## Midterm \#3 of ECE301, Prof. Wang's section

8-9pm Thursday, November 17, 2011, EE 170,

1. Please make sure that it is your name printed on the exam booklet. Enter your student ID number, e-mail address, and signature in the space provided on this page, NOW!
2. This is a closed book exam.
3. This exam contains multiple choice questions and work-out questions. For multiple choice questions, there is no need to justify your answers. You have one hour to complete it. The students are suggested not spending too much time on a single question, and working on those that you know how to solve.
4. There are 16 pages in the exam booklet. Use the back of each page for rough work.
5. Neither calculators nor help sheets are allowed.

> Name:

## Student ID:

> E-mail:

Signature:

Question 1: [15\%, Work-out question, Outcomes 3, 4, and 5]

1. $[15 \%]$ A discrete-time signal $x[n]$ is described as follows

$$
x[n]= \begin{cases}2 & \text { if } n=0  \tag{1}\\ 1 & \text { if } n=1 \text { or }-1 \\ 0 & \text { if } n=2\end{cases}
$$

and $x[n]$ is periodic with period 4 .
Find the Fourier transform $X\left(e^{j \omega}\right)$ and plot it for the range of $\omega=-2 \pi$ to $2 \pi$.

Question 2: [25\%, Work-out question, Outcomes 3, 4, and 5]

1. [6\%] $x(t)=\frac{\sin (2 t)}{t}$, plot $x(t)$ for the range of $t=-2 \pi$ to $2 \pi$. Carefully mark the height of the main lobe and the locations it intersects the horizontal axis.
2. [4\%] Find the Fourier transform $X(j \omega)$ and plot it for the range of $\omega=-12$ to 12 .
3. [15\%] $y(t)=x(t) \cdot \sin (3 t) \cdot \sin (6 t)$. Find the Fourier transform $Y(j \omega)$ and plot it for the range of $\omega=-12$ to 12 . If you do not know the answer to the previous question, you can write $Y(j \omega)$ in terms of $X(j \omega)$. You will get 12 points if your answer is correct.

Question 3: [32\%, Work-out question, Outcomes 4 and 5]

1. $[10 \%]$ Compute the Fourier transform $X(j \omega)$ of $x(t)=\frac{(\sin (3 \pi(t-0.5)))^{2}}{(\pi(t-0.5))^{2}}$ and write down its expression.
2. [6\%] Is $X(j \omega)$ periodic? Compute the value of $\int_{-\infty}^{\infty} X(j \omega) d \omega$.
3. $[10 \%]$ Compute the Fourier transform $X\left(e^{j \omega}\right)$ of $x[n]=n e^{-n} \mathcal{U}[n]$.
4. [6\%] Is $X\left(e^{j \omega}\right)$ periodic? Compute the value of $\sum_{n=1}^{\infty} n e^{-n}$. (Hint: You need to use the answer from the previous question.)

Question 4: [28\%, Work-out question, Outcomes 3, 4, and 5] Consider an AM system, which sends the input signal $x(t)$ over a cos carrier of frequency $\frac{1}{\pi} \approx 0.637 \mathrm{~Hz}$.

More specifically, we denote the input signal as $x(t)$ and use $y(t)$ to denote the AM modulated signal, which will be sent out by the AM transmitter.

1. [4\%] What is the value of the carrier frequency $\omega_{c}$ with the unit being ( $\mathrm{rad} / \mathrm{sec}$ ) ? Write down the input/output relationship (equation) between $x(t)$ and $y(t)$.
2. [12\%] Consider the receiver end. To demodulate the original signal from $y(t)$, the receiver first construct $w(t)=2 \cdot y(t) \cdot \cos \left(\omega_{c} t\right)$ and then passes $w(t)$ through a low-pass filter with cutoff frequency $\omega_{\text {cutoff }}$.
Suppose we also know that $x(t)=\cos (t)$, plot the Fourier transforms $X(j \omega), Y(j \omega)$, and $W(j \omega)$ for the range of $\omega=-6$ to 6 .
3. [12\%] Suppose that Prof. Wang forgot the expression of the impulse response of a low-pass filter, and decided to pass $w(t)$ through an LTI system with frequency response

$$
H(j \omega)= \begin{cases}1+0.5 \omega & \text { if }-2<\omega \leq 0  \tag{2}\\ 1-0.5 \omega & \text { if } 0<\omega \leq 2 \\ 0 & \text { otherwise }\end{cases}
$$

