

**Midterm #3 of ECE301, Section 2**  
8–9pm, Monday, November 9, 2015, ARMS 1010.

1. Please make sure that it is your name printed on the exam booklet. Enter your student ID number, and signature in the space provided on this page, **NOW!**
2. This is a closed book exam.
3. This exam contains multiple choice questions and work-out questions. For multiple choice questions, there is no need to justify your answers. You have one hour to complete it. The students are suggested not spending too much time on a single question, and working on those that you know how to solve.
4. Use the back of each page for rough work.
5. Neither calculators nor help sheets are allowed.

Name:

Student ID:

I certify that I have neither given nor received unauthorized aid on this exam.

Signature:

Date:

*Question 1:* [18%, Work-out question, Learning Objectives 1, 2, 3] Consider a discrete-time signal

$$x[n] = \begin{cases} 0.5e^{jn} & \text{if } 1 \leq n \leq 10 \\ 0.5 & \text{if } 11 \leq n \leq 20. \\ \text{is periodic with period 20} \end{cases} \quad (1)$$

Denote the Fourier series coefficient of  $x[n]$  by  $(a_k, \omega_0)$ .

1. [6%] Find the value of  $a_0$ .
2. [6%] Find the value of  $\sum_{k=0}^{19} a_k$ .
3. [6%] Find the value of  $\sum_{k=0}^{19} |a_k|^2$ .

Hint: You may need to use the following formulas: If  $|r| \neq 1$ , then we have

$$\sum_{k=1}^K ar^{k-1} = \frac{a(1-r^K)}{1-r}. \quad (2)$$

If  $|r| < 1$ , then we have

$$\sum_{k=1}^{\infty} ar^{k-1} = \frac{a}{1-r}. \quad (3)$$

If  $|r| \geq 1$ , then we have

$$\sum_{k=1}^{\infty} ar^{k-1} \text{ does not exist.} \quad (4)$$



Question 2: [23%, Work-out question, Learning Objectives 2, 3, and 4] Consider a CT-LTI system, for which the impulse response is

$$h(t) = \frac{\sin(5t) - \sin(3t)}{\pi t} \quad (5)$$

Consider a CT signal:

$$x(t) = \sum_{k=0}^{\infty} \sin(k\pi/2t). \quad (6)$$

1. [10%] Find the Fourier series  $(a_k, \frac{2\pi}{N})$  of  $x(t)$ . Specifically, (i) find the expression of  $a_k$  for any arbitrary  $k$ , (ii) find the period  $N$ , and (iii) plot  $a_k$  for the range of  $-2 \leq k \leq 2$ .
2. [13%] Let  $y(t)$  denote the output when the input is  $x(t)$ . Find the Fourier series  $(b_k, \frac{2\pi}{N})$  of  $x(t)$ . Specifically, (i) find the expression of  $b_k$ , (ii) find the period  $N$ , and (iii) plot  $b_k$  for the range of  $-3 \leq k \leq 3$ .

Hint 1: If you do not know the answer to this sub-question, you can find the Fourier transform  $H(j\omega)$  of  $h(t)$ . You will receive 10 points if your answer is correct.

Hint 2: The following values may be of use.

$$\frac{\pi}{2} \approx 1.571 \quad (7)$$

$$\pi \approx 3.142 \quad (8)$$

$$\frac{3\pi}{2} \approx 4.712 \quad (9)$$

$$2\pi \approx 6.283 \quad (10)$$

$$\frac{5\pi}{2} \approx 7.854 \quad (11)$$



Question 3: [25%, Work-out question, Learning Objectives 4, and 5]

1. [9%] Consider a continuous time signal  $x(t)$  with its Fourier transform  $X(j\omega)$  being

$$X(j\omega) = (1 + j)\delta(\omega - 3) + (1 - j)\delta(\omega + 3) + \delta(\omega) \quad (12)$$

Find the expression of  $x(t)$ .

