

# Lecture 20

Note Title

3/26/2012

\* Irregular LDPC code ensemble

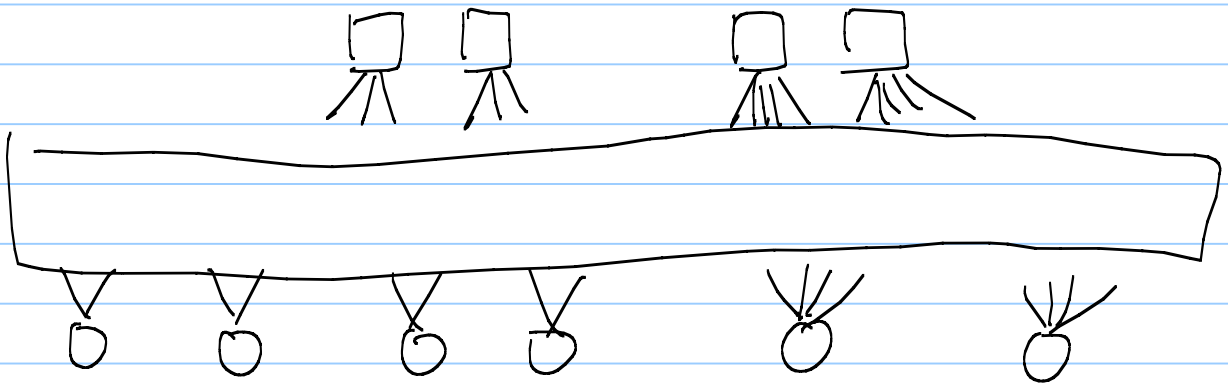
Example:  $\frac{2}{3}n$  variables of degree 2

$\frac{1}{3}n$  variables of degree 4

$\frac{1}{3}n$  check nodes of degree 3

$\frac{1}{3}n$  check nodes of degree 5

\* Consistency condition; Code rate



## \* Irregular LDPC code ensemble

Example:  $\frac{2}{3}n$  variables of degree 2

$\frac{1}{3}n$  variables of degree 4

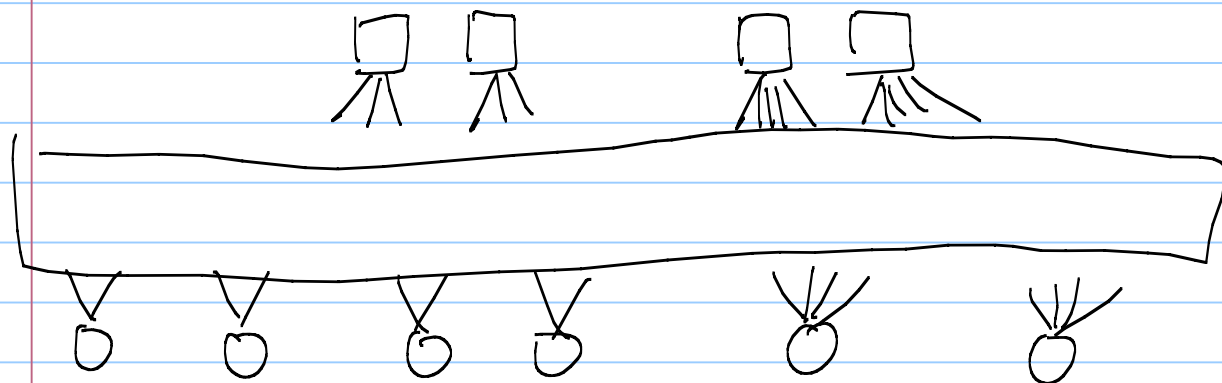
$\frac{1}{3}n$  check nodes of degree 3

$\frac{1}{3}n$  check nodes of degree 5

## \* Consistency condition

$$\text{Total \# edges: } \frac{2}{3}n \times 2 + \frac{1}{3}n \times 4 = \frac{1}{3}n \times 3 + \frac{1}{3}n \times 5$$

When  $n=6$ , we have the following FG.



