

Contouring Control of Biaxial Systems Based on a New Task Coordinate Frame

Chuxiong Hu, Bin Yao, and Qingfeng Wang

Abstract—This paper proposes a new task coordinate frame (TCF) for contouring control of biaxial systems. Existing task coordinate frames are only locally defined based on the desired contouring trajectory to be tracked and the calculated contour error is an approximation to the actual contour error only. As such, they are applicable to contouring tasks with small curvature and little actual trajectory tracking errors only. In contrast, the proposed task coordinate frame is globally defined based on the geometry of the desired contour only and the resulting coordinate errors correspond to the actual contouring error and the tangential error on the contour directly. To demonstrate the high contouring performance nature of the proposed task coordinate frame, the system dynamics of a biaxial linear motor gantry is transformed into this task coordinate frame. A discontinuous projection based adaptive robust controller (ARC) which explicitly takes into account the dynamic coupling effect is then employed to improve the contouring performance under both parametric uncertainties and uncertain nonlinearities. Comparative experimental results obtained on a high-speed industrial biaxial gantry driven by linear motors are presented to verify that the proposed TCF is effective for achieving excellent contouring performance even when large-curvature contouring control tasks are of concern.

I. INTRODUCTION

Contour-following is one of the most ubiquitous motion control problems encountered by servomechanisms such as integrated circuit (IC) manufacturing equipments, robot manipulators, and computer numerical control (CNC) machine tools. Earlier researches on the coordinated contouring control [1]–[7] use the cross-coupled control (CCC) strategy first proposed by Koren [8], in which coupling actions was introduced in the servo controllers to strengthen the motion coordination of multi-axes. However, the proposed representations of contouring errors are approximate and effective only for applications with small curvature path. In addition, the presented algorithms were based on the traditional control theories for linear time-invariant systems and cannot address the dynamic coupling phenomena (e.g. Coriolis force) well when tracking curved contours. Chiu and Tomizuka [9] formulated the contouring control problem into a task coordinate frame “attached” to the desired contour.

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Under the task coordinate formulation, a control law could be designed to assign different dynamics to the normal and tangential directions relative to the desired contour. Since then many contouring control schemes based on task coordinate approaches have been reported [10]–[14]. However, the task coordinate frames used in all these work are locally defined based on the desired trajectory to be tracked on the desired contour, which is valid only for applications with very small actual tracking errors and small curvatures. They would not be sufficient when large curvature contouring is of main concern or significant tracking errors exhibit due to sudden disturbances.

In this paper, as opposed to approximate estimations of contouring errors [5], [7], [10], a new task coordinate frame will be developed based on the geometry of the desired contour only. Such a task coordinate frame (TCF) is globally defined and has nothing to do with the specific desired trajectory to be tracked on the contour. The calculated contouring error is directly related to the actual contouring error and not affected by the curvature of the desired contour. To test the effectiveness of the proposed TCF, the system dynamics of a biaxial linear motor gantry is transformed into this task coordinate frame, and a discontinuous projection based adaptive robust controller (ARC) [16], [17] which explicitly takes into account the dynamic coupling effect is then employed to improve the contouring performance under both parametric uncertainties and uncertain nonlinearities. For comparison, the traditional TCF [11]–[14] and the proposed TCF with the same ARC controller are both implemented on a high-speed industrial biaxial gantry driven by LC-50-200 linear motors with a linear encoder resolution of 0.5 μm . Comparative experimental results obtained on the biaxial gantry demonstrate that the proposed algorithm is able to achieve excellent contouring performance even when the tasks of high-speed and large-curvature contouring are required.

