

EE 637 Final
May 3, Spring 2017

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Instructions:

- This is a 120 minute exam containing **five** problems.
- Each problem is worth 20 points for a total score of 100 points
- You may **only** use your brain and a pencil (or pen) to complete this exam.
- You **may not** use your book, notes, or a calculator, or any other electronic or physical device for communicating or accessing stored information.

Good Luck.

Fact Sheet

- Function definitions

$$\text{rect}(t) \triangleq \begin{cases} 1 & \text{for } |t| < 1/2 \\ 0 & \text{otherwise} \end{cases}$$

$$\Lambda(t) \triangleq \begin{cases} 1 - |t| & \text{for } |t| < 1 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{sinc}(t) \triangleq \frac{\sin(\pi t)}{\pi t}$$

- CTFT

$$X(f) = \int_{-\infty}^{\infty} x(t) e^{-j2\pi f t} dt$$

$$x(t) = \int_{-\infty}^{\infty} X(f) e^{j2\pi f t} df$$

- CTFT Properties

$$x(-t) \stackrel{CTFT}{\longleftrightarrow} X(-f)$$

$$x(t - t_0) \stackrel{CTFT}{\longleftrightarrow} X(f) e^{-j2\pi f t_0}$$

$$x(at) \stackrel{CTFT}{\longleftrightarrow} \frac{1}{|a|} X(f/a)$$

$$X(t) \stackrel{CTFT}{\longleftrightarrow} x(-f)$$

$$x(t) e^{j2\pi f_0 t} \stackrel{CTFT}{\longleftrightarrow} X(f - f_0)$$

$$x(t)y(t) \stackrel{CTFT}{\longleftrightarrow} X(f) * Y(f)$$

$$x(t) * y(t) \stackrel{CTFT}{\longleftrightarrow} X(f)Y(f)$$

$$\int_{-\infty}^{\infty} x(t)y^*(t)dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(f)Y^*(f)df$$

- CTFT pairs

$$\text{sinc}(t) \stackrel{CTFT}{\longleftrightarrow} \text{rect}(f)$$

$$\text{rect}(t) \stackrel{CTFT}{\longleftrightarrow} \text{sinc}(f)$$

For $a > 0$

$$\frac{1}{(n-1)!} t^{n-1} e^{-at} u(t) \stackrel{CTFT}{\longleftrightarrow} \frac{1}{(j2\pi f + a)^n}$$

- CSFT

$$F(u, v) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) e^{-j2\pi(ux+vy)} dx dy$$

$$f(x, y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u, v) e^{j2\pi(ux+vy)} du dv$$

- DTFT

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n}$$

$$x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

- DTFT pairs

$$a^n u(n) \stackrel{DTFT}{\longleftrightarrow} \frac{1}{1 - ae^{-j\omega}}$$

- Rep and Comb relations

$$\text{rep}_T[x(t)] = \sum_{k=-\infty}^{\infty} x(t - kT)$$

$$\text{comb}_T[x(t)] = x(t) \sum_{k=-\infty}^{\infty} \delta(t - kT)$$

$$\text{comb}_T[x(t)] \stackrel{CTFT}{\longleftrightarrow} \frac{1}{T} \text{rep}_{\frac{1}{T}}[X(f)]$$

$$\text{rep}_T[x(t)] \stackrel{CTFT}{\longleftrightarrow} \frac{1}{T} \text{comb}_{\frac{1}{T}}[X(f)]$$

- Sampling and Reconstruction

$$y(n) = x(nT)$$

$$Y(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X\left(\frac{\omega - 2\pi k}{2\pi T}\right)$$

$$s(t) = \sum_{k=-\infty}^{\infty} y(k) \delta(t - kT)$$

$$S(f) = Y(e^{j2\pi f T})$$

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Problem 1.(20pt)

Let $s(x, y) = \text{sinc}(\alpha x, \beta y)$ where x and y have units of inches and α and β are positive constants with units of inch^{-1} . Furthermore, let $g(m, n) = s(mA, nB)$ be a discrete-space function where A and B are the sampling intervals in the x and y directions, respectively, in inches.