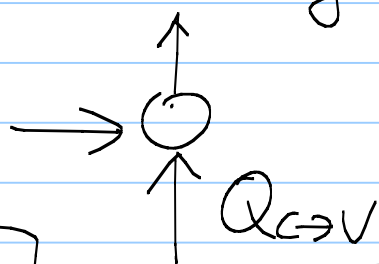
\* It is critical to keep the edge count consistent when we design the sockets of the var. & chk. nodes separately.

\* It is a rate  $\frac{1}{3}$  code  
( $n$  var nodes,  $\frac{2}{3}n$  check nodes)

Q: How to perform the DE analysis?

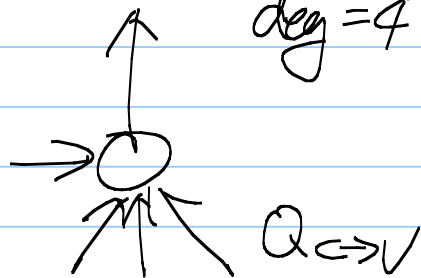
Variable

deg=2



$Q_{C \rightarrow V}$

deg=4



$Q_{C \rightarrow V}$

Observation 1

again, all incoming messages are i.i.d. due to the random interleaver

Check

deg=3



$P_{V \rightarrow C}$

deg 5



$P_{V \rightarrow C}$

Observation 2

assume we have the incoming message distribution  $P_{V \rightarrow C}$  &  $Q_{C \rightarrow V}$ , we can compute

Q: How to compute the update rule?

$$P_{V \rightarrow C}^{(t)} = \lambda_2 P_{V \rightarrow C}^{\text{deg}=2} + \lambda_4 P_{V \rightarrow C}^{\text{deg}=4}$$

Further average

$$= \lambda_2 \left( F_{\text{var}}(P^{(0)}, 2, Q_{C \rightarrow V}^{(t-1)}) \right) + \lambda_4 \left( F_{\text{var}}(P^{(0)}, 4, Q_{C \rightarrow V}^{(t-1)}) \right)$$

$P_{V \rightarrow C}^{\text{deg} 2}$   
 $P_{V \rightarrow C}^{\text{deg} 4}$   
 $Q_{C \rightarrow V}^{\text{deg} 3}$   
 $Q_{C \rightarrow V}^{\text{deg} 5}$

$$Q_{C \rightarrow V}^{(t)} = P_3 \left( F_{\text{chk}}(P_{V \rightarrow C}^{(t)}, 3) \right) + P_5 \left( F_{\text{chk}}(P_{V \rightarrow C}^{(t)}, 5) \right)$$

Q: What are the suitable values of

$\lambda_2, \lambda_4, \rho_3, \rho_5$ ?

Ans:

$$\lambda_2 = \frac{\frac{2}{3} n \times 2}{\frac{8}{3} n} = \frac{1}{2}$$

edges emitting from a deg 2 node

$$\lambda_4 = \frac{1}{2}$$

total # of edges

$$\rho_3 = \frac{\frac{1}{3} n \times 3}{\frac{8}{3} n} = \frac{3}{8}$$

edges emitting from a deg 3; check node

total # of edges

$$\rho_5 = \frac{5}{8}$$

We call  $\lambda_2, \lambda_4$ , the "edge-based" var degree distribution

$\rho_3, \rho_5$  the edge-based check deg distribution

edge-based

For general deg distributions  $\lambda_2, \dots, \lambda_{\max d_v}$   
 $\rho_2, \dots, \rho_{\max d_c}$

$$Q^{(t)} = \sum_{d=2}^{\max d_c} \rho_d F_{chk}(p^{(t)}, d-1)$$

$$p^{(t)} = \sum_{d=2}^{\max d_v} \lambda_d F_{var}(p^{(0)}, Q^{(t-1)}, d-1)$$

In the coding community, the deg distribution is usually described by a polynomial

$$\lambda(x) = \sum_{k=2}^{\max d_v} \lambda_k x^{k-1}$$

$$\rho(x) = \sum_{k=2}^{\max d_c} \rho_k x^{k-1}$$

and we say a  $(\lambda, \rho)$  irregular code ensemble

Exercise:

$$\text{given } \lambda(x) = 0.5x^2 + 0.5x^3$$

$$\rho(x) = 0.5x^4 + 0.5x^5$$

Exercise:

We require

$$\lambda(1) = \rho(1) = 1$$

Construct one such irr. code

The code rate of an irregular  $(\lambda, \rho)$  code

$$1 - \frac{\int_0^1 \rho(s) ds}{\int_0^1 \lambda(s) ds} \text{ is}$$

This special form of deg polynomials  
is very convenient.

$$\text{Ex: } \lambda(x) = 0.5x^2 + 0.5x^3 \quad \rho(x) = 0.5x^4 + 0.5x^5$$

Write down the DE formulas for BECs.