II. PROBLEM FORMULATION

A. Task Coordinate Frame

We will approach the definition of the task coordinate frame from a new perspective as follows. Traditionally, the task coordinate frame for a biaxial system is locally defined at the desired position $P_d(x_d(t), y_d(t))$ based on the desired trajectory on the desired contour given by

$$\mathbf{q}_d(t) = [x_d(t), y_d(t)]^T \quad (1)$$

Such a definition depend on not only the geometry of the desired contour (e.g., a circle or an ellipsoidal) but also

Thus, the equation (7) can be transformed into the task coordinates as

$$\mathbf{M}_t \ddot{\mathbf{r}} + \mathbf{B}_t \dot{\mathbf{r}} + \mathbf{C}_t \dot{\mathbf{r}} + \mathbf{A}_t \mathbf{S}_f(\dot{\mathbf{q}}) = \mathbf{u}_t + \mathbf{d}_t + \tilde{\Delta} \quad (9)$$

where

$$\begin{aligned} \mathbf{M}_t &= \mathbf{J}^{-T} \mathbf{M} \mathbf{J}^{-1}, \mathbf{B}_t = \mathbf{J}^{-T} \mathbf{B} \mathbf{J}^{-1}, \mathbf{C}_t = -\mathbf{J}^{-T} \mathbf{M} \mathbf{J}^{-1} \dot{\mathbf{J}} \mathbf{J}^{-1}, \\ \mathbf{A}_t &= \mathbf{J}^{-T} \mathbf{A}, \mathbf{d}_t = \mathbf{J}^{-T} \mathbf{d}_n, \mathbf{u}_t = \mathbf{J}^{-T} \mathbf{u}, \tilde{\Delta} = \mathbf{J}^{-T} \tilde{\mathbf{d}} \end{aligned} \quad (10)$$

It is well known that equation (9) has several properties [14]: (P1) \mathbf{M}_t is a symmetric positive definite (s.p.d.) matrix with $\mu_1 \mathbf{I} \leq \mathbf{M}_t \leq \mu_2 \mathbf{I}$ where μ_1 and μ_2 are two positive scalars; (P2) The matrix $\mathbf{N}_t = \dot{\mathbf{M}}_t - 2\mathbf{C}_t$ is a skew-symmetric matrix. In other words, $\mathbf{s}^T \mathbf{N}_t \mathbf{s} = 0, \forall \mathbf{s}$; (P3) $\mathbf{M}_t, \mathbf{B}_t, \mathbf{C}_t, \mathbf{A}_t, \mathbf{d}_t$ and \mathbf{u}_t in (9) can be linearly parameterized by a set of unknown parameters defined as $\theta = [\theta_1, \dots, \theta_8]^T = [M_1, M_2, B_1, B_2, A_1, A_2, d_{n1}, d_{n2}]^T$.

In general, θ cannot be known exactly. For example, the payload of the biaxial gantry depends on tasks. However, the extent of parametric uncertainties can be predicted. Therefore, the following practical assumption can be made.

Assumption 1: The extent of parametric uncertainties and uncertain nonlinearities is known. More precisely,

$$\begin{aligned} \theta &\in \Omega_\theta \triangleq \{\theta : \theta_{min} \leq \theta \leq \theta_{max}\} \\ \tilde{\Delta} &\in \Omega_\Delta \triangleq \{\tilde{\Delta} : \|\tilde{\Delta}\| \leq \delta_\Delta\} \end{aligned} \quad (11)$$

where $\theta_{min} = [\theta_{1min}, \dots, \theta_{8min}]^T$, and $\theta_{max} = [\theta_{1max}, \dots, \theta_{8max}]^T$ are known constant vectors and δ_Δ is a known function.

The control objective is to synthesize a control input \mathbf{u}_t such that $\mathbf{q} = [x, y]^T$ tracks a desired contour $\mathbf{q}_d(t) = [x_d, y_d]^T$ which is assumed to be at-least second-order differentiable. In the proposed task coordinate system, such an objective is achieved by simply regulating r_f to zero and letting r_g follow the desired trajectory of $r_{gd}(t) = r_g(x_d(t), y_d(t))$.