2. [3%] Is  $x(t)$  a real-valued signal? Use one quick sentence to justify your answer.
3. [1%] Consider a continuous time signal  $y(t)$ :

$$y(t) = \sum_{k=-2}^2 \delta(t - 2\pi k) \quad (13)$$

Is  $y(t)$  periodic?

4. [10%] Find the expression of its Fourier transform  $Y(j\omega)$ .
5. [2%] Is  $Y(j\omega)$  periodic? If so, what is its period? If not, use a quick sentence to justify why  $Y(j\omega)$  is not periodic.



*Question 4:* [15%, Work-out question, Learning Objectives 4, 5, and 6] Consider the following signal

$$x(t) = \begin{cases} t & \text{if } 0 \leq t < 2 \\ 4 - t & \text{if } 2 \leq t < 4 \\ 0 & \text{otherwise.} \end{cases} \quad (14)$$

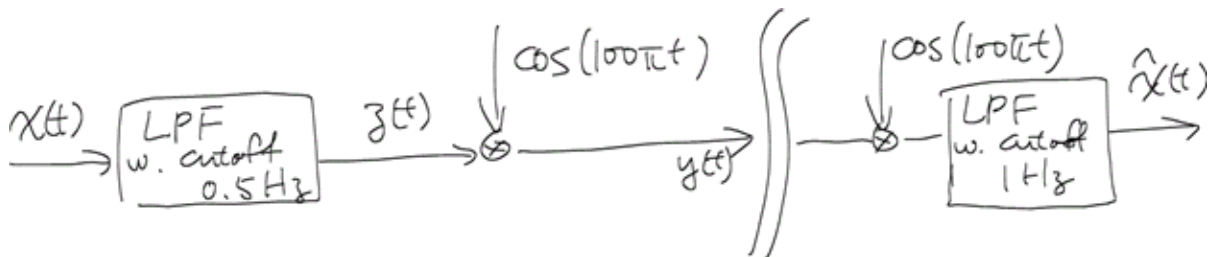
1. [2%] Plot  $x(t)$  for the range of  $-7 < t < 7$ .
2. [13%] Find the Fourier transform  $X(j\omega)$  of  $x(t)$ .





Question 5: [19%, Multiple Choices] Consider the following AM transmission system. For the transmitter side, for whatever signal  $x(t)$ , we first pass it through a low-pass filter with cut-off frequency 0.5Hz. Then we multiply it with  $\cos(100\pi t)$ .

For the receiver side, we first multiply the received signal by  $\cos(100\pi t)$  and then pass it through a low-pass filter with cut-off frequency 1Hz. The overall system flow chart is as follows.



Answer the following questions:

- [3%] What is the expression of the impulse response of a low-pass filter with cutoff frequency 0.5Hz.
- [3%] If the original signal  $x(t)$  has  $X(j\omega) = \mathcal{U}(\omega + 5) - \mathcal{U}(\omega - 5)$ . What is the Fourier transform  $Z(j\omega)$  of  $z(t)$ ?

Hint: If you do not know the answer to the previous subquestion, you can assume that a low-pass filter with cutoff frequency 0.5Hz has  $H(j\omega)$  being  $\mathcal{U}(\omega+3) - \mathcal{U}(\omega-3)$ . You will get full credit if your  $Z(j\omega)$  is correct.

- [9%] Plot the signal  $y(t)$  for the range of  $-4 < t < 4$ .

Hint 1: If you do not know how to plot  $y(t)$ , you can simply plot  $z(t)$  instead. You will receive 5 points if your plot of  $z(t)$  is correct.

Hint 2: If you do not know the answers of the previous questions, you can answer the following question instead: For a frequency-domain signal  $X_1(j\omega) = \mathcal{U}(\omega + 3) - \mathcal{U}(\omega - 3)$ , plot assume  $x_1(t)$  for the range of  $-\pi < t < \pi$ . You will receive 5 points if your answer is correct.

- [4%, Learning Objective 1] If you play the signals  $z(t)$  and  $\hat{x}(t)$  using the speakers of your computer. Will these two signals sound identical? Use a couple of sentences to justify your answer.