Let $z(t)$ denote the output after passing $w(t)$ through this system. Find the expression of $z(t)$.
If you do not know the answer of Q4.2, you can assume that $w(t)=\sum_{n=-\infty}^{\infty} 2^{-|n|} e^{j n t}$ and solve this question. You will still get 12 points if your answer is correct.

Discrete-time Fourier series

$$
\begin{align*}
x[n] & =\sum_{k=\langle N\rangle} a_{k} e^{j k(2 \pi / N) n}  \tag{1}\\
a_{k} & =\frac{1}{N} \sum_{n=\langle N\rangle} x[n] e^{-j k(2 \pi / N) n} \tag{2}
\end{align*}
$$

Continuous-time Fourier series

$$
\begin{align*}
x(t) & =\sum_{k=-\infty}^{\infty} a_{k} e^{j k(2 \pi / T) t}  \tag{3}\\
a_{k} & =\frac{1}{T} \int_{T} x(t) e^{-j k(2 \pi / T) t} d t \tag{4}
\end{align*}
$$

Continuous-time Fourier transform

$$
\begin{align*}
x(t) & =\frac{1}{2 \pi} \int_{-\infty}^{\infty} X(j \omega) e^{j \omega t} d \omega  \tag{5}\\
X(j \omega) & =\int_{-\infty}^{\infty} x(t) e^{-j \omega t} d t \tag{6}
\end{align*}
$$

Discrete-time Fourier transform

$$
\begin{align*}
x[n] & =\frac{1}{2 \pi} \int_{2 \pi} X(j \omega) e^{j \omega n} d \omega  \tag{7}\\
X\left(e^{j \omega}\right) & =\sum_{n=-\infty}^{\infty} x[n] e^{-j \omega n} \tag{8}
\end{align*}
$$

Laplace transform

$$
\begin{align*}
x(t) & =\frac{1}{2 \pi} e^{\sigma t} \int_{-\infty}^{\infty} X(\sigma+j \omega) e^{j \omega t} d \omega  \tag{9}\\
X(s) & =\int_{-\infty}^{\infty} x(t) e^{-s t} d t \tag{10}
\end{align*}
$$

Z transform

$$
\begin{align*}
x[n] & =r^{n} \mathcal{F}^{-1}\left(X\left(r e^{j \omega}\right)\right)  \tag{11}\\
X(z) & =\sum_{n=-\infty}^{\infty} x[n] z^{-n} \tag{12}
\end{align*}
$$