III. DISCONTINUOUS PROJECTION

Let $\hat{\theta}$ denote the estimate of θ and $\tilde{\theta}$ the estimation error (i.e., $\tilde{\theta} = \hat{\theta} - \theta$). In view of (11), the following adaptation law with discontinuous projection modification can be used

$$\dot{\hat{\theta}} = Proj_{\hat{\theta}}(\Gamma \tau) \quad (12)$$

where $\Gamma > 0$ is a diagonal matrix, τ is an adaptation function to be synthesized later. The projection mapping $Proj_{\hat{\theta}}(\bullet) = [Proj_{\hat{\theta}_1}(\bullet_1), \dots, Proj_{\hat{\theta}_8}(\bullet_8)]^T$ is defined in [16] as

$$Proj_{\hat{\theta}_i}(\bullet_i) = \begin{cases} 0 & \text{if } \hat{\theta}_i = \theta_{imax} \text{ and } \bullet_i > 0 \\ 0 & \text{if } \hat{\theta}_i = \theta_{imin} \text{ and } \bullet_i < 0 \\ \bullet_i & \text{otherwise} \end{cases} \quad (13)$$

It can be shown that for any adaptation function τ , the projection mapping used in (13) guarantees

$$(P4) \quad \hat{\theta} \in \Omega_\theta \triangleq \{\hat{\theta} : \theta_{imin} \leq \hat{\theta} \leq \theta_{imax}\} \quad (14)$$

$$(P5) \quad \tilde{\theta}^T (\Gamma^{-1} Proj_{\hat{\theta}}(\Gamma \tau) - \tau) \leq 0, \forall \tau \quad (15)$$

IV. ADAPTIVE ROBUST CONTROL (ARC) LAW

Define a switching-function-like quantity as

$$\mathbf{s} = \dot{\mathbf{e}} + \Lambda \mathbf{e} = \dot{\mathbf{r}} - \dot{\mathbf{r}}_{eq}, \dot{\mathbf{r}}_{eq} \triangleq \dot{\mathbf{r}}_d - \Lambda \mathbf{e} \quad (16)$$

where $\mathbf{e} = \mathbf{r}(t) - \mathbf{r}_d(t)$ is the output tracking error, and $\Lambda > 0$ is a constant diagonal matrix. Define a positive semi-definite (p.s.d.) function

$$V(t) = \frac{1}{2} \mathbf{s}^T \mathbf{M}_t(\mathbf{r}) \mathbf{s} \quad (17)$$

Differentiating V yields

$$\dot{V}(t) = \mathbf{s}^T [\mathbf{u}_t - \mathbf{M}_t \ddot{\mathbf{r}}_{eq} - \mathbf{B}_t \dot{\mathbf{r}} - \mathbf{C}_t \dot{\mathbf{r}}_{eq} - \mathbf{A}_t \mathbf{S}_f(\dot{\mathbf{q}}) + \mathbf{d}_t + \tilde{\Delta}] \quad (18)$$

where $\ddot{\mathbf{r}}_{eq} \triangleq \ddot{\mathbf{r}}_d - \Lambda \dot{\mathbf{e}}$, and (P2) is used to eliminate the term $\frac{1}{2} \mathbf{s}^T \dot{\mathbf{M}}_t(\mathbf{r}) \mathbf{s}$. Furthermore, since it follows from (P3) that

$$\mathbf{M}_t \ddot{\mathbf{r}}_{eq} + \mathbf{B}_t \dot{\mathbf{r}} + \mathbf{C}_t \dot{\mathbf{r}}_{eq} + \mathbf{A}_t \mathbf{S}_f(\dot{\mathbf{q}}) - \mathbf{d}_t = -\Psi(\mathbf{r}, \dot{\mathbf{r}}, \dot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq}) \theta \quad (19)$$

where $\Psi(\mathbf{r}, \dot{\mathbf{r}}, \dot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq})$ is a 2×8 matrix of known functions, known as the *regressor*. Equation (18) can be rewritten as

$$\dot{V}(t) = \mathbf{s}^T [\mathbf{u}_t + \Psi(\mathbf{r}, \dot{\mathbf{r}}, \dot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq}) \theta + \tilde{\Delta}] \quad (20)$$

