

# 1.1. AM DSB - Amplitude Modulation, Double Side Band

## 1.2 - Init. Code Read

$W_1 = \text{LPF Width}$

$W_2 = \text{Modulation Shift for } x_1 - \cosine$

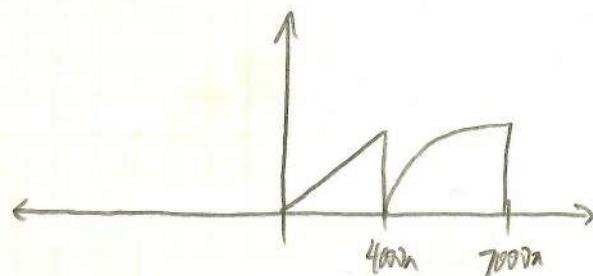
$W_3 = \text{Mod. shift } x_2 - \sin$

$3000\pi$

$4000\pi$

$7000\pi$

Q1.2 - Look at possible LSbs.



$$W_1 = 3000\pi$$

$x_1$  biggest LPF =  $4k\pi$

$x_2$  biggest =  $3k\pi$

Q1.3 - BW of  $x_2-h = 2500 \text{ Hz}$

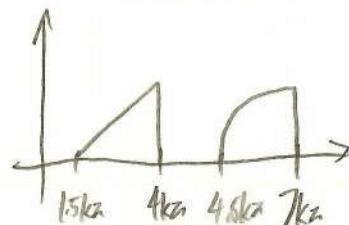
Q1.4 -  $W_4 = x_1 \text{ BPF for } x_1$

★  $0 \rightarrow 1.5\pi \times 10^3$  is A-signal.

$$\boxed{W_4 = 2000\pi}$$

$$\boxed{W_5 = 2000\pi}$$

Plot A



Q1.5  $W_6$  is upper side of BPF

$$0 \leq W_6 - W_7 \leq 4500\pi$$

$$W_6 = 7000\pi$$

$$2500\pi \leq W_7 \leq 7000\pi$$

Q1.6  $W_8$ : End LPF. Should be equal to BW of 1 side band.

$$W_8 = 2500\pi$$

$W_{11}$ : Upper side of transmitted  $X_2$

$$W_{11} = 7000\pi$$

$W_{12}$ : Lower side of transmitted  $X_2$

$$W_{12} = 4500\pi$$

$W_{13}$ :  $X_1$  Demod

$$W_{13} = 4000\pi$$

\*  $\pi$  is already in the code.

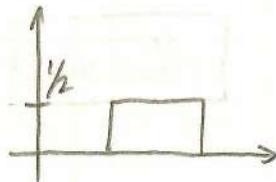
$W_{14}$ :  $X_2$  Demod

$$W_{14} = 7000\pi$$

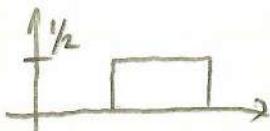
Q1.7a. No.

Q1.7b Observe the mod/demod. filters:

$$h_{\text{one}}: \text{sinc}(W_4 \cdot t) \cos(W_5 \cdot t) \rightarrow \text{FT:}$$



$$h_{\text{three}}: \text{sinc}(W_9 \cdot t) \cos(W_{10} \cdot t) \rightarrow \text{FT:}$$



Both filters also reduce amplitude by  $\frac{1}{2}$ .

Q1.7c: To fix on student end, multiply  $h_{\text{three}}$  by 4.

Note: Will cause noise to also increase.

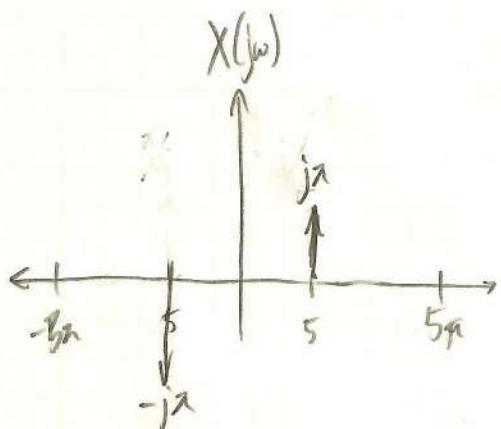
Q1.7d Yes

Q2.1 - Asym Demod - Defining characteristic is that sync. demod uses a carrier signal exactly in phase and in sync with the transmitted signal.

