\neq \frac{1}{2}|\text{psinc}_1| + \frac{1}{2}|\text{psinc}_2|$
 for N small

3. a. (21) Derive a decimation-in-time FFT algorithm for a 9-point DFT, and draw a complete flow diagram for the algorithm.
- b. (4) Calculate the approximate number of complex operations required to compute the 9-point FFT and compare with the number of complex operations required to compute the 9-point DFT directly.

(a) $9 = 3 \times 3$, so split the 9-pt DFT into 3 3-pt DFTs using decimation in time

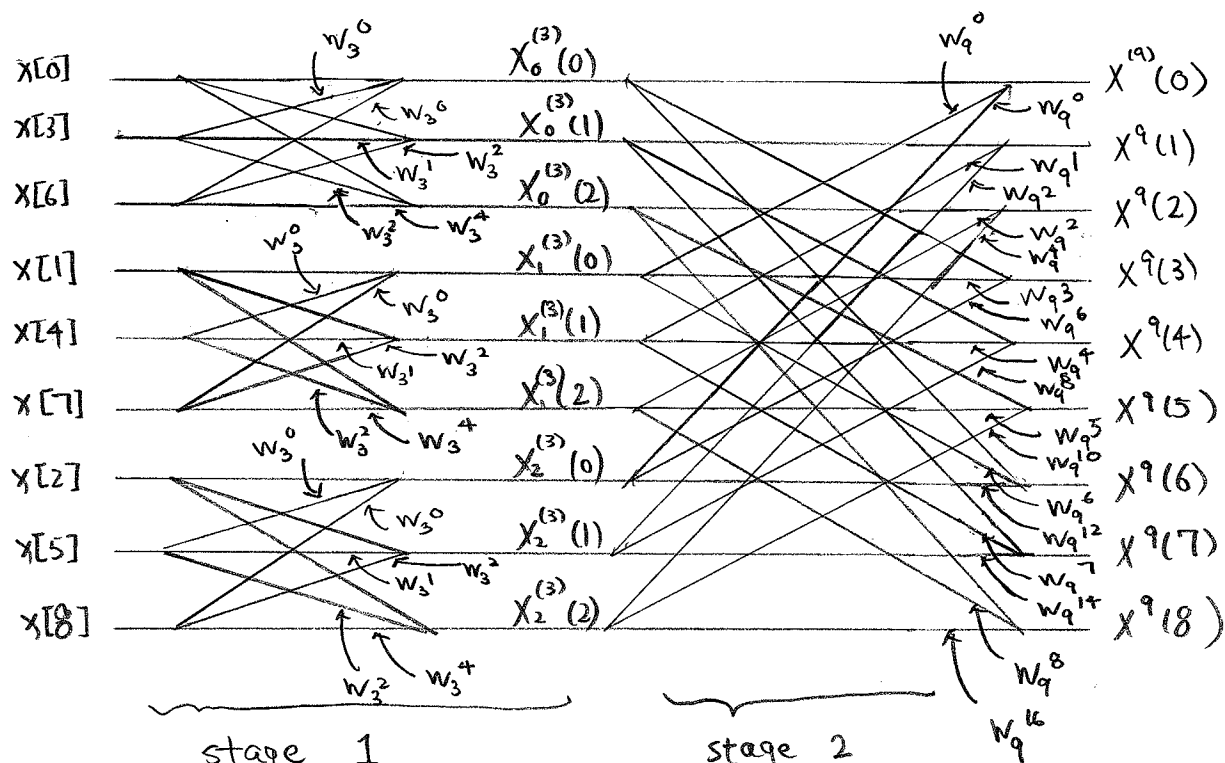
$$\begin{aligned}
 X^{(9)}(k) &= \sum_{n=0}^8 x[n] e^{-j2\pi kn/9} \\
 &= \sum_{n=0}^2 x[3n] e^{-j2\pi k(3n)/9} + \sum_{n=0}^2 x[3n+1] e^{-j2\pi k(3n+1)/9} \\
 &\quad + \sum_{n=0}^2 x[3n+2] e^{-j2\pi k(3n+2)/9} \quad \downarrow \\
 &= X_0^{(3)}(k) + e^{-j\frac{2\pi}{9}k} X_1^{(3)}(k) + e^{-j\frac{2\pi}{9}2k} X_2^{(3)}(k) \\
 &= X_0^{(3)}(k) + W_9^k X_1^{(3)}(k) + W_9^{2k} X_2^{(3)}(k)
 \end{aligned}$$

For each 3-pt DFT

$$\begin{aligned}
 Y^{(3)}(k) &= \sum_{n=0}^2 y[n] e^{-j\frac{2\pi}{3}kn} \\
 &= y[0] + y[1] e^{-j\frac{2\pi}{3}k} + y[2] e^{-j\frac{4\pi}{3}k} \\
 &= y[0] + W_3^k y[1] + W_3^{2k} y[2]
 \end{aligned}$$

↓ diagram on next page

3. (continued)



(b)

$$\begin{aligned}
 & 3(3)(2) = 18 \text{ multiplies} + 9(2) = 18 \text{ multiplies} = 36 \text{ multiplies} \\
 & 3(3)(2) = 18 \text{ adds} + 9(2) = 18 \text{ adds} = 36 \text{ adds} \\
 \Rightarrow & 18 + 18 + 18 + 18 = 72 \text{ complex operations} \\
 & \text{counting additions and multiplications}
 \end{aligned}$$

fft {

$$\text{DFT} \left\{ \begin{aligned} & N^2 \text{ multiplies} = 81 \\ & + \\ & N(N-1) = N^2 - N \text{ adds} = 72 \end{aligned} \right.$$

153 complex operations
counting additions and multiplications

4. (25 pts) Let θ be a random variable that is uniformly distributed on the interval $[0, 2\pi)$. Define a new random variable $X = \cos(2\pi f_0 t + \theta)$ for some fixed value of f_0 and a fixed value of t .

a) (8) Find the mean of X .

b) (8) Find the variance of X .

