Midterm #2 of ECE301, Section 3 (CRN 17102-004) 6:30–7:30pm, Thursday, October 11, 2018, PHYS112.

- 1. Please make sure that it is your name printed on the exam booklet. Enter your student ID number, and signature in the space provided on this page, **NOW!**
- 2. This is a closed book exam.
- 3. This exam contains multiple choice questions and work-out questions. For multiple choice questions, there is no need to justify your answers. You have one hour to complete it. The students are suggested not spending too much time on a single question, and working on those that you know how to solve.
- 4. Use the back of each page for rough work.
- 5. Neither calculators nor help sheets are allowed.

Name:

Student ID:

As a Boiler Maker pursuing academic excellence, I pledge to be honest and true in all that I do. Accountable together — We are Purdue.

Signature:

Date:

Question 1: [22%, Work-out question, Learning Objectives 1, 2, and 3]

- 1. [4%] What is the definition of *impulse response*?
- 2. [12%] Consider a DT-LTI system with the input/output relationship being

$$y[n] = \sum_{k=-\infty}^{n-2} x[k]e^{n-k-2}$$
(1)

Find the impulse response h[n].

3. [6%] Is the above system causal? Please carefully write down your reasons. A yes/no answer without reasons will not receive any point.

Hint: If you do not know the answer to the previous subquestion, you can assume h[n] = 3 - U[n] - 2U[n-2]. You will receive full points if your answer is correct. However, please indicate carefully if you are solving the alternative question.

Question 2: [15%, Work-out question, Learning Objectives 1, 2, and 3] Consider the following three signals:

$$x_1(t) = e^{-|t|} (2)$$

$$x_2(t) = e^{j3t} \tag{3}$$

1. [15%] Compute the convolution $y(t) = x_1(t) * x_2(t)$.

Question 3: [15%, Work-out question, Learning Objectives 2, 3, and 4] Consider a CT-LTI system with impulse response

$$h(t) = \begin{cases} e^{-2t} & \text{if } 0 < t\\ e^t & \text{if } t \le 0 \end{cases}$$

$$\tag{4}$$

Find the output y(t) when then input is $x(t) = e^{j3t} + e^{-j0.5\pi t}$.

Hint: There is no need to greatly simplify your answers. Some expressions of the form of $\frac{3}{3+4j}e^{j\sqrt{2}t}$ would suffice.

Question 4: [20%, Work-out question, Learning Objectives 4 and 5]

1. [12%] Consider a periodic CT signal

$$x(t) = \cos(3.5\pi t + 1) + \sin(2.5\pi t).$$
(5)

Find the Fourier series representation of x(t).

Hint: The following formula may be useful:

$$\cos(\theta) = \frac{e^{j\theta} + e^{-j\theta}}{2} \tag{6}$$

$$\sin(\theta) = \frac{e^{j\theta} - e^{-j\theta}}{2j}.$$
(7)

2. [8%] Continue from the previous question. Consider another CT signal $y(t) = \frac{d}{dt}x(t)$. Find the Fourier series representation of y(t).

Hint: If you do not know the answer to the previous question, you can assume $y(t) = \sin(0.5\pi t)$. You will receive 4 points if your answer is correct.

Question~5: [8%, Work-out question, Learning Objective 4] Consider a DT signal

$$x[n] = \begin{cases} 2\pi & \text{if } 0 \le n \le 2\\ 0 & \text{if } 3 \le n \le 49 \\ x[n] \text{ is periodic with period 50} \end{cases}$$
(8)

Find the DTFS representation of x[n].

Question 6: [20%, Multiple Choices, Learning Objective 1]

The following questions are multiple-choice questions and there is no need to justify your answers. Consider the following two systems:

System 1: When the input is $x_1(t)$, the output is

$$y_1(t) = \int_{-\infty}^{t-2} e^{-|x_1(s)| - |s|} ds$$
(9)

System 2: When the input is $x_2[n]$, the output is

$$y_2[n] = \begin{cases} \int_{s=-\infty}^{x_2[n]} e^s ds & \text{if } n \le 0\\ \int_{s=x_2[n]}^{\infty} e^{-s} ds & \text{if } n > 0 \end{cases}$$
(10)

Answer the following questions

- 1. [4%, Learning Objective 1] Is System 1 memoryless? Is System 2 memoryless?
- 2. [4%, Learning Objective 1] Is System 1 causal? Is System 2 causal?
- 3. [4%, Learning Objective 1] Is System 1 stable? Is System 2 stable?
- 4. [4%, Learning Objective 1] Is System 1 linear? Is System 2 linear?
- 5. [4%, Learning Objective 1] Is System 1 time-invariant? Is System 2 time-invariant?