Asym does not - typically frequency is much higher, and phase doesn't matter.

Q2.2

$$x(t) = \sin(5t - \pi) \xrightarrow{\text{FT}} \frac{\pi}{j} (\delta(\omega-5) - \delta(\omega+5)) e^{-j\omega \cdot \frac{\pi}{5}}$$



$$\text{As. } f(t) \delta(t-a) = f(a) \delta(t-a)$$

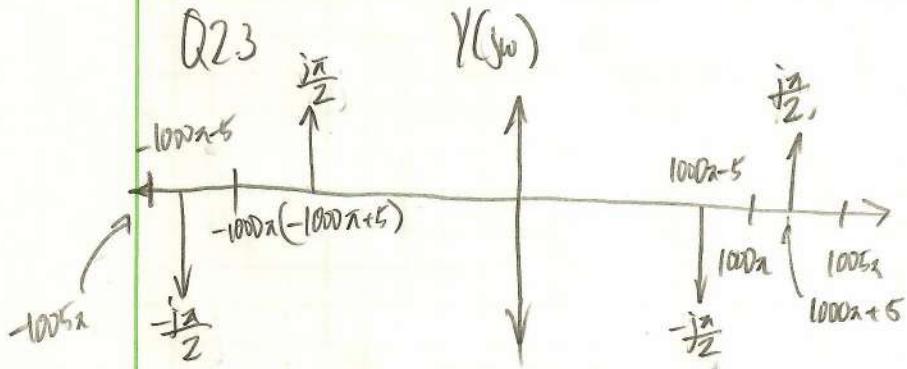
$$X(j\omega) = \frac{\pi}{j} \delta(\omega-5) e^{-j\omega} - \frac{\pi}{j} \delta(\omega+5) e^{j\omega}$$

$$e^{-j\omega} = e^{j\pi} = -1$$

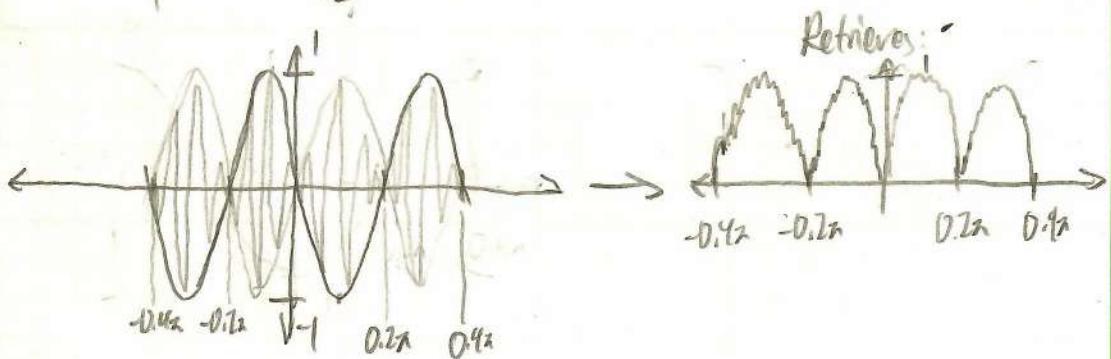
$$= -\frac{\pi}{j} \delta(\omega-5) + \frac{\pi}{j} \delta(\omega+5)$$