Noting the structure of (20), the following ARC law is proposed:

$$\mathbf{u}_t = \mathbf{u}_a + \mathbf{u}_s, \mathbf{u}_a = -\Psi(\mathbf{r}, \dot{\mathbf{r}}, \dot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq}) \hat{\theta} \quad (21)$$

where \mathbf{u}_a is the adjustable model compensation needed for achieving perfect tracking, and \mathbf{u}_s is a robust control law to be synthesized later. Substituting (21) into (20), and then simplifying the resulting expression leads to

$$\dot{V}(t) = \mathbf{s}^T [\mathbf{u}_s - \Psi(\mathbf{r}, \dot{\mathbf{r}}, \dot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq}) \tilde{\theta} + \tilde{\Delta}] \quad (22)$$

The robust control function \mathbf{u}_s consists of two terms:

$$\mathbf{u}_s = \mathbf{u}_{s1} + \mathbf{u}_{s2}, \mathbf{u}_{s1} = -\mathbf{K} \mathbf{s} \quad (23)$$

where \mathbf{u}_{s1} is used to stabilize the nominal system, and it is a simple proportional feedback with \mathbf{K} being a symmetric positive definite matrix in this case, and \mathbf{u}_{s2} is a robust feedback used to attenuate the effect of model uncertainties. Noting Assumption 1 and (P4), there exists a \mathbf{u}_{s2} such that the following two conditions are satisfied

$$\begin{aligned} i \quad & \mathbf{s}^T \left\{ \mathbf{u}_{s2} - \Psi(\mathbf{r}, \dot{\mathbf{r}}, \dot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq}) \tilde{\theta} + \tilde{\Delta} \right\} \leq \eta \\ ii \quad & \mathbf{s}^T \mathbf{u}_{s2} \leq 0 \end{aligned} \quad (24)$$

where η is a design parameter that can be arbitrarily small. One smooth example of \mathbf{u}_{s2} satisfying (24) is given by $\mathbf{u}_{s2} = -\frac{1}{4\eta} H^2 \mathbf{s}$, where H is a smooth function satisfying $H \geq \|\theta_M\| \|\Psi(\mathbf{r}, \dot{\mathbf{r}}, \dot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq})\| + \delta_\Delta$, and $\theta_M = \theta_{max} - \theta_{min}$.

Theorem 1: Suppose the adaptation function in (12) is chosen as

$$\tau = \Psi^T(\mathbf{r}, \dot{\mathbf{r}}, \dot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq}) \mathbf{s} \quad (25)$$

Then, the ARC control law (21) guarantees that:

A. In general, all signals are bounded. Furthermore, the positive semi-definite function $V(t)$ defined by (17) is bounded above by

$$V(t) \leq \exp(-\lambda t) V(0) + \frac{\eta}{\lambda} [1 - \exp(-\lambda t)] \quad (26)$$

where $\lambda = 2\sigma_{\min}(\mathbf{K})/\mu_2$, and $\sigma_{\min}(\cdot)$ denotes the minimum eigenvalue of a matrix. Note that $\sigma_{\min}(\mathbf{K})$ is real and positive since \mathbf{K} is symmetric positive definite.

B. Suppose there exist parametric uncertainties only after a finite time t_0 , i.e., $\tilde{\Delta} = 0, \forall t \geq t_0$. Then, in addition to result A, zero final tracking error is also achieved, i.e., $\mathbf{e} \rightarrow 0$ and $\mathbf{s} \rightarrow 0$ as $t \rightarrow \infty$.

