Midterm #2 of ECE301-003, 004 (CRN 17101, 17102) 8–9pm, Thursday, October 10, 2019, WTHR 200.

- 1. Please make sure that it is your name printed on the exam booklet. Enter your student ID number, and signature in the space provided on this page, **NOW!**
- 2. This is a closed book exam.
- 3. This exam contains multiple choice questions and work-out questions. For multiple choice questions, there is no need to justify your answers. You have one hour to complete it. The students are suggested not spending too much time on a single question, and working on those that you know how to solve.
- 4. Use the back of each page for rough work.
- 5. Neither calculators nor help sheets are allowed.

Name:

Student ID:

As a Boiler Maker pursuing academic excellence, I pledge to be honest and true in all that I do. Accountable together — We are Purdue.

Signature:

Date:

Question 1: [20%, Work-out question, Learning Objectives 1, 2, and 3]

- 1. [2%] What does the acronym *LTI* stand for?
- 2. [14%] Consider a DT-LTI system with the impulse response being

$$h[n] = \begin{cases} n+1 & \text{if } -1 \le n \le 2\\ 5-n & \text{if } 3 \le n \le 5\\ 0 & \text{otherwise} \end{cases}$$
(1)

Suppose we use x[n] = U[n+1] - U[n] + U[n-2] - U[n-3] as the input to this LTI system and denote the corresponding output by y[n]. Plot y[n] for the range of $-8 \le n \le 8$.

Hint 1: There is no need to find the mathematical expression of y[n]. We only ask you to plot y[n].

Hint 2: It may be useful to plot x[n] first.

3. [4%] Is the above system causal? Please carefully write down your reasons. A yes/no answer without reasons will not receive any point.

Question 2: [15%, Work-out question, Learning Objectives 1, 2, and 3] Consider the following two signals:

$$x_1(t) = e^{-2t}U(t)$$
 (2)

$$x_2(t) = e^{-jt} U(t - \pi)$$
(3)

1. Compute the convolution $y(t) = x_1(t) * x_2(t)$.

Hint 1: There is no need to greatly simplify your answers. Some expressions of the form of $\frac{3}{3+4j}e^{j\sqrt{2}t}$ would suffice.

Question 3: [18%, Work-out question, Learning Objectives 2, 3, and 4]

Consider a DT-LTI system with impulse response

$$h[n] = \begin{cases} e^{(-0.5+3j)n} & \text{if } n \ge 0\\ 0 & \text{otherwise} \end{cases}$$
(4)

Find the output y[n] when the input is $x[n] = \cos(3n + \pi/3)$.

Hint 1: There is no need to greatly simplify your answers. Some expressions of the form of $\frac{3}{3+4j}e^{j\sqrt{2}t} + \frac{8}{3-2j}e^{j\sqrt{3}t}$ would suffice. Hint 2: The following formulas may be useful:

$$\cos(\theta) = \frac{e^{j\theta} + e^{-j\theta}}{2} \tag{5}$$

$$\sin(\theta) = \frac{e^{j\theta} - e^{-j\theta}}{2j} \tag{6}$$

$$\sum_{k=1}^{\infty} ar^{k-1} = \frac{a}{1-r} \quad \text{if } |r| < 1.$$
(7)

Question 4: [19%, Work-out question, Learning Objectives 4 and 5]

1. [14%] Consider a periodic CT signal

$$x(t) = \begin{cases} \pi & \text{if } 0 < t < 2\\ 0 & \text{if } 2 < t < 5 \end{cases}$$
(8)
periodic with period $T = 5$

Find the Fourier series coefficients a_k of x(t).

2. [5%] Find the values of $\sum_{k=-\infty}^{\infty} |a_k|^2$. Hint 1: If you do not know the answer of the previous subquestion, you can assume that

$$a_k = \begin{cases} \frac{\sin(0.7k)}{k} & \text{if } k \neq 0\\ 0.7 & \text{if } k = 0. \end{cases}$$
(9)

You will receive full credit if your answer is correct.

Hint 2: You may need to use the tables in the end of the exam booklet.

Question 5: [8%, Work-out question, Learning Objective 4] Consider a DT periodic signal of period N = 8. Suppose we also know that its FS coefficients are $a_1 = 1 + j$, $a_7 = 1 - j$, and $a_k = 0$ for k = 0, 2, 3, 4, 5, 6.

Find the expression of x[n].

Hint: In this question, x[n] is a real-valued signal. Your answer cannot have the imaginary number j in the expression. You need to simplify the expression so that there is no j inside.