$$= j\pi \delta(\omega-5) - j\pi \delta(\omega+5)$$

Q2.3

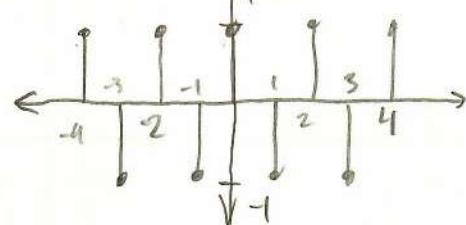


Q2.4



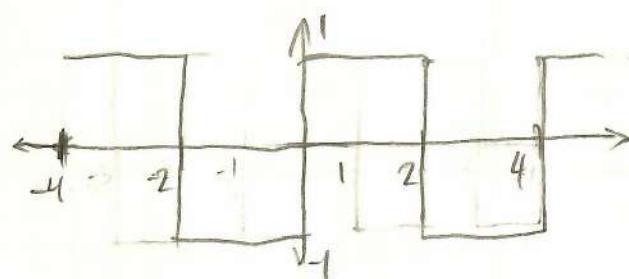
$$3.1 \quad x[n] = x[nT] = x[n/f]$$

$$x[n] = \cos(1.5\pi \cdot 2n) = \cos(3\pi n)$$



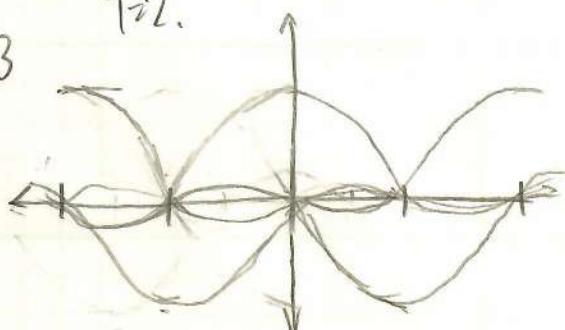
3.2

$$x_{\text{doh}}(t)$$

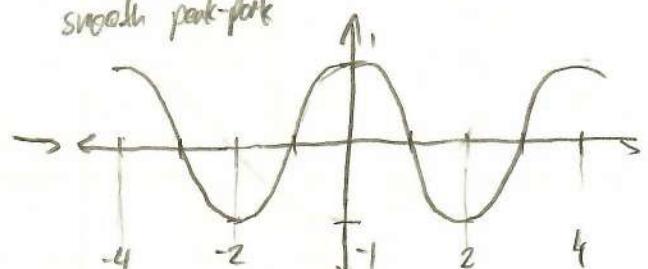


3.3

T/2.



This should  
smooth peak-park



Looks like  $\cos(0.5\pi t) +$

3.4 - A.  $\cos(0.5\pi t)$

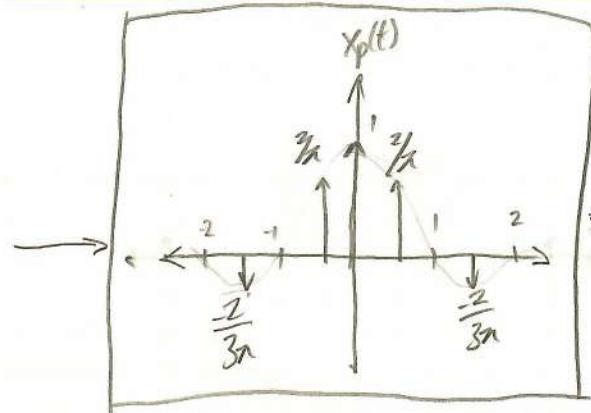
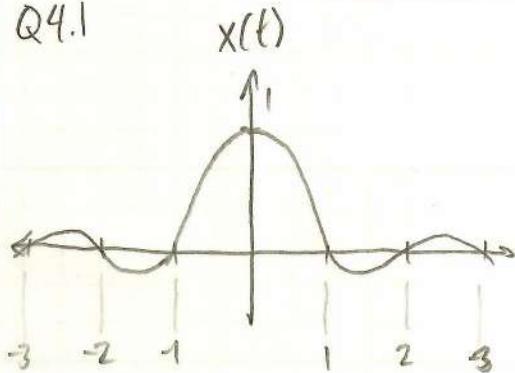
3.5. No. Reconstructed  $x_{\text{doh}}(t)$  is nowhere close,

No sampling could fix not sampling at Nyquist rate -

it's should have been sampled at 3Hz.

Because of that, sampling is off!