- Calculate $S(u, v)$, the CSFT of $s(x, y)$.
- Give the cutoff frequencies of $s(x, y)$ in both x and y directions in units of cycles per inch.
- Specify the Nyquist sampling rates in units of cycles per inch along the x and y directions that are required to ensure perfect reconstruction. Also, specify constraints on the associated values of A and B in terms of α and β .
- Calculate $G(e^{j\mu}, e^{j\nu})$, the DSFT of $g(m, n)$.
- Sketch the function $G(e^{j\mu}, e^{j\nu})$ on the region $(\mu, \nu) \in (-\pi, \pi) \times (-\pi, \pi)$ when $A = B = 1$ and $\alpha = \beta = \frac{1}{2}$.

$$(a) \quad S(u, v) = \frac{1}{\alpha \beta} \text{rect}\left(\frac{u}{\alpha}, \frac{v}{\beta}\right)$$

$$(b) \quad u_c = \frac{\alpha}{2} \quad v_c = \frac{\beta}{2}$$

$$(c) \quad u_N = \alpha \quad v_N = \beta$$

sampling rate must be greater than or equal to Nyquist frequencies

$$\frac{1}{A} > \alpha, \quad \frac{1}{B} > \beta.$$

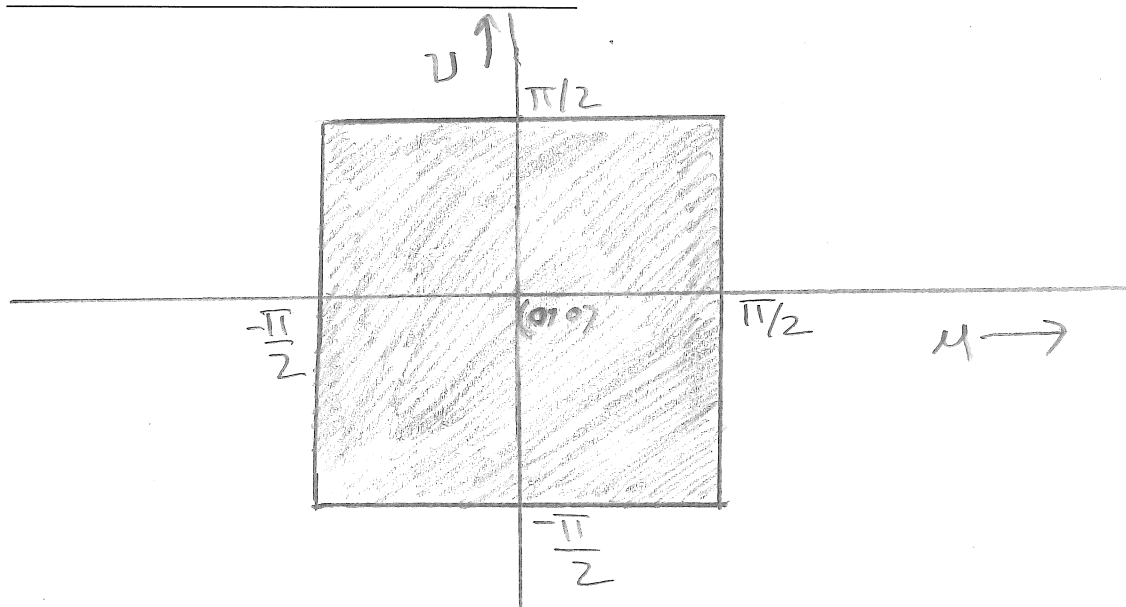
$$(d) \quad G(e^{j\mu}, e^{j\nu}) = \frac{1}{\alpha \beta} \frac{1}{AB} \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \text{rect}\left(\frac{u - 2\pi k}{2\pi \alpha A}, \frac{v - 2\pi l}{2\pi \beta B}\right)$$

(e) When $A = B = 1$ and $\alpha = \beta = \frac{1}{2}$, then

$$G(e^{j\mu}, e^{j\nu}) = 4 \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \text{rect}\left(\frac{u - 2\pi k}{\pi}, \frac{v - 2\pi l}{\pi}\right)$$

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$G(\omega, \phi) = 4$ in shaded region



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Problem 2.(20pt)

Consider a focal plane array with detector cell sizes $T \times T$, where T has units of inches. Let $g(x, y)$ be the incoming light signal, and let $s(m, n)$ be the measurement from the $(m, n)^{th}$ detector cell. The measurements $s(m, n)$ are related to the input light signal as

$$s(m, n) = \int_{\mathbb{R}^2} h(x - Tm, y - Tn) g(x, y) dx dy ,$$

where $h(x, y) = \text{rect}(\frac{x}{T}, \frac{y}{T})$. Furthermore, define

$$\tilde{g}(x, y) = h(x, y) ** g(x, y) ,$$

where $**$ represents 2-D convolution.