Discrete-time Fourier series

$$x[n] = \sum_{k=\langle N \rangle} a_k e^{jk(2\pi/N)n} \quad (1)$$

$$a_k = \frac{1}{N} \sum_{n=\langle N \rangle} x[n] e^{-jk(2\pi/N)n} \quad (2)$$

Continuous-time Fourier series

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk(2\pi/T)t} \quad (3)$$

$$a_k = \frac{1}{T} \int_T x(t) e^{-jk(2\pi/T)t} dt \quad (4)$$

Continuous-time Fourier transform

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega \quad (5)$$

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt \quad (6)$$

Discrete-time Fourier transform

$$x[n] = \frac{1}{2\pi} \int_{2\pi} X(j\omega) e^{j\omega n} d\omega \quad (7)$$

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x[n] e^{-j\omega n} \quad (8)$$

Laplace transform

$$x(t) = \frac{1}{2\pi} e^{\sigma t} \int_{-\infty}^{\infty} X(\sigma + j\omega) e^{j\omega t} d\omega \quad (9)$$

$$X(s) = \int_{-\infty}^{\infty} x(t) e^{-st} dt \quad (10)$$

Z transform

$$x[n] = r^n \mathcal{F}^{-1}(X(re^{j\omega})) \quad (11)$$

$$X(z) = \sum_{n=-\infty}^{\infty} x[n] z^{-n} \quad (12)$$

TABLE 3.1 PROPERTIES OF CONTINUOUS-TIME FOURIER SERIES

| Property  | Section | Periodic Signal  | Fourier Series Coefficients  |
|---|---------|--|--|
|   |         | $x(t)$ } Periodic with period $T$ and<br>$y(t)$ } fundamental frequency $\omega_0 = 2\pi/T$  | $a_k$<br>$b_k$   |
| Linearity   | 3.5.1   | $Ax(t) + By(t)$  | $Aa_k + Bb_k$  |
| Time Shifting   | 3.5.2   | $x(t - t_0)$   | $a_k e^{-jk\omega_0 t_0} = a_k e^{-jk(2\pi/T)t_0}$   |
| Frequency Shifting  |         | $e^{jM\omega_0 t} x(t) = e^{jM(2\pi/T)t} x(t)$   | $a_{k-M}$  |
| Conjugation   | 3.5.6   | $x^*(t)$   | $a_{-k}^*$   |
| Time Reversal   | 3.5.3   | $x(-t)$  | $a_{-k}$   |
| Time Scaling  | 3.5.4   | $x(\alpha t), \alpha > 0$ (periodic with period $T/\alpha$ )   | $a_k$  |
| Periodic Convolution  |         | $\int_T x(\tau)y(t - \tau)d\tau$   | $T a_k b_k$  |
| Multiplication  | 3.5.5   | $x(t)y(t)$   | $\sum_{l=-\infty}^{+\infty} a_l b_{k-l}$   |
| Differentiation   |         | $\frac{dx(t)}{dt}$   | $jk\omega_0 a_k = jk \frac{2\pi}{T} a_k$   |
| Integration   |         | $\int_{-\infty}^t x(t) dt$ (finite valued and periodic only if $a_0 = 0$ )   | $\left(\frac{1}{jk\omega_0}\right) a_k = \left(\frac{1}{jk(2\pi/T)}\right) a_k$  |
| Conjugate Symmetry for Real Signals                                   | 3.5.6   | $x(t)$ real  | $\begin{cases} a_k = a_{-k}^* \\ \Re\{a_k\} = \Re\{a_{-k}\} \\ \Im\{a_k\} = -\Im\{a_{-k}\} \\  a_k  =  a_{-k}  \\ \angle a_k = -\angle a_{-k} \end{cases}$ |
| Real and Even Signals   | 3.5.6   | $x(t)$ real and even   | $a_k$ real and even  |
| Real and Odd Signals  | 3.5.6   | $x(t)$ real and odd  | $a_k$ purely imaginary and odd   |
| Even-Odd Decomposition of Real Signals                                |         | $\begin{cases} x_e(t) = \mathcal{E}\{x(t)\} & [x(t) \text{ real}] \\ x_o(t) = \mathcal{O}\{x(t)\} & [x(t) \text{ real}] \end{cases}$ | $\begin{cases} \Re\{a_k\} \\ \Im\{a_k\} \end{cases}$   |
| Parseval's Relation for Periodic Signals                              |         |  |  |
| $\frac{1}{T} \int_T  x(t) ^2 dt = \sum_{k=-\infty}^{+\infty}  a_k ^2$ |         |  |  |

