

ECE 438 Lecture

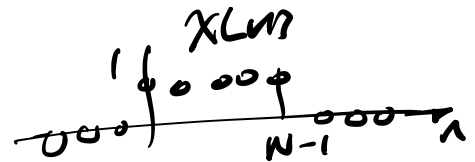
1/27/2023

DTFT

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n}$$

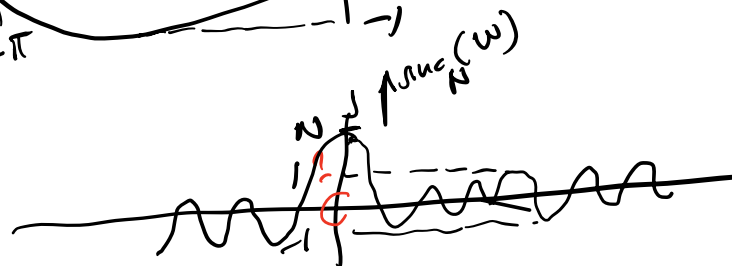
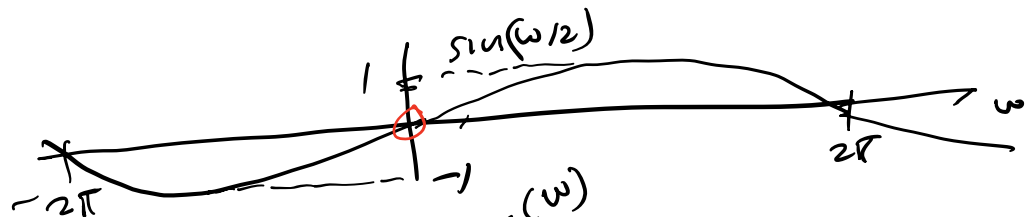
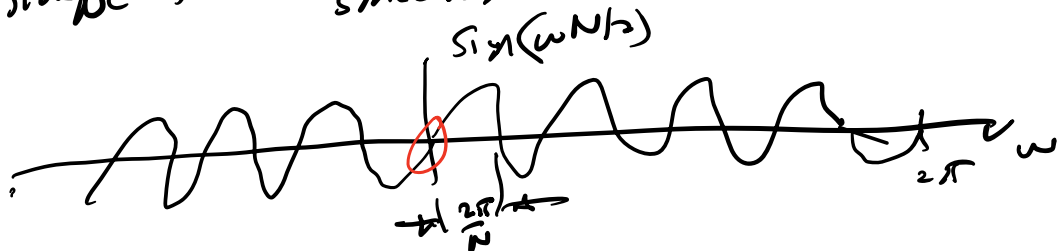
Example

$$x[n] = \begin{cases} 1, & 0 \leq n \leq N-1 \\ 0, & \text{else} \end{cases}$$



$$X(\omega) = \text{sinc}_N(\omega) e^{j\omega(N-1)/2}$$

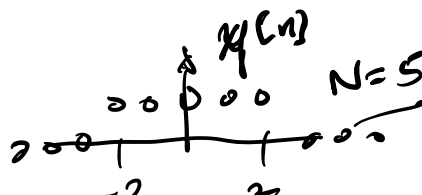
$$\text{sinc}_N(\omega) = \frac{\sin(\omega N/2)}{\sin(\omega/2)}$$



Taylor series expansion for $\sin(\theta) = \theta + \dots$

$$\lim_{\omega \rightarrow 0} \text{sinc}_N(\omega) = \lim_{\omega \rightarrow 0} \frac{\omega N/2}{\omega} = N$$

for N odd, $x[n]$ has an odd number of points $= 1 \Rightarrow \frac{N-1}{2}$ is an integer



$$y[n] = x\left[n + \frac{N-1}{2}\right]$$

$$\Rightarrow Y(\omega) = X(\omega) e^{j\left(\frac{N-1}{2}\right)\omega}$$

$$\therefore Y(\omega) = \text{sinc}_N(\omega)$$

sinc function is defined on formulas sheet

This material is from Module 1.3.3.

NEW TOPIC (Module 1.4.4)

Continuous-time signals & systems
 $\uparrow$
 (CT)

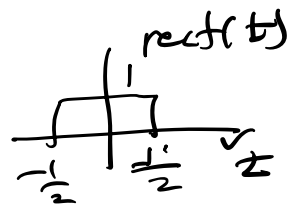
Why study CT in a course on DSP?

Ans. 1: We live in an analog or continuous-parameter world.

Ans. 2: The theory of CT signals & systems is more symmetric than that of DT signals & systems

Special CT signals:

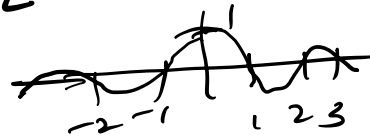
$$\textcircled{1} \text{rect}(t) = \begin{cases} 1, & |t| < 1/2 \\ 0, & \text{else} \end{cases}$$



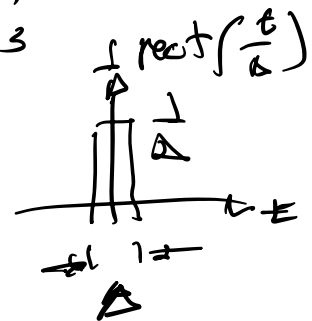
$$\textcircled{2} \text{sinc}(t) \equiv \frac{\sin(\pi t)}{\pi t}$$

sinc(t)

why not $\frac{\sin(t)}{t}$?