| Property | Section | Periodic Signal | Fourier Series Coefficients |
| :---: | :---: | :---: | :---: |
|  |  |  |  |
|  |  | $x(t)\}$ Periodic with period T and | $a_{k}$ |
|  |  | $y(t)\}$ fundamental frequency $\omega_{0}=2 \pi / T$ |  |
| Linearity <br> Time Shifting <br> Frequency Shifting <br> Conjugation <br> Time Reversal <br> Time Scaling |  | $\begin{aligned} & A x(t)+B y(t) \\ & x\left(t-t_{0}\right) \\ & e^{j M \omega_{0} t} x(t)=e^{j M(2 \pi / T) t} x(t) \\ & x^{*}(t) \\ & x(-t) \\ & x(\alpha t), \alpha>0(\text { periodic with period } T / \alpha) \end{aligned}$ | $A a_{k}+B b_{k}$ |
|  | 3.5.1 |  | $a_{k} e^{-j k \omega_{0} t_{0}}=a_{k} e^{-j k(2 \pi / T)_{0}}$ |
|  | 3.5.2 |  | $a_{k-M}$ |
|  |  |  | $a_{-k}^{*}$ |
|  | 3.5.6 |  | $a_{-k}$ |
|  | 3.5.5.4 |  | $a_{k}$ |
| Periodic Convolution | 3.5 .5 | $\int_{T} x(\tau) y(t-\tau) d \tau$ | $T a_{k} b_{k}$ |
|  |  | $x(t) y(t)$ | $\sum_{l=-\infty}^{+\infty} a_{l} b_{k-l}$ |
|  |  | $\underline{d x(t)}$ | $j k \omega_{0} a_{k}=j k \frac{2 \pi}{T} a_{k}$ |
| Differentiation |  | $\int^{t} x(t) d t \stackrel{(\text { finite valued and }}{\text { nerindic only if } \left.a_{0}=0\right)}$ | $\left(\frac{1}{j k \omega_{0}}\right) a_{k}=\left(\frac{1}{j k(2 \pi / T)}\right) a_{2}$ |
| Conjugate Symmetry for Real Signals | 3.5 .6 | $x(t)$ real | $\left\{\begin{array}{l} a_{k}=a_{-k}^{*} \\ \mathcal{Q e}_{\mathcal{L}}\left\{a_{k}\right\}=\mathcal{R e}_{\mathscr{L}}\left\{a_{-k}\right\} \\ \mathfrak{g}_{n}\left\{a_{k}\right\}=-\mathfrak{S n}_{n}\left\{a_{-k}\right\} \\ \left\|a_{k}\right\|=\left\|a_{-k}\right\| \\ \Varangle a_{k}=-\Varangle a_{-k} \end{array}\right.$ |
| Real and Even Signals Real and Odd Signals Even-Odd Decomposition of Real Signals | $\begin{aligned} & 3.5 .6 \\ & 3.5 .6 \end{aligned}$ | $x(t)$ real and even <br> $x(t)$ real and odd $\begin{cases}x_{o}(t)=\mathcal{E}_{v}\{x(t)\} & {[x(t) \text { real }]} \\ x_{o}(t)=\mathcal{O} d\{x(t)\} & {[x(t) \text { real }]}\end{cases}$ | $a_{k}$ real and even <br> $a_{k}$ purely imaginary and dd <br> $\mathfrak{R e}\left\{a_{k}\right\}$ <br> $j \mathfrak{g}_{n}\left\{a_{k}\right\}$ |

Parseval's Relation for Periodic Signals

$$
\frac{1}{T} \int_{T}|x(t)|^{2} d t=\sum_{k=-\infty}^{+\infty}\left|a_{k}\right|^{2}
$$

three examples, we illustrate this. The last example in this section then demonstratestir properties of a signal can be used to characterize the signal in great detail.

## Example 3.6

Consider the signal $g(t)$ with a fundamental period of 4 , shown in Figure 3.10 . could determine the Fourier series representation of $g(t)$ directly from the analysiser tion (3.39). Instead, we will use the relationship of $g(t)$ to the symmetric periodic $4=$ wave $x(t)$ in Example 3.5. Referring to that example, we see that, with $T=t=$ $T_{1}=1$,

$$
g(t)=x(t-1)-1 / 2
$$

Thus, in general, none of the finite partial sums in eq. (3.52) yield the exact values of $x(t)$, and convergence issues, such as those considered in Section 3.4, arise as we consider the problem of evaluating the limit as the number of terms approaches infinity.

### 3.7 PROPERTIES OF DISCRETE-TIME FOURIER SERIES

There are strong similarities between the properties of discrete-time and continuous-time Fourier series. This can be readily seen by comparing the discrete-time Fourier series properties summarized in Table 3.2 with their continuous-time counterparts in Table 3.1.