three examples, we illustrate this. The last example in this section then demonstrates how properties of a signal can be used to characterize the signal in great detail.

### Example 3.6

Consider the signal  $g(t)$  with a fundamental period of 4, shown in Figure 3.10. We could determine the Fourier series representation of  $g(t)$  directly from the analysis equation (3.39). Instead, we will use the relationship of  $g(t)$  to the symmetric periodic square wave  $x(t)$  in Example 3.5. Referring to that example, we see that, with  $T = 4$  and  $T_1 = 1$ ,

$$g(t) = x(t - 1) - 1/2. \quad (3.40)$$

Thus, in general, *none* of the finite partial sums in eq. (3.52) yield the exact values of  $x(t)$ , and convergence issues, such as those considered in Section 3.4, arise as we consider the problem of evaluating the limit as the number of terms approaches infinity.

### 3.7 PROPERTIES OF DISCRETE-TIME FOURIER SERIES

There are strong similarities between the properties of discrete-time and continuous-time Fourier series. This can be readily seen by comparing the discrete-time Fourier series properties summarized in Table 3.2 with their continuous-time counterparts in Table 3.1.

**TABLE 3.2** PROPERTIES OF DISCRETE-TIME FOURIER SERIES

| Property                               | Periodic Signal  | Fourier Series Coefficients  |
|--|--|--|
|  | $x[n]$ } Periodic with period $N$ and<br>$y[n]$ } fundamental frequency $\omega_0 = 2\pi/N$  | $a_k$ } Periodic with<br>$b_k$ } period $N$  |
| Linearity                              | $Ax[n] + By[n]$  | $Aa_k + Bb_k$  |
| Time Shifting                          | $x[n - n_0]$   | $a_k e^{-jk(2\pi/N)n_0}$   |
| Frequency Shifting                     | $e^{jM(2\pi/N)n} x[n]$   | $a_{k-M}$  |
| Conjugation                            | $x^*[n]$   | $a_{-k}^*$   |
| Time Reversal                          | $x[-n]$  | $a_{-k}$   |
| Time Scaling                           | $x_{(m)}[n] = \begin{cases} x[n/m], & \text{if } n \text{ is a multiple of } m \\ 0, & \text{if } n \text{ is not a multiple of } m \end{cases}$<br>(periodic with period $mN$ ) | $\frac{1}{m} a_k$ (viewed as periodic)<br>(with period $mN$ )  |
| Periodic Convolution                   | $\sum_{r=(N)} x[r]y[n-r]$  | $Na_k b_k$   |
| Multiplication                         | $x[n]y[n]$   | $\sum_{l=(N)} a_l b_{k-l}$   |
| First Difference                       | $x[n] - x[n-1]$  | $(1 - e^{-jk(2\pi/N)})a_k$   |
| Running Sum                            | $\sum_{k=-\infty}^n x[k]$ (finite valued and periodic only)<br>(if $a_0 = 0$ )   | $\left(\frac{1}{1 - e^{-jk(2\pi/N)}}\right)a_k$  |
| Conjugate Symmetry for Real Signals    | $x[n]$ real  | $\begin{cases} a_k = a_{-k}^* \\ \Re\{a_k\} = \Re\{a_{-k}\} \\ \Im\{a_k\} = -\Im\{a_{-k}\} \\  a_k  =  a_{-k}  \\ \angle a_k = -\angle a_{-k} \end{cases}$ |
| Real and Even Signals                  | $x[n]$ real and even   | $a_k$ real and even  |
| Real and Odd Signals                   | $x[n]$ real and odd  | $a_k$ purely imaginary and odd   |
| Even-Odd Decomposition of Real Signals | $\begin{cases} x_e[n] = \mathcal{E}\{x[n]\} & [x[n] \text{ real}] \\ x_o[n] = \mathcal{O}\{x[n]\} & [x[n] \text{ real}] \end{cases}$   | $\begin{cases} \Re\{a_k\} \\ j\Im\{a_k\} \end{cases}$  |