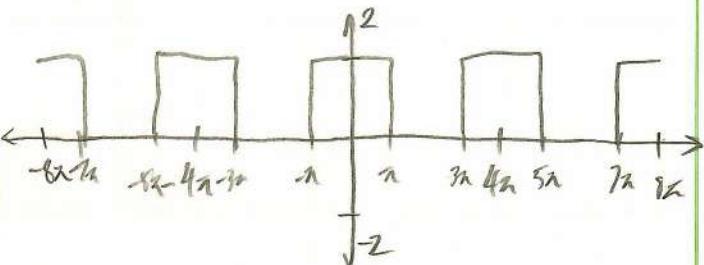
Q4.1



$$\sin\left(\frac{0.5\pi}{at}\right) \sim 1$$

$$Q4.2 \quad \sin(\pi t) \cdot \sum_{n=0}^{\infty} \delta(t - 0.5n)$$

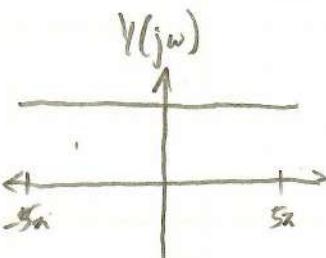
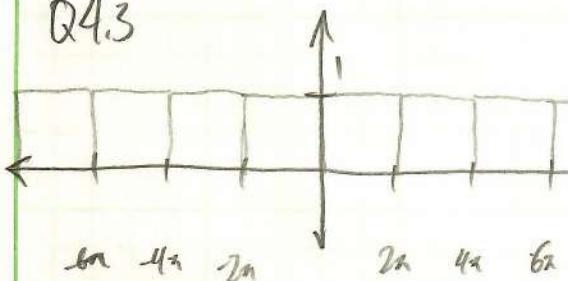
$$\frac{1}{2\pi} \left( \begin{cases} 1 & |\omega| < \pi \\ 0 & \text{else} \end{cases} \right) * \left( 4\pi \sum_{k=0}^{\infty} \delta(\omega - 4\pi k) \right)$$



Hint: As  $\delta(t-a)*f(t) = f(t-a)$ :

$$w(t) = \sin(3t-3) + \sin(3t-6) - \sin(3t-8)$$

Q4.3

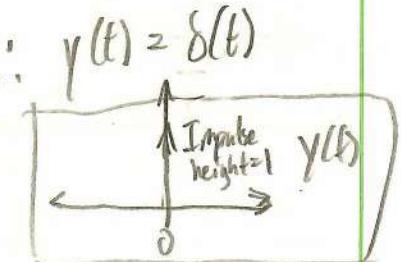


Making sense of this:

$$y(t) = \frac{\sin(\pi t) \cos(\pi t)}{\pi t} \cdot \sum_{n=0}^{\infty} \delta(t - 0.5n)$$

$$= \frac{1}{2} \cdot \frac{\sin(2\pi t)}{\pi t} \cdot \underbrace{\sum_{n=0}^{\infty} \delta(t - 0.5n)}_{\text{sampling}}$$

A Note:  
the sampling interval always makes  $\sin(2\pi t) \rightarrow 0$



Q5.

$$\cos(3t + 1.5) + \cos(6t + 3) + \cos(9t + 4.5) + \cos(12t + 6)$$

even function -  
remove negative

$$\frac{1}{2}e^{\frac{3}{2}j}e^{3jt} + \frac{1}{2}e^{\frac{3}{2}j}e^{-3jt} + \dots \quad \text{The key is the } \underline{\text{exponents.}}$$

We can similarly break down

the rest.

Matching it up to  $jk\frac{2n}{T}t$  for F.S. Inspection

Exponents are:

$$3jt \quad 6jt \quad 9jt \quad 12jt \rightarrow \text{If } \frac{2n}{T} = 3, \text{ ok } \checkmark \rightarrow \boxed{T = \frac{2n}{3}}$$

$\therefore$  By inspection...

$a_1 = \frac{1}{2}e^{\frac{3}{2}j}$	$a_1 = \frac{1}{2}e^{-\frac{3}{2}j}$
$a_2 = \frac{1}{2}e^{3j}$	$a_2 = \frac{1}{2}e^{-3j}$
$a_3 = \frac{1}{2}e^{\frac{9}{2}j}$	$a_3 = \frac{1}{2}e^{-\frac{9}{2}j}$
$a_4 = \frac{1}{2}e^{6j}$	$a_4 = \frac{1}{2}e^{-6j}$