TABLE 3.2 PROPERTIES OF DISCRETE-TIME FOURIER SERIES

| Property | Periodic Signal | Fourier Series Coefficients |
| :---: | :---: | :---: |
|  | $\left.\begin{array}{l} x[n] \\ y[n] \end{array}\right\} \begin{aligned} & \text { Periodic with period } N \text { and } \\ & \text { fundamental frequency } \omega_{0}=2 \pi / N \end{aligned}$ | $\left.\begin{array}{l} a_{k} \\ b_{k} \end{array}\right\} \begin{aligned} & \text { Periodic with } \\ & \text { period } N \end{aligned}$ |
| Linearity <br> Time Shifting Frequency Shifting Conjugation Time Reversal | $\begin{aligned} & A x[n]+B y[n] \\ & x\left[n-n_{0}\right] \\ & e^{j M(2 \pi / N) n} x[n] \\ & x^{*}[n] \\ & x[-n] \end{aligned}$ | $\begin{aligned} & A a_{k}+B b_{k} \\ & a_{k} e^{-j k(2 \pi N) n_{0}} \\ & a_{k-M} \\ & a_{-k}^{*} \\ & a_{-k} \end{aligned}$ |
| Time Scaling | $x_{(m)}[n]= \begin{cases}x[n / m], & \text { if } n \text { is a multiple of } m \\ 0, & \text { if } n \text { is not a multiple of } m\end{cases}$ (periodic with period $m N$ ) | $\frac{1}{m} a_{k}\binom{$ viewed as periodic }{ with period $m N}$ |
| Periodic Convolution | $\sum_{r=(N)} x[r] y[n-r]$ | $N a_{k} b_{k}$ |
| Multiplication | $x[n] y[n]$ | $\sum_{l=\{N\rangle} a_{l} b_{k-l}$ |
| First Difference | $x[n]-x[n-1]$ | $\left(1-e^{-j k(2 \pi / N)}\right) a_{k}$ |
| Running Sum <br> Conjugate Symmetry for Real Signals | $\sum_{k=-\infty}^{n} x[k]\binom{\text { finite valued and periodic only }}{\text { if } a_{0}=0}$ | $\begin{aligned} & \left(\frac{1}{\left(1-e^{-j k(2 \pi / N)}\right)}\right) a_{k} \\ & \left\{\begin{array}{l} a_{k}=a_{-k}^{*} \\ \mathcal{P}_{e}\left\{a_{k}\right\}=\mathcal{R} e\left\{a_{-k}\right\} \end{array}\right. \end{aligned}$ |
|  | $x[n]$ real | $\left\{\begin{array}{l} \mathscr{S}_{n}\left\{a_{k}\right\}=\left\{a_{k}\right\}=-\mathfrak{I n}_{n}\left\{a_{-k}\right\} \\ \left\|a_{k}\right\|=\left\|a_{-k}\right\| \\ \Varangle a_{k}=-\Varangle a_{-k} \end{array}\right.$ |
| Real and Even Signals <br> Real and Odd Signals | $x[n]$ real and even <br> $x[n]$ real and odd | $a_{k}$ real and even <br> $a_{k}$ purely imaginary and odd |
| en-Odd Decomposition <br> of Real Signals | $\begin{cases}x_{e}[n]=\mathcal{E}_{\ell}\{x[n]\} & {[\mathrm{x}[\mathrm{n}] \text { real }]} \\ x_{o}[n]=0 d\{x[n]\} & {[\mathrm{x}[\mathrm{n}] \text { real }]}\end{cases}$ | $\begin{aligned} & \mathcal{R e}_{e}\left\{a_{k}\right\} \\ & j \mathscr{S}_{m}\left\{a_{k}\right\} \end{aligned}$ |
|  | Parseval's Relation for Periodic Signals $\frac{1}{N} \sum_{n=\{N\rangle}\|x[n]\|^{2}=\sum_{k=\{N\rangle}\left\|a_{k}\right\|^{2}$ |  |

### 4.6 TABLES OF FOURIER PROPERTIES AND OF BASIC FOURIER TRANSFORM PAIRS

In the preceding sections and in the problems at the end of the chapter, we have considered some of the important properties of the Fourier transform. These are summarized in Table 4.1, in which we have also indicated the section of this chapter in which each property has been discussed.