- Show that $s(m, n) = \tilde{g}(Tm, Tn)$, and use this result to draw a block-diagram of a signal-processing model of the focal-plane array.
- Calculate an expression for $\tilde{G}(u, v)$ the CSFT of $\tilde{g}(x, y)$ in terms of $G(u, v)$.
- Calculate an expression for $S(e^{j\mu}, e^{j\nu})$ the DSFT of $s(m, n)$ in terms $G(u, v)$.
- What is the maximum spatial frequency that the sampling system can reconstruct based on Nyquist theory?
- Calculate the frequency response of a linear space-invariant digital filter, $P(e^{j\mu}, e^{j\nu})$ that can compensate for the effect of the detector cell.

$$(a) \quad \tilde{g}(x, y) = h(x, y) ** g(x, y)$$

$$\Rightarrow \tilde{g}(\epsilon, \eta) = \int_{\mathbb{R}^2} h(\epsilon - x, \eta - y) g(x, y) dx dy$$

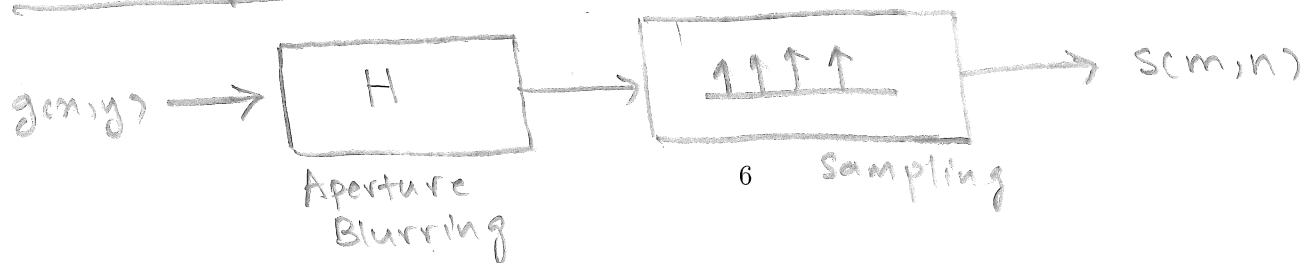
$$= \int_{\mathbb{R}^2} h(x - \epsilon, y - \eta) g(x, y) dx dy \quad \because h \text{ is symmetric.}$$

Now let's sample $\tilde{g}(\epsilon, \eta)$ at $\epsilon = mT, \eta = nT$.

$$\Rightarrow \tilde{g}(mT, nT) = \int_{\mathbb{R}^2} h(x - mT, y - nT) g(x, y) dx dy$$

$$= s(m, n)$$

Block Diagram



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$$\begin{aligned} \text{(b)} \quad \tilde{G}(u, v) &= H(u, v) G(u, v) \\ &= T^2 \text{sinc}(uT, vT) G(u, v) \end{aligned}$$

$$\begin{aligned} \text{(c)} \quad S(e^{j\mu}, e^{j\nu}) &= \frac{1}{T^2} \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \tilde{G}\left(\frac{\mu - 2\pi k}{2\pi T}, \frac{\nu - 2\pi l}{2\pi T}\right) \\ &= \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} \text{sinc}\left(\frac{\mu - 2\pi k}{2\pi}, \frac{\nu - 2\pi l}{2\pi}\right) G\left(\frac{\mu - 2\pi k}{2\pi T}, \frac{\nu - 2\pi l}{2\pi T}\right) \end{aligned}$$

(d) Maximum spatial frequency that can be reconstructed is $\frac{1}{T}$ cycles per inch in both horizontal and vertical directions.

$$\text{(e)} \quad P(e^{j\mu}, e^{j\nu}) = \frac{1}{\text{sinc}\left(\frac{\mu}{2\pi}, \frac{\nu}{2\pi}\right)}$$

This filter would compensate for the roll-off.

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Problem 3.(20pt)

Let $Y \in \mathbb{R}^N$ be a vector containing the pixels in an image window. We can model Y as

$$Y = \mu X + W$$

where $\mu \in \mathbb{R}^N$ is a deterministic column vector of length N , $X \in \mathbb{R}$ is scalar valued Gaussian random variable with mean 0 and variance σ^2 , and $W \in \mathbb{R}^N$ is a Gaussian random vector of i.i.d. noise components, each with distribution $N(0, \sigma_w)$.

Intuitively, Y is composed of a signal μ modulated by an amplitude X obscured by additive noise W . Our objective is to estimate the value of X from the observations Y . To do this, we will form a MMSE linear estimator for S given by

$$\hat{X} = Y^t \theta$$

where $\theta \in \mathbb{R}^N$ is a vector of coefficients.