Or else

Full expression:

$$\begin{aligned}
 & \left(\frac{1}{2}e^{\frac{3}{2}j}\right)e^{3jt} + \left(\frac{1}{2}e^{\frac{3}{2}j}\right)e^{-3jt} + \left(\frac{1}{2}e^{3j}\right)e^{6jt} + \left(\frac{1}{2}e^{3j}\right)e^{-6jt} \\
 & + \left(\frac{1}{2}e^{\frac{9}{2}j}\right)e^{9jt} + \left(\frac{1}{2}e^{\frac{9}{2}j}\right)e^{-9jt} + \left(\frac{1}{2}e^{6j}\right)e^{12jt} + \left(\frac{1}{2}e^{6j}\right)e^{-12jt}
 \end{aligned}$$

$$b. \quad x[n] = \frac{1}{2} e^{j\pi n} + \frac{1}{4} e^{j\frac{1}{4}\pi n} + \frac{1}{8} e^{j\frac{1}{8}\pi n}$$

Allow  $h[n]$  to form our bounds.

$$\sum_{k=-\infty}^{\infty} x[n-k] h[k] = \sum_{k=0}^{q_1} x[n-k]$$

$$= \sum_{k=0}^{q_1} \left( \frac{1}{2} e^{j\pi n} e^{-jk} + \frac{1}{4} e^{j\frac{1}{4}\pi n} e^{-jk} + \frac{1}{8} e^{j\frac{1}{8}\pi n} e^{-jk} \right) \quad \text{Each is dividable into its own sum.}$$

$$H-L+1=50 \quad = \frac{1}{2} e^{j\pi n} \sum_{k=0}^{q_1} (e^{-jk}) + \frac{1}{4} e^{j\frac{1}{4}\pi n} \sum_{k=0}^{q_1} (e^{-jk}) + \frac{1}{8} e^{j\frac{1}{8}\pi n} \sum_{k=0}^{q_1} (e^{-jk})$$

$$= \frac{1}{2} e^{j\pi n} \frac{e^{-j50\pi} (1 - e^{-j50})}{1 - e^{-j\pi}} + \frac{1}{4} e^{j\frac{1}{4}\pi n} \frac{e^{-j\frac{25}{2}\pi} (1 - e^{-j\frac{25}{2}\pi})}{1 - e^{-j\pi/4}} + \frac{1}{8} e^{j\frac{1}{8}\pi n} \frac{e^{-j\frac{5}{4}\pi} (1 - e^{-j\frac{5}{4}\pi})}{1 - e^{-j\pi/8}}$$

$$\star e^{-j50\pi} = 1 \\ e^{-j\pi} = -1$$

$$\star e^{-j\frac{25}{2}\pi} = -j$$

$$\boxed{= 0 + \frac{1}{4} e^{j\frac{1}{4}\pi n} \cdot \frac{-j(1+j)}{1 - e^{-j\pi/4}} + \frac{1}{8} e^{j\frac{1}{8}\pi n} \cdot \frac{e^{-j\frac{5}{4}\pi} (1 - e^{-j\frac{5}{4}\pi})}{1 - e^{-j\pi/8}}}$$

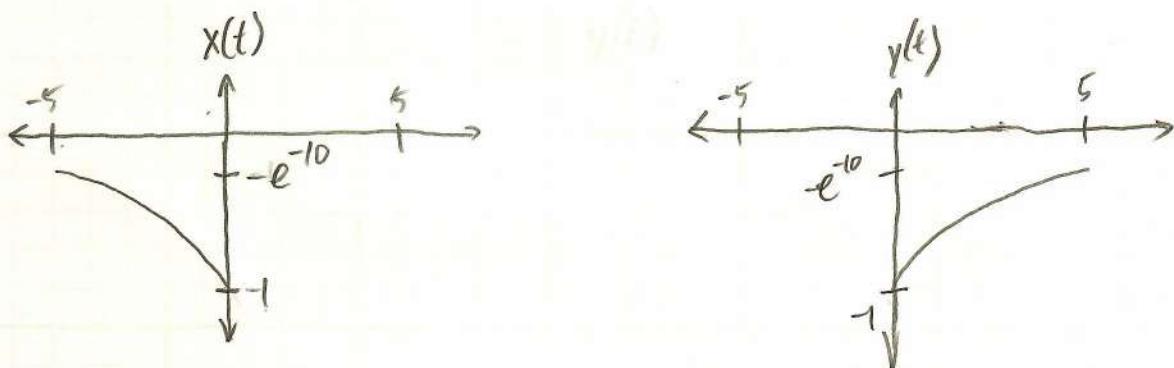
7.1  $X(j\omega) = -\frac{1}{2-j\omega} \rightarrow$  If we use time reversal