In Table 4.2, we have assembled a list of many of the basic and important Fourier transform pairs. We will encounter many of these repeatedly as we apply the tools of

TABLE 4.1 PROPERTIES OF THE FOURIER TRANSFORM


Parseval's Relation for Aperiodic Signals

$$
\int_{-\infty}^{+\infty}|x(t)|^{2} d t=\frac{1}{2 \pi} \int_{-\infty}^{+\infty}|X(j \omega)|^{2} d \omega
$$

## FORM PAIRS

, we have consid. re summarized in which each prop. important Fourier upply the tools of
transform
; $\omega$ )
, $-\theta) d \theta$
$\cdot(0) \delta(\omega)$

## $-j \omega)$

$\mathcal{P}_{\mathcal{e}}\{X(-j \omega)\}$
$-\mathscr{S}_{n}\{X(-j \omega)\}$
$-j \omega) \mid$
$\lceil X(-j \omega)$
ven
tginary and odd

TABLE 4.2 BASIC FOURIER TRANSFORM PAIRS

| Signal | Fourier transform | Fourier series coefficients (if periodic) |
| :---: | :---: | :---: |
| $\sum_{k=-\infty}^{+\infty} a_{k} e^{j k \omega_{0 j} t}$ | $2 \pi \sum_{k=-\infty}^{+\infty} a_{k} \delta\left(\omega-k \omega_{0}\right)$ | $a_{k}$ |
| $e^{j \omega_{0}{ }^{\prime}}$ | $2 \pi \delta\left(\omega-\omega_{0}\right)$ | $\begin{aligned} & a_{1}=1 \\ & a_{k}=0, \quad \text { otherwise } \end{aligned}$ |
| $\cos \omega_{0} t$ | $\pi\left[\delta\left(\omega-\omega_{0}\right)+\delta\left(\omega+\omega_{0}\right)\right]$ | $\begin{aligned} & a_{1}=a_{-1}=\frac{1}{2} \\ & a_{k}=0, \quad \text { otherwise } \end{aligned}$ |
| $\sin \omega_{0} t$ | $\frac{\pi}{j}\left[\delta\left(\omega-\omega_{0}\right)-\delta\left(\omega+\omega_{0}\right)\right]$ | $\begin{aligned} & a_{1}=-a_{-1}=\frac{1}{2 j} \\ & a_{k}=0, \quad \text { otherwise } \end{aligned}$ |
| $x(t)=1$ | $2 \pi \delta(\omega)$ | $a_{0}=1, \quad a_{k}=0, k \neq 0$ <br> (this is the Fourier series representation for ) |
| Periodic square wave $x(t)= \begin{cases}1, & \|t\|<T_{1} \\ 0, & T_{1}<\|t\| \leq \frac{T}{2}\end{cases}$ <br> and $x(t+T)=x(t)$ | $\sum_{k=-\infty}^{+\infty} \frac{2 \sin k \omega_{0} T_{1}}{k} \delta\left(\omega-k \omega_{0}\right)$ | $\frac{\omega_{0} T_{1}}{\pi} \operatorname{sinc}\left(\frac{k \omega_{0} T_{1}}{\pi}\right)=\frac{\sin k \omega_{0} T_{1}}{k \pi}$ |
| $\sum_{n=-\infty}^{+\infty} \delta(t-n T)$ | $\frac{2 \pi}{T} \sum_{k=-\infty}^{+\infty} \delta\left(\omega-\frac{2 \pi k}{T}\right)$ | $a_{k}=\frac{1}{T}$ for all $k$ |
| $x(t) \begin{cases}1, & \|t\|<T_{1} \\ 0, & \|t\|>T_{1}\end{cases}$ | $\frac{2 \sin \omega T_{1}}{\omega}$ | - |
| $\frac{\sin W t}{\pi t}$ | $X(j \omega)= \begin{cases}1, & \|\omega\|<W \\ 0, & \|\omega\|>W\end{cases}$ | - |
| $\delta(t)$ | 1 | - |
| $u(t)$ | $\frac{1}{j \omega}+\pi \delta(\omega)$ | - |
| $\delta\left(t-t_{0}\right)$ | $e^{-j \omega t_{0}}$ | - |
| $e^{-a t} u(t), \mathcal{R} e\{a\}>0$ | $\frac{1}{a+j \omega}$ | - |
| $t e^{-a t} u(t), \mathcal{R e}\{a\}>0$ | $\frac{1}{(a+j \omega)^{2}}$ | - |
| $\begin{aligned} & \frac{n^{n-1}}{(n-1)!} e^{-a t} u(t), \\ & \mathfrak{Q}\{a\}>0 \end{aligned}$ | $\frac{1}{(a+j \omega)^{n}}$ | - |