Discrete-time Fourier series

$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk(2\pi/N)n} \tag{1}$$

$$a_k = \frac{1}{N} \sum_{n = \langle N \rangle} x[n] e^{-jk(2\pi/N)n}$$
⁽²⁾

Continuous-time Fourier series

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk(2\pi/T)t}$$
(3)

$$a_k = \frac{1}{T} \int_T x(t) e^{-jk(2\pi/T)t} dt \tag{4}$$

Continuous-time Fourier transform

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$
⁽⁵⁾

$$X(j\omega) = \int_{-\infty}^{\infty} x(t)e^{-j\omega t}dt$$
(6)

Discrete-time Fourier transform

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(j\omega) e^{j\omega n} d\omega$$
⁽⁷⁾

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$
(8)

Laplace transform

$$x(t) = \frac{1}{2\pi} e^{\sigma t} \int_{-\infty}^{\infty} X(\sigma + j\omega) e^{j\omega t} d\omega$$
(9)

$$X(s) = \int_{-\infty}^{\infty} x(t)e^{-st}dt$$
(10)

Z transform

$$x[n] = r^{n} \mathcal{F}^{-1}(X(re^{j\omega})) \tag{11}$$

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n}$$
(12)

	Section	Periodic Signal	Fourier Series Coefficients
Property	Section		a_k
		x(t) Periodic with period T and $y(t)$ fundamental frequency $\omega_0 = 2\pi/T$	b_k
		Ax(t) + By(t)	$Aa_k + Bb_k$
Linearity	3.5.1	(4 4)	$a_k e^{-jk\omega_0 t_0} = a_k e^{-jk(2\pi/T)t_0}$
Time Shifting	3.5.2	$x(t - t_0) e^{jM\omega_0 t} x(t) = e^{jM(2\pi/T)t} x(t)$	a_{k-M}
Frequency Shifting	3.5.6	$x^*(t)$	a^*_{-k}
Conjugation	3.5.0	r(-t)	a_{-k}
Time Reversal	3.5.5 3.5.4	x(-t) $x(\alpha t), \alpha > 0$ (periodic with period T/α)	a_k
Time Scaling	5.5.4		T a b
Periodic Convolution		$\int_{T} x(\tau) y(t-\tau) d\tau$	$Ta_k b_k$
1 chodie - Fill		51	$\sum_{n=1}^{+\infty} a b_{n-1}$
Multiplication	3.5.5	x(t)y(t)	$\sum_{l=-\infty}^{+\infty} a_l b_{k-l}$
Multiplication			277
		$\frac{dx(t)}{dt}$	$jk\omega_0 a_k = jk\frac{2\pi}{T}a_k$
Differentiation			
		$\int_{-\infty}^{t} x(t) dt$ (finite valued and periodic only if $a_0 = 0$)	$\left(\frac{1}{ik\omega_0}\right)a_k = \left(\frac{1}{ik(2\pi/T)}\right)$
Integration		$\int_{-\infty} x(t) dt$ periodic only if $a_0 = 0$	$(j\kappa\omega_0)$ $(j\kappa(2\pi/1))$
e e			$\int a_k = a_{-k}^*$
			$(\operatorname{Re}\{a_k\} = \operatorname{Re}\{a_{-k}\}$
			$g_{m}\{a_{1}\} = -g_{m}\{a_{-1}\}$
Conjugate Symmetry for	3.5.6	x(t) real	$\begin{cases} \Re \cdot \{a_k\} = \Re \cdot \{a_{-k}\} \\ \mathfrak{G}_{\mathcal{M}}\{a_k\} = -\mathfrak{G}_{\mathcal{M}}\{a_{-k}\} \\ a_k = a_{-k} \\ \mathfrak{F}_{\mathcal{A}}a_k = -\mathfrak{F}_{\mathcal{A}}a_{-k} \end{cases}$
Real Signals			$ a_k - a_{-k} $
Real Dignalo			
		x(t) real and even	a_k real and even
Real and Even Signals	3.5.6	x(t) real and odd	a_k purely imaginary and o
Real and Odd Signals	3.5.6	$(x(t) - \mathcal{E}_{\infty}(x(t)) - [x(t) real]$	$\Re e\{a_k\}$
Even-Odd Decomposition		$\begin{cases} x_e(t) = \mathcal{E}_{\mathcal{V}}\{x(t)\} & [x(t) \text{ real}] \\ x_o(t) = \mathcal{O}d\{x(t)\} & [x(t) \text{ real}] \end{cases}$	$j\mathfrak{Gm}\{a_k\}$
of Real Signals		$[x_o(t) = \bigcup \{x(t)\} \ [x(t) \ teat]$	
		Parseval's Relation for Periodic Signals	
		$\frac{1}{T}\int_{T} x(t) ^{2}dt = \sum_{k=-\infty}^{+\infty} a_{k} ^{2}$	

PROPERTIES OF CONTINUOUS-TIME FOURIER SERIES

three examples, we illustrate this. The last example in this section then demonstrates how properties of a signal can be used to characterize the signal in great detail.