Proof: From of (P1), we have

$$\frac{1}{2}\mu_1\|\mathbf{s}\|^2 \leq V \leq \frac{1}{2}\mu_2\|\mathbf{s}\|^2 \quad (27)$$

From (22), (23), condition i of (24) and (27), it follows that

$$\begin{aligned} \dot{V}(t) &= -\mathbf{s}^T \mathbf{K} \mathbf{s} + \mathbf{s}^T [\mathbf{u}_{s2} - \Psi(\mathbf{r}, \dot{\mathbf{r}}, \ddot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq}) \tilde{\boldsymbol{\theta}} + \tilde{\Delta}] \\ &\leq -\sigma_{\min}(\mathbf{K})\|\mathbf{s}\|^2 + \eta = -\lambda V + \eta \end{aligned} \quad (28)$$

which leads to (26) and thus proves result A of Theorem 1. Now consider the situation in B of Theorem 1, i.e., $\tilde{\Delta} = 0, \forall t \geq t_0$. Choose a positive definite function $V_\theta(t)$ as

$$V_\theta(t) = V(t) + \frac{1}{2} \tilde{\boldsymbol{\theta}}^T \Gamma^{-1} \tilde{\boldsymbol{\theta}} \quad (29)$$

From (28), condition ii of (24) and (P5), it follows that

$$\dot{V}_\theta(t) \leq -\mathbf{s}^T \mathbf{K} \mathbf{s} - \mathbf{s}^T \Psi(\mathbf{r}, \dot{\mathbf{r}}, \ddot{\mathbf{r}}_{eq}, \ddot{\mathbf{r}}_{eq}) \tilde{\boldsymbol{\theta}} + \tilde{\boldsymbol{\theta}}^T \Gamma^{-1} \dot{\tilde{\boldsymbol{\theta}}} \leq -\mathbf{s}^T \mathbf{K} \mathbf{s} \quad (30)$$

which shows that $\mathbf{s} \in L_2 \cap L_\infty$. It is easy to check that $\dot{\mathbf{s}}$ is bounded. So, \mathbf{s} is uniformly continuous. By Barbalat's lemma, $\mathbf{s} \rightarrow 0$ as $t \rightarrow \infty$. ■

V. EXPERIMENTAL SETUP AND RESULTS

A. Experiment setup

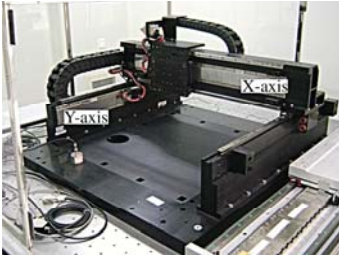


Fig. 2. A biaxial linear motor driven gantry system

To test the proposed controllers' effectiveness and study contour-following motion control of the biaxial gantry system, a biaxial Anorad gantry from Rockwell Automation is set up as a test-bed. As shown in Figure 2, the two axes of the gantry are mounted orthogonally with X-axis on top of Y-axis. The position sensors of the gantry are two linear encoders with a resolution of $0.5\mu\text{m}$ after quadrature. The velocity signal is obtained by the difference of two consecutive position measurements. Standard least-square identification is performed to obtain the parameters of the biaxial gantry and it is found that nominal values of the gantry system parameters without loads are $M_1 = 0.12 \text{ V/m} \cdot \text{s}^2$, $M_2 = 0.5 \text{ V/m} \cdot \text{s}^2$, $B_1 = 0.45 \text{ V/m} \cdot \text{s}$, $B_2 = 0.65 \text{ V/m} \cdot \text{s}$, $A_1 = 0.1 \text{ V}$, $A_2 = 0.15 \text{ V}$. The bounds of the parametric variations are chosen as $\boldsymbol{\theta}_{\min} = [0.06, 0.45, 0.35, 0.5, 0.05, 0.08, -0.5, -1]^T$ and $\boldsymbol{\theta}_{\max} = [0.25, 0.6, 0.55, 0.8, 0.15, 0.25, 0.5, 1]^T$