Parseval's Relation for Periodic Signals

$$\frac{1}{N} \sum_{n=(N)} |x[n]|^2 = \sum_{k=(N)} |a_k|^2$$

## 4.6 TABLES OF FOURIER PROPERTIES AND OF BASIC FOURIER TRANSFORM PAIRS

In the preceding sections and in the problems at the end of the chapter, we have considered some of the important properties of the Fourier transform. These are summarized in Table 4.1, in which we have also indicated the section of this chapter in which each property has been discussed.

In Table 4.2, we have assembled a list of many of the basic and important Fourier transform pairs. We will encounter many of these repeatedly as we apply the tools of

**TABLE 4.1** PROPERTIES OF THE FOURIER TRANSFORM

| Section | Property                                  | Aperiodic signal  | Fourier transform  |
|---------|---|---|--|
|         |   | $x(t)$<br>$y(t)$  | $X(j\omega)$<br>$Y(j\omega)$   |
| -----   |   |   |  |
| 4.3.1   | Linearity                                 | $ax(t) + by(t)$   | $aX(j\omega) + bY(j\omega)$  |
| 4.3.2   | Time Shifting                             | $x(t - t_0)$  | $e^{-j\omega t_0} X(j\omega)$  |
| 4.3.6   | Frequency Shifting                        | $e^{j\omega_0 t} x(t)$  | $X(j(\omega - \omega_0))$  |
| 4.3.3   | Conjugation                               | $x^*(t)$  | $X^*(-j\omega)$  |
| 4.3.5   | Time Reversal                             | $x(-t)$   | $X(-j\omega)$  |
| 4.3.5   | Time and Frequency Scaling                | $x(at)$   | $\frac{1}{ a } X\left(\frac{j\omega}{a}\right)$  |
| 4.4     | Convolution                               | $x(t) * y(t)$   | $X(j\omega)Y(j\omega)$   |
| 4.5     | Multiplication                            | $x(t)y(t)$  | $\frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\theta)Y(j(\omega - \theta))d\theta$   |
| 4.3.4   | Differentiation in Time                   | $\frac{d}{dt}x(t)$  | $j\omega X(j\omega)$   |
| 4.3.4   | Integration                               | $\int_{-\infty}^t x(t)dt$   | $\frac{1}{j\omega} X(j\omega) + \pi X(0)\delta(\omega)$  |
| 4.3.6   | Differentiation in Frequency              | $tx(t)$   | $j \frac{d}{d\omega} X(j\omega)$   |
| 4.3.3   | Conjugate Symmetry for Real Signals       | $x(t)$ real   | $\begin{cases} X(j\omega) = X^*(-j\omega) \\ \Re\{X(j\omega)\} = \Re\{X(-j\omega)\} \\ \Im\{X(j\omega)\} = -\Im\{X(-j\omega)\} \\  X(j\omega)  =  X(-j\omega)  \\ \angle X(j\omega) = -\angle X(-j\omega) \end{cases}$ |
| 4.3.3   | Symmetry for Real and Even Signals        | $x(t)$ real and even  | $X(j\omega)$ real and even   |
| 4.3.3   | Symmetry for Real and Odd Signals         | $x(t)$ real and odd   | $X(j\omega)$ purely imaginary and odd  |
| 4.3.3   | Even-Odd Decomposition for Real Signals   | $x_e(t) = \mathcal{E}\{x(t)\}$ [ $x(t)$ real]<br>$x_o(t) = \mathcal{O}\{x(t)\}$ [ $x(t)$ real]          | $\Re\{X(j\omega)\}$<br>$j\Im\{X(j\omega)\}$  |
| -----   |   |   |  |
| 4.3.7   | Parseval's Relation for Aperiodic Signals |   |  |
|         |   | $\int_{-\infty}^{+\infty}  x(t) ^2 dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty}  X(j\omega) ^2 d\omega$ |  |