Ans: Suppose we are facing a BEC w. erasure  $\varepsilon$ .

$$p^{(0)} = \varepsilon, \quad p^{(1)} = p^{(0)}$$

$$q^{(t)} = \sum_{k=2}^{\max dc} p_k (1 - (1 - p^{(t)})^{k-1})$$

$$= 1 - P(1 - p^{(t)})$$

$$p^{(t+1)} = \sum_{k=2}^{\max dc} \lambda_k (p^{(0)} (q^{(t)})^{k-1})$$

$$= p^{(0)} \lambda(q^{(t)})$$

Sometimes the overall density evolution can  
be summarized as

$$p^{(t+1)} = \varepsilon \cdot \lambda(1 - P(1 - p^{(t)}))$$

\* DE analysis thus applies to irregular LDPC codes as well

\* How to <sup>use DE to</sup> design your own good LDPC codes of a given code rate  $R$ ?

① Fix  $\max d_v$   $\max d_c$ . Randomly choose  $\lambda_2, \lambda_3, \dots, \lambda_{\max d_v}$  that satisfies the rate constraint  $\rho_2, \rho_3, \dots, \rho_{\max d_c} \left| - \frac{\int_0^1 \rho dx}{\int_0^1 \lambda dx} = R \right.$

② Suppose the channel is parametrized by  $\sigma$ , Run DE for  $t_0$  say  $t_0 = 100$ , to see whether  $P_{\text{ber}}^{(t)} < 0.001$  (means convergence to zero)

③ Run a binary search to find  $\sigma^*$  of the given  $\lambda, \rho$

Hill-climbing algorithm, or any other general optimization techniques

④ Slightly perturb  $\lambda$ ,  $\rho$  while maintaining the code rate  $R$ , let  $\lambda^{\text{new}}$ ,  $\rho^{\text{new}}$  denote the perturbed version.

find the  $Q^{*,\text{new}}$  for  $\lambda^{\text{new}}$ ,  $\rho^{\text{new}}$

⑤ Retain the better degree distribution pair. Repeat ④

⑤① Sometimes we may be trapped in a local optimum point. Choose a different starting point & repeat ① — ⑤

⑥ With the optimized  $\lambda^*$ ,  $\rho^*$  construct a code with the desired codeword length. See the previous exercise.

Sometimes we have to quantify the  $\lambda^*$ ,  $\rho^*$  to fit the constraint of using a finite  $n$

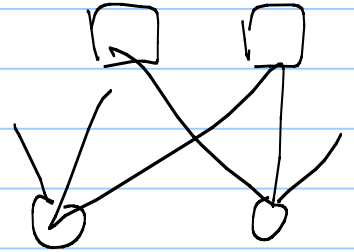


⑦ Arbitrarily select an interleaver with  
no obvious short cycles ex: cycle of

length 2



length 4



⑧ Convert the graph to the  $H$  matrix  
&  $G$  matrix for encoding

Voila! Very good performance LDPC  
codes.

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DE holds for ① any cycle-free FG,

or any asymptotically-cycle-free networks.

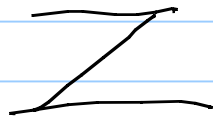
② Symmetry condition  $\Rightarrow$  all-zero codeword

③ identically distributed messages

$\Rightarrow$  remove the dependence on the  
subscript.

Pros: DE is thus quite general & predicts the performance well. (Think general large FG, rather than error control codes)

Cons: ① The symmetry condition ② is too strong.

Ex:  channel for optical storage.