$$\textcircled{3} \delta(t) = \lim_{\Delta \rightarrow 0} \frac{1}{\Delta} \text{rect}\left(\frac{t}{\Delta}\right)$$



area $A_s = 1$ for all Δ

properties of $\delta(t)$

$$\textcircled{a} \int_{-\infty}^{\infty} x(t) \delta(t-t_0) dt = x(t_0)$$

sifting or sampling property

$$\textcircled{b} x(t) \delta(t-t_0) = x(t_0) \delta(t-t_0)$$

i.e. $\int x(\tau) \delta(\tau - \underline{t - t_0}) d\tau = x(t - t_0)$

$$\textcircled{c} \delta(\underline{at+b}) \equiv \frac{1}{|a|} \delta(\underline{\frac{t}{a} + \frac{b}{a}})$$

proof:

$$\int_{-\infty}^{\infty} x(t) \delta(at+b) dt =$$

$$\text{let } \tau = at \Rightarrow d\tau = a dt \Rightarrow dt = \frac{d\tau}{a}$$

(assume $a > 0$) $t = \frac{\tau}{a}$

have $\frac{1}{a} \int_{-\infty}^{\infty} x\left(\frac{\tau}{a}\right) \delta(\tau+b) d\tau = x\left(\frac{b}{a}\right) \frac{1}{a}$

Note that also by sifting property

$$\int_{-\infty}^{\infty} x(t) \delta\left(t + \frac{b}{a}\right) dt = \frac{1}{|a|} x\left(\frac{b}{a}\right)$$

↑ accounts for
 $a < 0$

$$\text{So } \delta(at+b) \equiv \frac{1}{|a|} \delta\left(t + \frac{b}{a}\right)$$

Continue with CTFT:

Forward: $X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt$

Inverse: $x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi f t} df$

Note symmetry

why not $\omega = 2\pi f$ rad/sec? f - Hz cycles/sec

① more symmetric

② H_2 is more intuitive than Adams
sample

Properties

① Linearity

$$a_1 x_1(t) + a_2 x_2(t) \xrightarrow{\text{CFT}} a_1 X_1(f) + a_2 X_2(f)$$

② Scaling & shifting

$$x\left(\frac{t-t_0}{a}\right) \xrightarrow{\text{CFT}} a X(af) e^{-j2\pi f t_0}$$

$\frac{1}{a}$ " " " " " "

③ Modulation

$$e^{j2\pi f_0 t} x(t) \xrightarrow{\text{CFT}} X(f - f_0)$$

④ Initial values

$$\begin{aligned} \text{a) } X(0) &= \int_{-\infty}^{\infty} x(t) e^{\frac{j2\pi f t}{1}} dt \Big|_{f=0} \\ &= \int_{-\infty}^{\infty} x(t) dt = A_x \end{aligned}$$

$$\begin{aligned} \text{b) } x(0) &= \int_{-\infty}^{\infty} X(f) e^{\frac{j2\pi f t}{1}} df \Big|_{t=0} \\ &= \int_{-\infty}^{\infty} X(f) df = A_X \end{aligned}$$

⑤ Parseval

$$\int |x(t)|^2 dt = \int |X(f)|^2 df$$

$$\text{or } E_x = E_x$$

⑥ Reciprocity

$$\text{Suppose } \underline{x(t)} \xrightarrow{\text{CTFT}} \underline{X(f)}$$

$$\text{Note let } y(t) = X(\underline{t})$$

$$\text{then } Y(f) = x(-f)$$

⑦ Convolution

$$\int x_1(\tau) x_2(t-\tau) d\tau \xrightarrow{\text{CTFT}} X_1(f) X_2(f)$$

$$x_1(t) * x_2(t) \xrightarrow{\text{CTFT}} X_1(f) X_2(f)$$

⑧ Product

$$x_1(t) x_2(t) \xrightarrow{\text{CTFT}} X_1(f) * X_2(f)$$

Transform pairs

① $\delta(t) \xrightarrow{\text{CTFT}} 1$

$$X(f) = \int_{-\infty}^{\infty} \delta(t) \underline{e^{-j2\pi ft}} dt = 1$$

② $1 \xrightarrow{\text{CTFT}} \underline{\delta(f)}$ by reciprocity

$$(3) e^{j2\pi f_0 t} \cdot \underline{1} \leftrightarrow \delta(f - f_0)$$

$$(4) \cos(2\pi f_0 t) = \frac{1}{2} \{ e^{j2\pi f_0 t} + e^{-j2\pi f_0 t} \}$$

$$\cos(2\pi f_0 t) \xleftrightarrow{\text{CTFT}} \frac{1}{2} \{ \delta(f - f_0) + \delta(f + f_0) \}$$

note that CTFT is ^{generally} not periodic

in $f \Rightarrow$ don't need to restrict domain of f

$$(5) \text{rect}(t) \xleftrightarrow{\text{CTFT}} \text{sinc}(f)$$

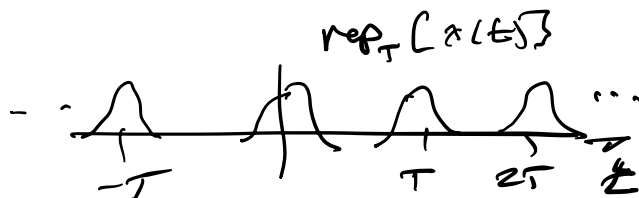
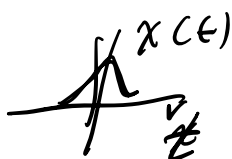
$$(6) \text{sinc}(t) \xleftrightarrow{\text{CTFT}} \text{rect}(f)$$

by reciprocity

Now we introduce two special operators;

① replication

$$\underline{\text{rep}_T[x(t)]} = \sum_{k=-\infty}^{\infty} x(t - kT)$$



② comb