TABLE 4.2 BASIC FOURIER TRANSFORM PAIRS

| Signal   | Fourier transform  | Fourier series coefficients<br>(if periodic)   |
|--|--|--|
| $\sum_{k=-\infty}^{+\infty} a_k e^{jk\omega_0 t}$                                    | $2\pi \sum_{k=-\infty}^{+\infty} a_k \delta(\omega - k\omega_0)$                         | $a_k$  |
| $e^{j\omega_0 t}$  | $2\pi \delta(\omega - \omega_0)$   | $a_1 = 1$<br>$a_k = 0$ , otherwise   |
| $\cos \omega_0 t$  | $\pi[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$                             | $a_1 = a_{-1} = \frac{1}{2}$<br>$a_k = 0$ , otherwise  |
| $\sin \omega_0 t$  | $\frac{\pi}{j}[\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$                   | $a_1 = -a_{-1} = \frac{1}{2j}$<br>$a_k = 0$ , otherwise  |
| $x(t) = 1$   | $2\pi \delta(\omega)$  | $a_0 = 1, a_k = 0, k \neq 0$<br>(this is the Fourier series representation for any choice of $T > 0$ )                 |
| Periodic square wave   |  |  |
| $x(t) = \begin{cases} 1, &  t  < T_1 \\ 0, & T_1 <  t  \leq \frac{T}{2} \end{cases}$ | $\sum_{k=-\infty}^{+\infty} \frac{2 \sin k\omega_0 T_1}{k} \delta(\omega - k\omega_0)$   | $\frac{\omega_0 T_1}{\pi} \operatorname{sinc}\left(\frac{k\omega_0 T_1}{\pi}\right) = \frac{\sin k\omega_0 T_1}{k\pi}$ |
| and<br>$x(t + T) = x(t)$   |  |  |
| $\sum_{n=-\infty}^{+\infty} \delta(t - nT)$  | $\frac{2\pi}{T} \sum_{k=-\infty}^{+\infty} \delta\left(\omega - \frac{2\pi k}{T}\right)$ | $a_k = \frac{1}{T}$ for all $k$  |
| $x(t) \begin{cases} 1, &  t  < T_1 \\ 0, &  t  > T_1 \end{cases}$                    | $\frac{2 \sin \omega T_1}{\omega}$   | —  |
| $\frac{\sin Wt}{\pi t}$  | $X(j\omega) = \begin{cases} 1, &  \omega  < W \\ 0, &  \omega  > W \end{cases}$          | —  |
| $\delta(t)$  | 1  | —  |
| $u(t)$   | $\frac{1}{j\omega} + \pi \delta(\omega)$   | —  |
| $\delta(t - t_0)$  | $e^{-j\omega t_0}$   | —  |
| $e^{-at} u(t), \operatorname{Re}\{a\} > 0$   | $\frac{1}{a + j\omega}$  | —  |
| $t e^{-at} u(t), \operatorname{Re}\{a\} > 0$   | $\frac{1}{(a + j\omega)^2}$  | —  |
| $\frac{t^{n-1}}{(n-1)!} e^{-at} u(t), \operatorname{Re}\{a\} > 0$                    | $\frac{1}{(a + j\omega)^n}$  | —  |

transform Chap. 4

FORM PAIRS

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which each prop-

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apply the tools of

transform

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(0) $\delta(\omega)$

( $j\omega$ )

$\operatorname{Re}\{X(-j\omega)\}$

$-\operatorname{Im}\{X(-j\omega)\}$

( $j\omega$ )

$X(-j\omega)$

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imaginary and odd