B. Experimental Results

The control algorithms are implemented using a dSPACE DS1103 controller board. The controller executes programs at a sampling period $T_s = 0.2\text{ms}$, which results in a velocity measurement resolution of 0.0025m/sec . To test the effectiveness of the proposed task coordinate frame, the following two task coordinate frames with ARC algorithms are compared:

C1: The traditional task coordinate frame presented in [11], [12] shown in Figure 3. Let x and y denote the horizontal and the vertical axes of a biaxial gantry system. P_d and P_a denote the position of the reference command and the actual position of the system at any time instant respectively. At the point P_d , the desired contour possesses a set of tangential and normal directions such that the contouring error can be approximated by the distance from P_a to the tangential line in the normal direction. The physical (x, y) coordinates can be transformed into the task $(\varepsilon_c, \varepsilon_t)$ coordinates by a linear time-varying transformation

$$\boldsymbol{\varepsilon} = \mathbf{T} \mathbf{e} \quad (31)$$

where $\boldsymbol{\varepsilon} = [\varepsilon_c, \varepsilon_t]^T$; $\mathbf{e} = [e_x, e_y]^T$, e_x and e_y denote the axial tracking errors of x and y axes, i.e. $e_x = x - x_d$, $e_y = y - y_d$, α denotes the angle between the tangential line to the horizontal X-axis, and the time-varying transformation matrix depends on the reference trajectories for the desired contour only and is given by

$$\mathbf{T} = \begin{bmatrix} -\sin\alpha & \cos\alpha \\ \cos\alpha & \sin\alpha \end{bmatrix} \quad (32)$$

The matrix \mathbf{T} is always unitary for all values of α , i.e., $\mathbf{T}^T = \mathbf{T}$ and $\mathbf{T}^{-1} = \mathbf{T}$. In this model, if the axial tracking errors are comparatively small to the curvature of the desired contour, then it yields a good approximation of the contouring error.

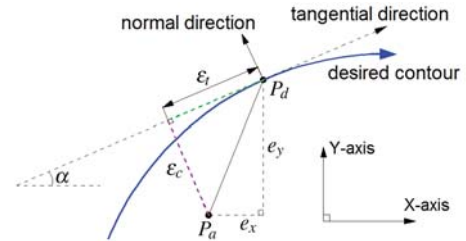


Fig. 3. An approximate contouring error model

C2: The proposed task coordinate frame in this paper.

In C1 and C2, the smooth functions $S_f(\dot{x})$ and $S_f(\dot{y})$ are chosen as $\frac{2}{\pi} \arctan(9000\dot{x})$ and $\frac{2}{\pi} \arctan(9000\dot{y})$. Λ is chosen as: $\Lambda = \text{diag}[100, 30]$. \mathbf{u}_{s2} in (23) is theoretically given in section IV. In implementation, a large enough constant feedback gain is used instead to simplify the resulting control law as done in [12]–[14]. With this simplification, the feedback gain in (23) is chosen as: $\mathbf{K} = \text{diag}[100, 60]$. The adaptation rates are set as $\Gamma = \text{diag}[10, 10, 10, 10, 1, 1, 10000, 10000]$. The initial parameter estimates are chosen as: $\hat{\boldsymbol{\theta}}(0) = [0.12, 0.5, 0.45, 0.65, 0.1, 0.15, 0, 0]^T$. Note that with these control gains, C1 is exactly same as that in [12].

The following test sets are performed:

- Set1 : To test the nominal contouring performance of the controllers, experiments are run without payload;
Set2 : A large step disturbance (a simulated 0.8 V electrical signal) is added to the input of Y axis at $t=10.4$ sec and removed at $t=13.4$ sec to test the performance robustness of each controller to disturbance.
Set3 : To test the performance robustness of the algorithms to parameter variations, a 5 kg payload is mounted on the gantry;

1) *Small-curvature elliptical contouring*: The gantry is first commanded to track a small-curvature contour given by

$$\mathbf{q}_d = \begin{bmatrix} x_d(t) \\ y_d(t) \end{bmatrix} = \begin{bmatrix} 0.2\sin(3t) \\ -0.1\cos(3t) + 0.1 \end{bmatrix} \quad (33)$$

which has an angular velocity of $w = 3\text{rad/sec}$.