Furthermore, define the covariance matrix of W given by

$$R_w = E[WW^t] ,$$

the covariance matrix of Y given by

$$R_y = E[YY^t] ,$$

and the cross-covariance column vector of Y and X given by

$$b = E[YX] .$$

- Give a simplified expression for R_w in terms of σ_w .
- Give a simplified expression for R_y in terms of σ_w , σ and μ .
- Give a simplified expression for b in terms of σ_w , σ and μ .
- Calculate an expression for the MSE given by $E[||X - \hat{X}||^2]$ in terms of σ^2 , σ_w^2 , θ , and μ .
- Use the above results to calculate a closed form expression for $\hat{X}$.

(a) $R_w = E[WW^t] = \sigma_w^2 I_N$ where I_N is $N \times N$ identity matrix.

(b)
$$\begin{aligned} R_y &= E[YY^t] \\ &= E[(\mu X + W)(\mu X + W)^t] \\ &= E[\mu \mu^t X^2 + 2\mu W^t X + WW^t] \\ &= \mu \mu^t E[X^2] + 2\mu E[W^t X] + E[WW^t] \end{aligned}$$

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Since W and X are independent, so

$$E[W^t X] = E[W^t] E[X] = 0$$

$$R_y = \mu \mu^t \sigma^2 + \sigma_w^2 I$$

$$\begin{aligned} (c) \quad b &= E[YX] \\ &= E[\mu X^2 + XW] \\ &= \mu E[X^2] + E[X] E[W] \\ &= \mu \sigma^2 \end{aligned}$$

$$\begin{aligned} (d) \quad E[\|X - \hat{X}\|^2] &= E[(X - Y^t \theta)^2] \\ &= E[X^2 - 2\theta^t Y X + \theta^t Y Y^t \theta] \\ &= E[X^2] - 2\theta^t E[YX] + \theta^t E[Y Y^t] \theta \\ &= \sigma^2 - 2\theta^t b + \theta^t R_y \theta \\ &= \sigma^2 - 2\theta^t \mu \sigma^2 + \theta^t (\mu \mu^t \sigma^2 + \sigma_w^2 I) \theta \end{aligned}$$

$$(e) \quad \frac{\partial}{\partial \theta} E[\|X - \hat{X}\|^2] = -2b + 2R_y \theta = 0$$

$$\Rightarrow \theta = R_y^{-1} b$$

$$\Rightarrow \theta = (\mu \mu^t \sigma^2 + \sigma_w^2 I)^{-1} \mu \sigma^2$$

Note: R_y is positive definite because it's the sum of a

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positive definite matrix $\sigma^2 I$ and a positive semidefinite matrix MM^T . Therefore θ is the unique global minimizer of $E[\|x - \hat{x}\|^2]$.

Since $\hat{x} = Y^T \theta$, so

$$\hat{x} = Y^T R_Y^{-1} b.$$

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Problem 4.(20pt)

Consider an image $f(x, y)$ with a CSFT given by $F(u, v)$ and a forward projection

$$\begin{aligned} p_\theta(r) &= \mathcal{FP}\{f(x, y)\} \\ &= \int_{-\infty}^{\infty} f(r \cos(\theta) - z \sin(\theta), r \sin(\theta) + z \cos(\theta)) dz. \end{aligned}$$

- a) Calculate the projections $p_\theta(r)$ of the function $f(x, y) = \delta(x, y)$.
- b) Calculate the projections $p_\theta(r)$ of the function $f(x, y) = \delta(x - 1, y)$.
- c) Calculate the projections $p_\theta(r)$ of the function $f(x, y) = \text{circ}(x, y)$.
- d) Calculate the projections $p_\theta(r)$ of the function $f(x, y) = \text{circ}(x - 1, y)$.

(a) since $\delta(x, y)$ is rotationally invariant, we shall find $p_0(r)$

$$\begin{aligned} p_0(r) &= p_0(r) = \int_{-\infty}^{\infty} \delta(r, z) dz = \int_{-\infty}^{\infty} \delta(r) \delta(z) dz \\ &= \delta(r). \end{aligned}$$

(b) Let's put $\begin{bmatrix} x' \\ y' \end{bmatrix} = \begin{bmatrix} x-1 \\ y \end{bmatrix}$.