$$Y(j\omega) = X(-j\omega) = \frac{-1}{2+j\omega}$$

7.2  $Y(j\omega)$  fits form for  $e^{-at} u(t)$   $a > 0$

$$\therefore Y(t) = -e^{-2t} u(t)$$
  

$$\therefore X(t) = -e^{2t} u(-t)$$



Q6

$$h_1(t) = e^{2j \sum_{k=1}^5 \sin(kt)} \quad e^{ikt} - e^{-ikt} = 2j \sin(kt)$$

$$= e^{2j \sin(t)} e^{2j \sin(2t)} e^{2j \sin(3t)} e^{2j \sin(4t)} e^{2j \sin(5t)}$$

Individually - each is periodic. All go between  $e^{2j}$ ,  $e^{-2j}$

Together - They do have a period at  $T = 2\pi$

$\therefore h_1(t)$  periodic.

$$\text{If } t \text{ replaced w/-t: } h_1(-t) = \frac{1}{h_1(t)}$$

$\therefore h_1(t)$  Not even or odd

The absolute value of this char = 1.  
magnitude

$\therefore h_1(t)$  has finite power

Suppose this is an impulse response.

$h_1(t)$  extends beyond a strict  $\delta$  at the middle.

$\therefore$  Not memoryless

If  $t \geq 1$ ,  $h_1(t) \neq 0$

$\therefore$  Not causal

$$\int_{-\infty}^{\infty} |h_1(t)| dt = \int_{-\infty}^{\infty} 1 dt = \infty$$

$\therefore$  Not Stable

$$Q8-2. \quad h_2[n] = \sin\left(\frac{1}{2}\pi n + \frac{1}{2}\pi\right) + B + C$$

$$h_2[n] = A + B + C$$

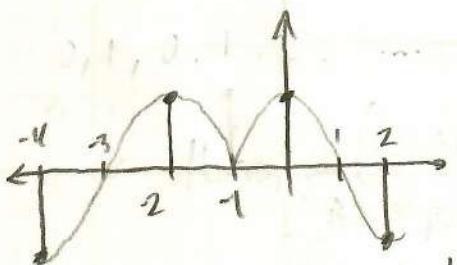
$$\frac{1}{2j} \left( e^{\frac{1}{2}jn\pi} e^{j\frac{\pi}{2}} - e^{-\frac{1}{2}jn\pi} e^{-j\frac{\pi}{2}} \right)$$

$$\frac{1}{2} \left( e^{j\frac{\pi n}{2}} + e^{-j\frac{\pi n}{2}} \right) = \cos\left(\frac{\pi n}{2}\right) = A$$

$$B: \sin\left(\frac{\pi}{2}|n|\right)$$

has discontinuity at  $n=1$  if CT.

So it is Not periodic.



If B not periodic,  $h_2[n]$  cannot be periodic.

$h_2[n]$  not periodic.

$h_2[n]$  Not even/odd.

$h_2[n]$  Finite Power

$h_2[n]$  cannot be even/odd due to asymmetry + offset

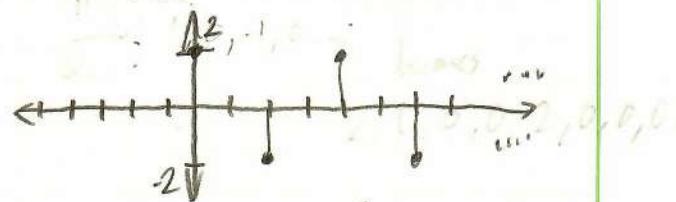
C: Eval. similarly.

$n$	C-val
-3	0
-2	0
-1	0
0	0
1	0
:	:

$$C=0$$

$$\therefore h_2[n] = \cos\left(\frac{\pi n}{2}\right) + \sin\left(0.5\pi|n|\right)$$

plot.



$h_2[n]$  has values outside of  $n \geq 0$

no values when  $n < 0$

will sum to  $\infty$  over  $\infty$  duration

$h_2[n]$  not memoryless,  
is causal  
not stable