table 5.1 PROPERTIES OF THE DISCRETE-TIME FOURIER TRANSFORM

| Section | Property | Aperiodic Signal | Fourier Transform |
| :---: | :---: | :---: | :---: |
|  |  | $x[n]$ | $X\left(e^{j \omega}\right)$ periodic with |
|  |  | $y[n]$ | $\left.Y\left(e^{j \omega}\right)\right\}$ period $2 \pi$ |
| 5.3.2 | Linearity | $a x[n]+b y[n]$ | $a X\left(e^{j \omega}\right)+b Y\left(e^{j \omega}\right)$ |
| 5.3.3 | Time Shifting | $x\left[n-n_{0}\right]$ | $e^{-j \omega n_{0}} X\left(e^{j \omega}\right)$ |
| 5.3.3 | Frequency Shifting | $e^{j \omega_{0} n} x[n]$ | $X\left(e^{j\left(\omega-\omega_{0}\right)}\right)$ |
| 5.3.4 | Conjugation | $x^{*}[n]$ | $X^{*}\left(e^{-j \omega}\right)$ |
| 5.3.6 | Time Reversal | $x[-n]$ | $X\left(e^{-j \omega}\right)$ |
| 5.3.7 | Time Expansion | $x_{(k)}[n]= \begin{cases}x[n / k], & \text { if } n=\text { multiple of } k \\ 0, & \text { if } n \neq \text { multiple of } k\end{cases}$ | $X\left(e^{j k \omega}\right)$ |
| 5.4 | Convolution | $x[n] * y[n]$ | $X\left(e^{j \omega}\right) Y\left(e^{j \omega}\right)$ |
| 5.5 | Multiplication | $x[n] y[n]$ | $\frac{1}{2 \pi} \int_{2 \pi} X\left(e^{j \theta}\right) Y\left(e^{j(\omega-\theta)}\right) d \theta$ |
| 5.3.5 | Differencing in Time | $x[n]-x[n-1]$ | $\left(1-e^{-j \omega}\right) X\left(e^{j \omega}\right)$ |
| 5.3.5 | Accumulation | $\sum_{k=-\infty}^{n} x[k]$ | $\frac{1}{1-e^{-j \omega}} X\left(e^{j \omega}\right)$ |
| 5.3.8 | Differentiation in Frequency | $n \times[n]$ | $\begin{aligned} & +\pi X\left(e^{j 0}\right) \sum_{k=-\infty}^{+\infty} \delta(\omega-2 \pi k) \\ & j \frac{d X\left(e^{j \omega}\right)}{d \omega} \end{aligned}$ |
| 5.3.4 | Conjugate Symmetry for Real Signals | $x[n]$ real | $\left\{\begin{array}{l} X\left(e^{j \omega}\right)=X^{*}\left(e^{-j \omega}\right) \\ \operatorname{Re}\left\{X\left(e^{j \omega}\right)\right\}=\mathcal{R e}^{-j}\left\{X\left(e^{-j \omega}\right)\right\} \\ \mathscr{I}_{n z\{ }\left\{X\left(e^{j \omega}\right)\right\}=-\mathcal{I}_{m}\left\{X\left(e^{-j \omega}\right)\right\} \\ \left\|X\left(e^{j \omega}\right)\right\|=\left\|X\left(e^{-j \omega}\right)\right\| \\ \Varangle X\left(e^{j \omega}\right)=-\Varangle X\left(e^{-j \omega}\right) \end{array}\right.$ |
| 5.3.4 | Symmetry for Real, Even Signals | $x[n]$ real an even | $X\left(e^{j \omega}\right)$ real and even . |
| 5.3.4 | Symmetry for Real, Odd Signals | $x[n]$ real and odd | $X\left(e^{j \omega}\right)$ purely imaginary and odd |
| 5.3.4 | Even-odd Decomposition of Real Signals | $\begin{array}{ll} x_{e}[n]=\mathcal{E v}\{x[n]\} & {[x[n] \text { real }]} \\ x_{o}[n]=\operatorname{dd}\{x[n]\} & {[x[n] \text { real }]} \end{array}$ |  |
| 5.3.9 | Parseval's Re $\sum_{n=-\infty}^{+\infty}\|x[n]\|$ | ation for Aperiodic Signals $=\frac{1}{2 \pi} \int_{2 \pi}\left\|X\left(e^{j \omega}\right)\right\|^{2} d \omega$ |  |