2) *Large-curvature elliptical contouring*: The gantry is also commanded to track a large-curvature contour given by

$$\mathbf{q}_d = \begin{bmatrix} x_d(t) \\ y_d(t) \end{bmatrix} = \begin{bmatrix} 0.2\sin(4t) \\ -0.02\cos(4t) + 0.02 \end{bmatrix} \quad (34)$$

which has an angular velocity of $w = 4\text{rad/sec}$.

TABLE I
CONTOURING RESULTS

Small-curvature contouring results				
	$\ \epsilon_c\ _{rms}(\mu m)$	$\epsilon_{CM}(\mu m)$	$\ u_x\ _{rms}(V)$	$\ u_y\ _{rms}(V)$
C1 (Set1)	3.00	12.01	0.36	0.47
C2 (Set1)	2.39	8.76	0.37	0.47
C1 (Set2)	3.37	38.25	0.36	0.62
C2 (Set2)	2.91	38.16	0.37	0.62
C1 (Set3)	3.27	14.94	0.41	0.50
C2 (Set3)	2.42	9.66	0.41	0.50
Large-curvature contouring results				
	$\ \epsilon_c\ _{rms}(\mu m)$	$\epsilon_{CM}(\mu m)$	$\ u_x\ _{rms}(V)$	$\ u_y\ _{rms}(V)$
C1 (Set1)	2.59	12.15	0.46	0.27
C2 (Set1)	1.55	6.87	0.46	0.30
C1 (Set2)	2.99	36.39	0.46	0.46
C2 (Set2)	1.97	36.23	0.47	0.47
C1 (Set3)	2.68	12.56	0.57	0.28
C2 (Set3)	1.63	6.92	0.56	0.31

The experimental results of small-curvature and large-curvature contouring in terms of all performance indexes after running the gantry for several periods are given in Table I, in which the following performance indexes are used to measure the quality of each algorithm: (i) $\|\epsilon_c\|_{rms} = (\frac{1}{T} \int_0^T |\epsilon_c|^2 dt)^{1/2}$, the root-mean-square (RMS) value of the contouring error to measure average contouring performance where T is the total running time; (ii) $\epsilon_{CM} = \max\{\|\epsilon_c\|\}$, the maximum absolute value of the contouring error to measure transient performance; and (iii) $\|u_i\|_{rms} = (\frac{1}{T} \int_0^T |u_i|^2 dt)^{1/2}$, the average control input of each axis to evaluate the amount of control effort. As seen from the table, C2 performs better than C1 in terms of all indexes using almost the same amount of control efforts for every test set. Figure 4 – 9 show the results of the low-curvature and large-curvature contouring performances. In the small-curvature contouring

results, the desired contour and the actual contours by C1 and C2 are partially shown in Figure 4 for Set1, which directly demonstrates that C2 achieve better contouring results than that of C1 – the contour of C2 are quite closed to the desired contour; for Set2, the contouring errors are displayed in Figure 5, it is seem that the added large disturbances do not affect the contouring performance much except the transient spikes when the sudden changes of the disturbances occur; for Set3, the contouring errors are given in Figure 6, which shows that both controllers achieve good steady-state contouring performance in spite of the change of inertia load, also verifying the performance robustness of the proposed ARC controllers to parameter variations. In the large-curvature contouring results, the contouring errors are displayed in Figure 7 for Set1, it is seem that the C2 achieves much better contouring performances than that of C1, which demonstrates that C2 can achieve excellent contouring performance even when large-curvature contouring task is required; this conclusion can also be validated by Figure 8 for Set2 and Figure 9 for Set3, in which the desired contours and the actual contours by C1 and C2 are partially shown for further comparison and demonstration. Overall, all these results demonstrate that the proposed task coordinate frame can achieve excellent contouring performance. And the proposed frame can achieve better contouring performances than that of the traditional task coordinate frame, especially under the tasks of large-curvature contouring.