Question 6: [20%, Multiple Choices, Learning Objective 1]

The following questions are multiple-choice questions and there is no need to justify your answers. Consider the following two systems:

System 1: When the input is $x_1(t)$, the output is

$$y_1(t) = \begin{cases} \frac{e^{jx_1(t-1)} + e^{-jx_1(t-1)}}{e^{jx_1(t-1)} - e^{-jx_1(t-1)}} & \text{if } x_1(t-1) \neq 0\\ 0 & \text{if } x_1(t-1) = 0 \end{cases}$$
(10)

System 2: When the input is $x_2[n]$, the output is

$$y_2[n] = \int_{s=-\infty}^{\max(x[n],0)} e^s ds$$
 (11)

Answer the following questions

- 1. [4%, Learning Objective 1] Is System 1 memoryless? Is System 2 memoryless?
- 2. [4%, Learning Objective 1] Is System 1 causal? Is System 2 causal?
- 3. [4%, Learning Objective 1] Is System 1 stable? Is System 2 stable?
- 4. [4%, Learning Objective 1] Is System 1 linear? Is System 2 linear?
- 5. [4%, Learning Objective 1] Is System 1 time-invariant? Is System 2 time-invariant?

Discrete-time Fourier series

$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk(2\pi/N)n} \tag{1}$$

$$a_k = \frac{1}{N} \sum_{n = \langle N \rangle} x[n] e^{-jk(2\pi/N)n} \tag{2}$$

Continuous-time Fourier series

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk(2\pi/T)t}$$
(3)

$$a_k = \frac{1}{T} \int_T x(t) e^{-jk(2\pi/T)t} dt \tag{4}$$

Continuous-time Fourier transform

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$
(5)

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt$$
(6)

Discrete-time Fourier transform

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(e^{j\omega}) e^{j\omega n} d\omega \tag{7}$$

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$
(8)

Laplace transform

$$x(t) = \frac{1}{2\pi} e^{\sigma t} \int_{-\infty}^{\infty} X(\sigma + j\omega) e^{j\omega t} d\omega$$
(9)

$$X(s) = \int_{-\infty}^{\infty} x(t)e^{-st}dt$$
(10)

Z transform

$$x[n] = r^{n} \mathcal{F}^{-1}(X(re^{j\omega})) \tag{11}$$

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$
(12)

IABLE 5.1 THOLEHILLO	Fourier Series Coefficients		
Property Section		Periodic Signal	
		$x(t)$ Periodic with period T and $y(t)$ fundamental frequency $\omega_0 = 2\pi/T$	a_k b_k
Linearity	3.5.1	Ax(t) + By(t)	$Aa_k + Bb_k$ $a_k e^{-jk\omega_0 t_0} = a_k e^{-jk(2\pi/T)t_0}$
Time Shifting	3.5.2	$x(t-t_0)$ $e^{jM\omega_0 t}x(t) = e^{jM(2\pi/T)t}x(t)$	a_{k-M}
Conjugation	3.5.6	$x^*(t)$	a_{-k}
Time Reversal	3.5.3 3.5.4	x(-t) $x(\alpha t), \alpha > 0$ (periodic with period T/α)	a_k
Periodic Convolution		$\int_{T} x(\tau) y(t-\tau) d\tau$	$Ta_k b_k$
Multiplication	3.5.5	x(t)y(t)	$\sum_{l=-\infty}^{+\infty}a_lb_{k-l}$
Differentiation		$\frac{dx(t)}{dt}$	$jk\omega_0 a_k = jk\frac{2\pi}{T}a_k$
Integration		$\int_{-\infty}^{t} x(t) dt$ (finite valued and periodic only if $a_0 = 0$)	$\left(\frac{1}{jk\omega_0}\right)a_k = \left(\frac{1}{jk(2\pi/T)}\right)a_k$ $\left(a_k = a^*\right)$
Conjugate Symmetry for Real Signals	3.5.6	x(t) real	$\begin{cases} a_k & \exists_{-k} \\ \Re e\{a_k\} = \Re e\{a_{-k}\} \\ \Im m\{a_k\} = -\Im m\{a_{-k}\} \\ a_k = a_{-k} \\ \not \propto a_k = - \not \ll a_{-k} \end{cases}$
Real and Even Signals Real and Odd Signals Even-Odd Decomposition of Real Signals	3.5.6 3.5.6	$\begin{aligned} x(t) \text{ real and even} \\ x(t) \text{ real and odd} \\ \begin{cases} x_e(t) = \mathcal{E}\upsilon\{x(t)\} & [x(t) \text{ real}] \\ x_o(t) = \mathcal{O}d\{x(t)\} & [x(t) \text{ real}] \end{cases} \end{aligned}$	a_k real and even a_k purely imaginary and odd $\operatorname{Re}\{a_k\}$ $j\operatorname{Im}\{a_k\}$
		Parseval's Relation for Periodic Signals	
		$\frac{1}{T}\int_{T} x(t) ^{2}dt = \sum_{k=-\infty}^{+\infty} a_{k} ^{2}$	

PROPERTIES OF CONTINUOUS-TIME FOURIER SERIES

three examples, we illustrate this. The last example in this section then demonstrates how properties of a signal can be used to characterize the signal in great detail.