a duality relationship between the discrete-time Fourier transform and the continuous-time Fourier series. This relation is discussed in Section 5.7.2.

### 5.7.1 Duality in the Discrete-Time Fourier Series

Since the Fourier series coefficients $a_{k}$ of a periodic signal $x[n]$ are themselves a periodic sequence, we can expand the sequence $a_{k}$ in a Fourier series. The duality property for discrete-time Fourier series implies that the Fourier series coefficients for the periodic sequence $a_{k}$ are the values of $(1 / N) x[-n]$ (i.e., are proportional to the values of the original

TABLE 5.2 BASIC DISCRETE-TIME FOURIER TRANSFORM PAIRS

| Signal | Fourier Transform | Fourier Series Coefficients (if periodic) |
| :---: | :---: | :---: |
| $\sum_{k=\langle N\rangle} a_{k} e^{j k(2 n / N) n}$ | $2 \pi \sum_{k=-\infty}^{+\infty} a_{k} \delta\left(\omega-\frac{2 \pi k}{N}\right)$ | $a_{k}$ |
| $e^{j \omega_{0} n}$ | $2 \pi \sum_{l=-\infty}^{+\infty} \delta\left(\omega-\omega_{0}-2 \pi l\right)$ | (a) $\begin{aligned} & \omega_{0}=\frac{2 \pi m}{N} \\ & a_{k}= \begin{cases}1, & k=m, m \pm N, m \pm 2 N, \ldots \\ 0, & \text { otherwise }\end{cases} \end{aligned}$ <br> (b) $\frac{\omega_{0}}{2 \pi}$ irrational $\Rightarrow$ The signal is aperiodic |
| $\cos \omega_{0} n$ | $\pi \sum_{l=-\infty}^{+\infty}\left\{\delta\left(\omega-\omega_{0}-2 \pi l\right)+\delta\left(\omega+\omega_{0}-2 \pi l\right)\right\}$ | (a) $\begin{aligned} \omega_{0} & =\frac{2 \pi m}{N} \\ a_{k} & = \begin{cases}\frac{1}{2}, & k= \pm m, \pm m \pm N, \pm m \pm 2 N \\ 0, & \text { otherwise }\end{cases} \end{aligned}$ <br> (b) $\frac{\omega_{0}}{2 \pi}$ irrational $\Rightarrow$ The signal is aperiodic |
| $\sin \omega_{0} n$ | $\frac{\pi}{j} \sum_{l=-\infty}^{+\infty}\left\{\delta\left(\omega-\omega_{0}-2 \pi l\right)-\delta\left(\omega+\omega_{0}-2 \pi l\right)\right\}$ | (a) $\begin{aligned} & \omega_{0}\end{aligned} \quad=\frac{2 \pi r}{N} \quad \begin{aligned} \frac{1}{2 j}, & k=r, r \pm N, r \pm 2 N, \ldots,\end{aligned}, \begin{aligned}-\frac{1}{2 j}, & k=-r ;-r \pm N,-r \pm 2 N \\ 0, & \text { otherwise }\end{aligned}$ <br> (b) $\frac{\omega_{0}}{2 \pi}$ irrational $\Rightarrow$ The signal is aperiodic |