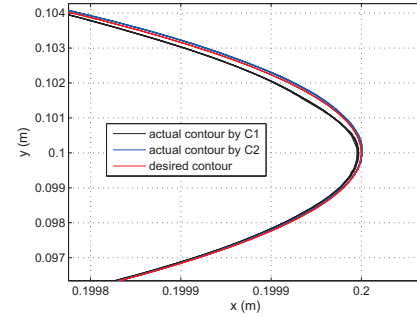


Fig. 4. Partial small-curvature contours of Set1

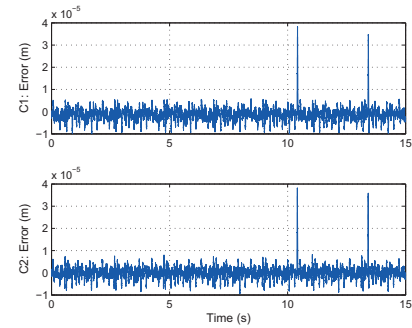


Fig. 5. Small-curvature contouring errors of Set2

VI. CONCLUSIONS

This paper proposed a global task coordinate frame (TCF) for contouring control of biaxial systems for high-speed and

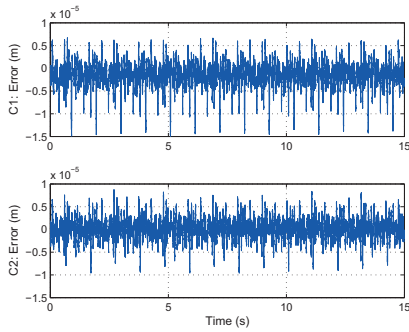


Fig. 6. Small-curvature contouring errors of Set3

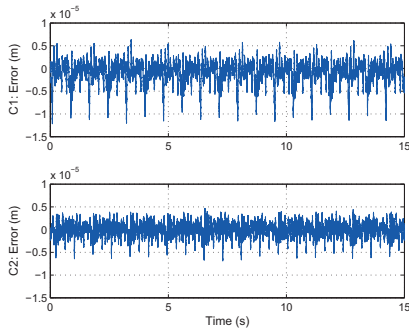


Fig. 7. Large-curvature contouring errors of Set1

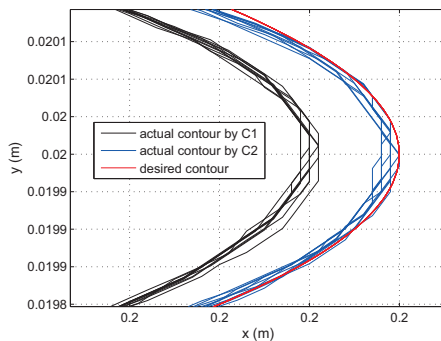


Fig. 8. Partial large-curvature contours of Set2

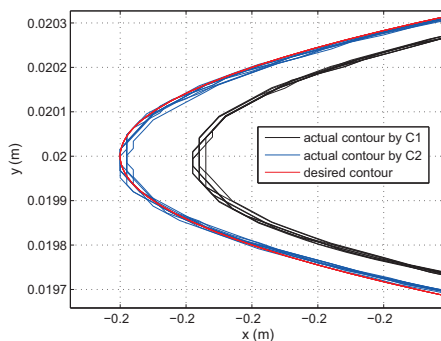


Fig. 9. Partial large-curvature contours of Set3

large-curvature contouring applications. Comparative experimental results obtained on a high-speed industrial biaxial gantry driven by linear motors were presented and verified the effectiveness and the excellent contouring performance of the proposed TCF based ARC controller for both high-speed small-curvature and large-curvature motions.

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