Example 3.6

Consider the signal g(t) with a fundamental period of 4, shown in Figure 3.10 is could determine the Figure 3.10 is could determine the Fourier series representation of g(t) directly from the analysis control (2.30). Instead, when a function of g(t) directly from the analysis control (2.30). tion (3.39). Instead, we will use the relationship of g(t) to the symmetric periodic space wave r(t) in Example 3.5. Before to the wave x(t) in Example 3.5. Referring to that example, we see that, with T = 4 at $T_{1} = 1$ $T_1 = 1,$

g(t) = x(t-1) - 1/2.

Sec. 3.7 Properties of Discrete-Time Fourier Series

Thus, in general, *none* of the finite partial sums in eq. (3.52) yield the exact values of x(t), and convergence issues, such as those considered in Section 3.4, arise as we consider the problem of evaluating the limit as the number of terms approaches infinity.

3.7 PROPERTIES OF DISCRETE-TIME FOURIER SERIES

There are strong similarities between the properties of discrete-time and continuous-time Fourier series. This can be readily seen by comparing the discrete-time Fourier series properties summarized in Table 3.2 with their continuous-time counterparts in Table 3.1.

TABLE 3.2	PROPERTIES	0F	DISCRETE-TIME	FOURIER	SERIES
-----------	------------	----	---------------	---------	--------

Property	Periodic Signal	Fourier Series Coefficients	
	$x[n]$ Periodic with period N and $y[n]$ fundamental frequency $\omega_0 = 2\pi/N$	$\left. \begin{array}{c} a_k \\ b_k \end{array} \right\}$ Periodic with b_k period N	
Linearity Time Shifting Frequency Shifting Conjugation Time Reversal Time Scaling	$Ax[n] + By[n]$ $x[n - n_0]$ $e^{jM(2\pi/N)n}x[n]$ $x^{*}[n]$ $x[-n]$ $x_{(m)}[n] = \begin{cases} x[n/m], & \text{if } n \text{ is a multiple of } m \\ 0, & \text{if } n \text{ is not a multiple of } m \end{cases}$ $(\text{periodic with period } mN)$	$Aa_{k} + Bb_{k}$ $a_{k}e^{-jk(2\pi/N)n_{0}}$ a_{k-M} a_{-k}^{*} a_{-k} $\frac{1}{m}a_{k}$ (viewed as periodic) (with period mN)	
Periodic Convolution Multiplication	$\sum_{\substack{r=\langle N\rangle\\x[n]y[n]}} x[r]y[n-r]$	Na_kb_k $\sum a_lb_{k-l}$	
First Difference	x[n] - x[n-1]	$\frac{1}{(1-e^{-jk(2\pi/N)})a_k}$	
Running Sum	$\sum_{k=-\infty}^{n} x[k] \left(\begin{array}{c} \text{finite valued and periodic only} \\ \text{if } a_0 = 0 \end{array} \right)$	$\left(\frac{1}{(1-e^{-jk(2\pi/N)})}\right)a_k$	
Conjugate Symmetry for Real Signals	x[n] real	$\left\{egin{array}{l} a_k &= a_{-k}^* \ { m Re}\{a_k\} &= { m Re}\{a_{-k}\} \ { m Sm}\{a_k\} &= -{ m Sm}\{a_{-k}\} \ a_k &= a_{-k} \ { m \sphericalangle} a_k &= -{ m \sphericalangle} a_{-k} \end{array} ight.$	
Real and Even Signals Real and Odd Signals	x[n] real and even $x[n]$ real and odd	a_k real and even a_k purely imaginary and odd	
Even-Odd Decomposition of Real Signals	$\begin{cases} x_e[n] = \delta v\{x[n]\} & [x[n] real] \\ x_o[n] = \mathbb{O}d\{x[n]\} & [x[n] real] \end{cases}$	$\mathbb{R}e\{a_k\}$ $j\mathcal{G}m\{a_k\}$	
	Parseval's Relation for Periodic Signals		
	$\frac{1}{N}\sum_{n=\langle N\rangle} x[n] ^2 = \sum_{k=\langle N\rangle} a_k ^2$,	
		······································	

Chap. 3

f eqs. iodic h M = 1; = 4.

sequence in (3.106), the ns, we have

(3.107)

onclude from

if values ov o represent 221