$$p_0(r') = \int_{-\infty}^{\infty} f\left(A \begin{bmatrix} r' \\ z' \end{bmatrix}\right) dz$$

$$\begin{aligned} \text{However } \begin{bmatrix} r' \\ z' \end{bmatrix} &= A^{-1} \begin{bmatrix} x' \\ y' \end{bmatrix} = A^{-1} \begin{bmatrix} x \\ y \end{bmatrix} + A^{-1} \begin{bmatrix} -1 \\ 0 \end{bmatrix} \\ &= \begin{bmatrix} r \\ z \end{bmatrix} - \begin{bmatrix} \cos \theta \\ -\sin \theta \end{bmatrix} \end{aligned}$$

$$\Rightarrow p_0(r') = p_0(r - \cos \theta).$$

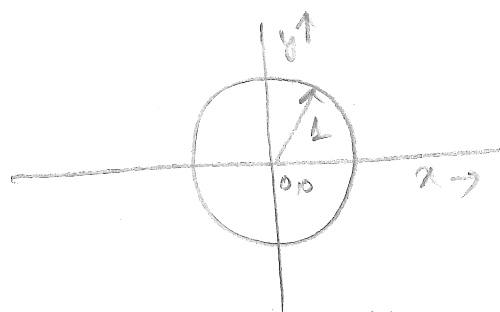
$$\text{So when } f(x, y) = \delta(x-1, y) \Rightarrow p_0(r) = \delta(r - \cos \theta).$$

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So if $f(x,y) = f(x-1,y)$

$$\Rightarrow p_0(r) = f(r - \cos \theta).$$

$$(c) \quad \text{circc}(x,y) = \begin{cases} 1 & \text{when } x^2 + y^2 \leq 1/2 \\ 0 & \text{otherwise} \end{cases}$$



This is also rotationally invariant, so

$$p_0(r) = p_0(r) = \int_{-\sqrt{\frac{1-r^2}{4}}}^{\sqrt{\frac{1-r^2}{4}}} 1 \cdot dz \quad \text{when } |r| \leq 1/2$$

$$= 2 \sqrt{\frac{1-r^2}{4}} \quad \text{when } |r| \leq 1/2$$

$$p_0(r) = \begin{cases} 2 \sqrt{\frac{1-r^2}{4}} & \text{when } r \leq 1 \\ 0 & \text{otherwise.} \end{cases}$$

(d) using result of part (b)

$$p_0(r) = \begin{cases} 2 \sqrt{\frac{1}{4} (r - \cos \theta)^2} & |r - \cos \theta| \leq 1/2 \\ 0 & |r - \cos \theta| > 1/2 \end{cases}$$

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Problem 5.(20pt)

Let $x(t)$ be a continuous-time signal, and $y(n) = x(Tn)$ be the associated discrete-time sampled signal.

a) Give an expression for the discrete approximation to

$$\left. \frac{dx(t)}{dt} \right|_{t=Tn}$$

the derivative of $x(t)$ at time $t = Tn$ in terms of the sampled signal $y(n)$.

b) Give an expression for the discrete approximation to

$$\left. \frac{d^2x(t)}{dt^2} \right|_{t=Tn}$$

the second derivative of $x(t)$ at time $t = Tn$ in terms of the sampled signal $y(n)$.

c) Give a general condition for detecting an edge in the signal based on the first discrete derivative.

d) Give a general condition for detecting an edge in the signal based on a combination of the first and second discrete derivatives.

e) What is the advantage of using both the first and second derivatives?

$$(a) \quad \frac{dx(t)}{dt} \approx \frac{x(t) - x(t-T)}{T}$$

$$\begin{aligned} \left. \frac{dx(t)}{dt} \right|_{t=nT} &\approx \frac{x(nT) - x(nT-T)}{T} \\ &= \frac{y(n) - y(n-1)}{T} \end{aligned}$$

$$(b) \quad \frac{d^2x(t)}{dt^2} \approx \frac{x'(t+T) - x'(t)}{T}$$

$$\left. \frac{d^2x(t)}{dt^2} \right|_{t=nT} \approx \frac{\frac{x(nT+T) - x(nT)}{T} - \frac{x(nT) - x(nT-T)}{T}}{T}$$

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$$= \frac{y(n+1) - y(n) - y(n) + y(n-1)}{T^2}$$

$$= \frac{-2y(n) + y(n-1) + y(n+1)}{T^2}$$

(c) Condition based on 1st derivative for edge detection:

$$|y(n) - y(n-1)| > K \quad \text{where } K \text{ is threshold.}$$

(d) conditions based on 1st + 2nd derivative for edge detection:

$$\text{Let } f_1(n) = y(n) - y(n-1) \quad \text{and} \quad f_2(n) = -2y(n) + y(n-1) + y(n+1)$$

Conditions

$$(1) \quad |f_1(n)| > K$$

$$(2) \quad f_2(n-1) f_2(n+1) < 0$$

where K_1 and K_2 are thresholds.

(e) By using both derivatives, we would have less false alarms in flat regions.

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