Now suppose that we define a second random variable $Y = \cos(2\pi f_0 s + \theta)$, where f_0 has the same value as before and s is a fixed parameter $\neq t$.

c) (9) Find the expected value of the product XY .

Note: The following formula may be of benefit to you in working this problem.

$$\cos \alpha \cos \beta = \frac{1}{2} \cos(\alpha + \beta) + \frac{1}{2} \cos(\alpha - \beta)$$

$g(\theta)$: this is a function of θ

$$\begin{aligned} (a) \quad E[X] &= E[\cos(2\pi f_0 t + \theta)] = E[g(\theta)] = \int_{-\infty}^{\infty} g(\theta) f_{\theta}(\theta) d\theta \\ &= \int_{-\infty}^{\infty} \cos(2\pi f_0 t + \theta) f_{\theta}(\theta) d\theta \\ &= \frac{1}{2\pi} \int_0^{2\pi} \cos(2\pi f_0 t + \theta) d\theta \quad \text{given that } \theta \text{ is uniformly distributed} \\ &= \frac{1}{2\pi} \sin(2\pi f_0 t + \theta) \Big|_0^{2\pi} \\ &= \frac{1}{2\pi} [\sin(2\pi f_0 t + 2\pi) - \sin(2\pi f_0 t)] \\ &= \frac{1}{2\pi} [\sin(2\pi f_0 t) - \sin(2\pi f_0 t)] \\ &= 0 \end{aligned}$$

$$\begin{aligned} (b) \quad E[X^2] &= E[\cos^2(2\pi f_0 t + \theta)] \\ &= \int_{-\infty}^{\infty} \cos^2(2\pi f_0 t + \theta) f_{\theta}(\theta) d\theta \quad \text{note: } \cos^2 x = \frac{1}{2} + \frac{1}{2} \cos 2x \\ &= \frac{1}{2\pi} \int_0^{2\pi} \left(\frac{1}{2} + \frac{1}{2} \cos(4\pi f_0 t + 2\theta) \right) d\theta \\ &= \frac{1}{2\pi} \int_0^{2\pi} \frac{1}{2} d\theta + \int_0^{2\pi} \frac{1}{2} \cos(4\pi f_0 t + 2\theta) d\theta \rightarrow 0 \text{ as before} \\ &= \frac{1}{2\pi} \times \pi = \frac{1}{2} \end{aligned}$$

$$\begin{aligned} \text{Var}(X) &= E[X^2] - (E[X])^2 = \frac{1}{2} - 0^2 \\ &= \frac{1}{2} \end{aligned}$$

4. (continued)

$$(c) \quad Y = \cos(2\pi f_0 s + \Theta)$$

$$\begin{aligned} \text{let } Z = XY &= \cos(2\pi f_0 t + \Theta) \cos(2\pi f_0 s + \Theta) \\ &= \frac{1}{2} \cos(2\pi f_0 t + \Theta + 2\pi f_0 s + \Theta) + \frac{1}{2} \cos(2\pi f_0 t + \Theta - 2\pi f_0 s - \Theta) \\ &= \frac{1}{2} \cos(2\pi f_0(t+s) + 2\Theta) + \frac{1}{2} \cos(2\pi f_0(t-s)) \end{aligned}$$

note that $Z = g(\Theta)$: ie. Z is a function of $\Theta \Rightarrow E[g(\Theta)] = \int_{-\infty}^{\infty} g(\theta) f_{\Theta}(\theta) d\theta$

$$\therefore E[XY] = E\left[\frac{1}{2} \cos(2\pi f_0(t+s) + 2\Theta) + \frac{1}{2} \cos(2\pi f_0(t-s))\right]$$

$$= \frac{1}{2} E[\cos(2\pi f_0(t+s) + 2\Theta)] + \frac{1}{2} E[\underbrace{\cos(2\pi f_0(t-s))}_{\text{this is not a function of } \Theta}]$$

$$= \frac{1}{2} \int_{-\infty}^{\infty} \cos(2\pi f_0(t+s) + 2\theta) f_{\Theta}(\theta) d\theta + \frac{1}{2} \cos(2\pi f_0(t-s))$$

$$= \frac{1}{2} \int_0^{2\pi} \cos(2\pi f_0(t+s) + 2\theta) d\theta \xrightarrow{\text{0 as before}} + \frac{1}{2} \cos(2\pi f_0(t-s))$$

$$= \frac{1}{2} \cos(2\pi f_0(t-s))$$

1. _____

2. _____

3. _____

4. _____

Total _____