| $x[n]=1$ | $2 \pi \sum_{l=-\infty}^{+\infty} \delta(\omega-2 \pi l)$ | $a_{k}= \begin{cases}1, & k=0, \pm N, \pm 2 N, \ldots \\ 0, & \text { otherwise }\end{cases}$ |
| Periodic square wave $x[n]= \begin{cases}1, & \|n\| \leq N_{1} \\ 0, & N_{1}<\|n\| \leq N / 2\end{cases}$ <br> and $x[n+N]=x[n]$ | $2 \pi \sum_{k=-\infty}^{+\infty} a_{k} \delta\left(\omega-\frac{2 \pi k}{N}\right)$ | $\begin{aligned} & a_{k}=\frac{\sin \left[(2 \pi k / N)\left(N_{1}+\frac{1}{2}\right)\right]}{N \sin [2 \pi k / 2 N]}, k \neq 0, \pm N, \pm 2 N, \\ & a_{k}=\frac{2 N_{1}+1}{N}, k=0, \pm N, \pm 2 N, \ldots \end{aligned}$ |
| $\sum_{k=-\infty}^{+\infty} \delta[n-k N]$ | $\frac{2 \pi}{N} \sum_{k=-\infty}^{+\infty} \delta\left(\omega-\frac{2 \pi k}{N}\right)$ | $a_{k}=\frac{1}{N}$ for all $k$ |
| $a^{n} u[n], \quad\|a\|<1$ | $\frac{1}{1-a e^{-j \omega}}$ | - |
| $x[n]= \begin{cases}1, & \|n\| \leq N_{1} \\ 0, & \|n\|>N_{1}\end{cases}$ | $\frac{\sin \left[\omega\left(N_{1}+\frac{1}{2}\right)\right]}{\sin (\omega / 2)}$ | - |
| $\begin{aligned} & \frac{\sin W n}{\pi n}=\frac{W}{\pi} \operatorname{sinc}\left(\frac{W n}{\pi}\right) \\ & 0<W<\pi \end{aligned}$ | $\begin{aligned} & X(\omega)= \begin{cases}1, & 0 \leq\|\omega\| \leq W \\ 0, & W<\|\omega\| \leq \pi\end{cases} \\ & X(\omega) \text { periodic with period } 2 \pi \end{aligned}$ | - |
| $\delta[n]$ | 1 | - |
| $u[n]$ | $\frac{1}{1-e^{-j \omega}}+\sum_{k=-\infty}^{+\infty} \pi \delta(\omega-2 \pi k)$ | $-$ |
| $\delta\left[n-n_{0}\right]$ | $e^{-j \omega \mu_{0}}$ |  |
| $(n+1) a^{n} u[n], \quad\|a\|<1$ | $\frac{1}{\left(1-a e^{-j \omega}\right)^{2}}$ |  |
| $\frac{(n+r-1)!}{n!(r-1)!} a^{n} u[n], \quad\|a\|<1$ | $\frac{1}{\left(1-a e^{-j \omega}\right)^{r}}$ |  |