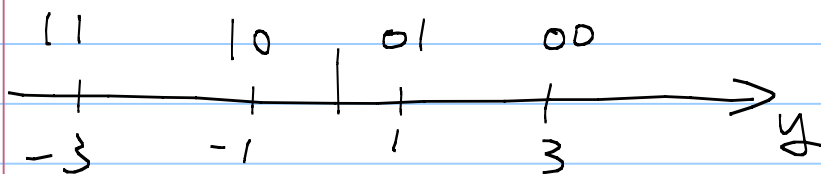
Example:

$$Y = 2(-1)^{X_1} + (-1)^{X_2} + N \quad \text{in the Euclidean space.}$$

$X_1, X_2$  are Bernoulli with  $p = 0.5$

$N$  is standard GSN.

the constellation points of this PAM are

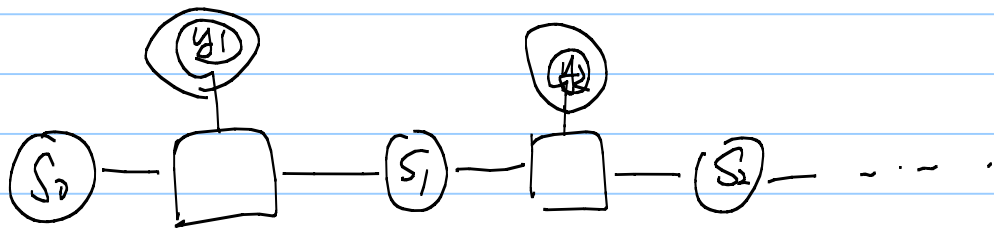


Note: This system is not symmetric, as the codewords 00, & 11 have smaller prob being misdetected as other codewords

$$P(\text{error} | \vec{X} = 00) < P(\text{error} | \vec{X} = 01)$$

② The density evolution computation is too complicated, since we are essentially computing the density of a random function when focusing on a general FG.

Ex: we have a <sup>linear</sup> Convolutional code,



Use the DE to analyze the performance

Ans: Theoretically doable as it is a cycle-free network.

However, each message  $\alpha$  or  $\beta$  is a  $|S|$ -dimensional vector.

We are tracing the density of  $|S|$ -dim vectors.  $\Rightarrow$  Hard.

On the other hand, analyzing convolutional codes is an important question as it relates to the turbo code performance

For the following, we will study more the DE & try to answer these challenges at least partially.

\* Deal with the 2nd drawback of DE  
(at least for general FGs), the complexity

Note: for binary LDPC codes, the exact DE  
is fast

\* Instead of tracing the exact density, we  
trace the approximate density.

We sacrifice some precision of the prediction  
for better efficiency.

of Gsn Approx

\* Gaussian Approximation (Too many variants)  
(We discuss some general procedures.)

Motivation 1: For AWGN channels:  $Y = (-1)^X + \sigma N$

the initial message distribution is

$$P_{m_i | X=0} \sim \text{Gsn} \left( \frac{2}{\sigma^2}, \frac{4}{\sigma^2} \right) \text{ Gsn.}$$

Motivation 2: If all incoming messages  
are Gsn, then the variable node  
message map

$$m_{X_i | C_j}^{(t+1)} = m_i^{(0)} + \sum_{l \in \partial(i) \setminus j} m_{C_l | X_i}^{(t)}$$

preserves the Gaussianity.

Remark: Unfortunately, the check node map

$$M_{C_j X_i}^{(t)} = 2 \tanh^{-1} \left( \prod_{k \in \partial j \setminus i} \tanh \left( \frac{m_{X_k C_j}^{(t)}}{2} \right) \right)$$

does not preserve the Gaussianity.

Motivation 3: If  $d_v$  is large,

$$m_{X_i C_j}^{(t+1)} = m_i^{(0)} + \underbrace{\sum_{k \in \partial i \setminus j} m_{C_k X_i}^{(t)}}_{d_v - 1 \text{ terms}}$$

then the var. message map will "restore" the Gaussianity due to the central limit theorem.

Remark: ① CLT requires normalization  
without normalization, simply taking convolution does not give you  $\mathcal{G}_{sn}$

② CLT requires medium-sized  $d_v$ .  
In practice,  $d_v$  is very small,  
(mostly 2 or 3, generally  $\leq 6$ )