Example 3.6

Consider the signal g(t) with a fundamental period of 4, shown in Figure 3.10 is could determine the Figure 3.10 is could determine the Fourier series representation of g(t) directly from the analysis control (2.30). Instead, when a function of g(t) directly from the analysis control (2.30). tion (3.39). Instead, we will use the relationship of g(t) to the symmetric periodic space wave r(t) in Example 3.5. Before to the wave x(t) in Example 3.5. Referring to that example, we see that, with T = 4 at $T_{1} = 1$ $T_1 = 1,$

g(t) = x(t-1) - 1/2.

Sec. 3.7 Properties of Discrete-Time Fourier Series

Thus, in general, *none* of the finite partial sums in eq. (3.52) yield the exact values of x(t), and convergence issues, such as those considered in Section 3.4, arise as we consider the problem of evaluating the limit as the number of terms approaches infinity.

3.7 PROPERTIES OF DISCRETE-TIME FOURIER SERIES

There are strong similarities between the properties of discrete-time and continuous-time Fourier series. This can be readily seen by comparing the discrete-time Fourier series properties summarized in Table 3.2 with their continuous-time counterparts in Table 3.1.

TABLE 3.2	PROPERTIES	0F	DISCRETE-TIME FOURIER SERIES
		U 1	

$y[n] \int \text{fundamental frequency } \omega_0 = 2\pi/N \qquad b_k \int p_k$ Linearity $Ax[n] + By[n] \qquad Aa_k + \\ x[n - n_0] \qquad a_{k}e^{-jk\ell}$ Prequency Shifting $e^{jM(2\pi/N)n}x[n] \qquad a_{k-M}$ a_{k-M} Conjugation $x^*[n] \qquad x^*[n] \qquad a_{k-M}$ $x[-n] \qquad a_{k-k}$ Time Reversal $x[-n] \qquad x[-n] \qquad a_{k-k}$ Time Scaling $x_{(m)}[n] = \begin{cases} x[n/m], \text{ if } n \text{ is a multiple of } m \\ 0, \text{ if } n \text{ is not a multiple of } m \\ (periodic with period mN) \end{cases}$ Periodic Convolution $\sum_{r=\langle N \rangle} x[r]y[n - r] \qquad Na_kb_k$ Multiplication $x[n]y[n] \qquad \sum_{r=\langle N \rangle} a_i b_i \\ (1 - e^{-ikk}) \qquad (1 - e^{-ikk}) \\ ($	Fourier Series Coefficient	
Time Shifting $x[n - n_0]$ $Ad_k + a_k e^{-jkt}$ Frequency Shifting $x[n - n_0]$ a_{km} Frequency Shifting $e^{jM(2\pi/N)n}x[n]$ a_{k-m} Conjugation $x^*[n]$ a_{k-m} Time Reversal $x[-n]$ a_{k-m} Time Scaling $x[n][n] = \begin{cases} x[n/m], & \text{if } n \text{ is a multiple of } m \\ 0, & \text{if } n \text{ is not a multiple of } m \\ 0, & \text{if } n \text{ is not a multiple of } m \\ periodic Convolution\sum_{r=\langle N \rangle} x[r]y[n-r]Na_kb_kMultiplicationx[n]y[n]\sum_{l=\langle N \rangle} a_l bFirst Differencex[n] - x[n-1](1 - e^{-1})Running Sum\sum_{k=-\infty}^n x[k] \begin{pmatrix} \text{finite valued and periodic only} \\ \text{if } a_0 = 0 \end{pmatrix}\begin{pmatrix} a_k = a \\ Re\{a_k\} \\ gm\{a_k \\ a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k = \\ < a_k$	riodic with riod N	
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of Real Signals $\begin{cases} x_e[n] = \mathcal{E}v\{x[n]\} & [x[n] real] \end{cases} \qquad $	•	
Parseval's Relation for Periodic Signals		
$\frac{1}{N}\sum_{n=\langle N\rangle} x[n] ^2 = \sum_{k=\langle N\rangle} a_k ^2$		

Chap. 3

f eqs. iodic h M = 1; = 4.

sequence in (3.106), the ns, we have

(3.107)

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if values ov o reptesent 221