

EE 637 Midterm I
February 21, Spring 2014

Name: (4 pt) _____

Instructions:

- This is a 50 minute exam containing **three** problems.
- You may **only** use your brain and a pencil (or pen) and the included “Fact Sheet” to complete this exam.
- You **may not** use or have access to your book, notes, any supplementary reference, a calculator, or any communication device including a cell-phone or computer.
- You **may not** communicate with any person other than the official proctor during the exam.

Good Luck.

Fact Sheet

- Function definitions

$$\text{rect}(t) \triangleq \begin{cases} 1 & \text{for } |t| < 1/2 \\ 0 & \text{otherwise} \end{cases}$$

$$\Lambda(t) \triangleq \begin{cases} 1 - |t| & \text{for } |t| < 1 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{sinc}(t) \triangleq \frac{\sin(\pi t)}{\pi t}$$

- CTFT

$$X(f) = \int_{-\infty}^{\infty} x(t)e^{-j2\pi ft} dt$$

$$x(t) = \int_{-\infty}^{\infty} X(f)e^{j2\pi ft} df$$

- CTFT Properties

$$x(-t) \stackrel{CTFT}{\Leftrightarrow} X(-f)$$

$$x(t - t_0) \stackrel{CTFT}{\Leftrightarrow} X(f)e^{-j2\pi ft_0}$$

$$x(at) \stackrel{CTFT}{\Leftrightarrow} \frac{1}{|a|} X(f/a)$$

$$X(t) \stackrel{CTFT}{\Leftrightarrow} x(-f)$$

$$x(t)e^{j2\pi f_0 t} \stackrel{CTFT}{\Leftrightarrow} X(f - f_0)$$

$$x(t)y(t) \stackrel{CTFT}{\Leftrightarrow} X(f) * Y(f)$$

$$x(t) * y(t) \stackrel{CTFT}{\Leftrightarrow} X(f)Y(f)$$

$$\int_{-\infty}^{\infty} x(t)y^*(t)dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(f)Y^*(f)df$$

- CTFT pairs

$$\text{sinc}(t) \stackrel{CTFT}{\Leftrightarrow} \text{rect}(f)$$

$$\text{rect}(t) \stackrel{CTFT}{\Leftrightarrow} \text{sinc}(f)$$

For $a > 0$

$$\frac{1}{(n-1)!} t^{n-1} e^{-at} u(t) \stackrel{CTFT}{\Leftrightarrow} \frac{1}{(j2\pi f + a)^n}$$

- CSFT

$$F(u, v) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} f(x, y) e^{-j2\pi(ux+vy)} dx dy$$

$$f(x, y) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} F(u, v) e^{j2\pi(ux+vy)} du dv$$

- DTFT

$$X(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x(n)e^{-j\omega n}$$

$$x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(e^{j\omega}) e^{j\omega n} d\omega$$

- DTFT pairs

$$a^n u(n) \stackrel{DTFT}{\Leftrightarrow} \frac{1}{1 - ae^{-j\omega}}$$

- Rep and Comb relations

$$\text{rep}_T[x(t)] = \sum_{k=-\infty}^{\infty} x(t - kT)$$

$$\text{comb}_T[x(t)] = x(t) \sum_{k=-\infty}^{\infty} \delta(t - kT)$$

$$\text{comb}_T[x(t)] \stackrel{CTFT}{\Leftrightarrow} \frac{1}{T} \text{rep}_{\frac{1}{T}}[X(f)]$$

$$\text{rep}_T[x(t)] \stackrel{CTFT}{\Leftrightarrow} \frac{1}{T} \text{comb}_{\frac{1}{T}}[X(f)]$$

- Sampling and Reconstruction

$$y(n) = x(nT)$$

$$Y(e^{j\omega}) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X\left(\frac{\omega - 2\pi k}{2\pi T}\right)$$

$$s(t) = \sum_{k=-\infty}^{\infty} y(k)\delta(t - kT)$$

$$S(e^{j\omega}) = Y(e^{j\omega T})$$

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Problem 1.(32pt)

Consider the following 2-D LSI systems. The first system (S1) has input $x(m, n)$ and output $y(m, n)$, and the second system (S2) has input $y(m, n)$ and output $z(m, n)$.

$$y(m, n) = ay(m, n - 1) + x(m, n) \quad (\text{S1})$$

$$z(m, n) = bz(m - 1, n) + y(m, n) \quad (\text{S2})$$

The third system (S3) is formed by the composition of (S1) and (S2) with input $x(m, n)$ and output $z(m, n)$ and impulse response $h_3(m, n)$.

- a) Calculate the 2-D impulse response, $h_1(m, n)$, of the first system (S1).
- b) Calculate the 2-D impulse response, $h_2(m, n)$, of the second system (S2).
- c) Calculate the 2-D impulse response, $h_3(m, n)$, of the complete system (S3).
- d) Calculate the 2-D transfer function, $H_1(z_1, z_2)$, of the first system (S1).
- e) Calculate the 2-D transfer function, $H_3(z_1, z_2)$, of the first system (S3).

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Problem 2.(35pt)

Consider a function, $f(x, y)$, with a CSFT of $F(u, v)$, and imagine that we have two pieces of information about the function.

- First, we know its projection at angle $\theta = 0$ is given by:

$$p_0(t) = \int_{-\infty}^{\infty} f(t, y) dy .$$

- Second, we know that it is rotationally invariant so that for all θ ,

$$f(x, y) = f \left(\begin{bmatrix} \cos \theta & \sin \theta \\ -\sin \theta & \cos \theta \end{bmatrix} \begin{bmatrix} x \\ y \end{bmatrix} \right) .$$

Furthermore, define the CTFT of the projection as

$$P_0(\rho) = \int_{-\infty}^{\infty} p_0(t) e^{-j2\pi\rho t} dt .$$

- Derive an expression for $P_0(\rho)$ in terms of $F(u, v)$.
- Derive an expression for $F(u, v)$ in terms of $P_0(\rho)$.
- Is it possible to exactly reconstruct $f(x, y)$ from these two pieces of information? If so, then how? If not, then why not?
- Derive an expression for $f(x, y)$ when $p_0(t) = \text{sinc}(t)$.

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Problem 3.(32pt)

Consider the discrete-time LTI system with input $x(n)$ and output $y(n)$ given by

$$y(n) = \sum_{k=-\infty}^{\infty} h(n-k)x(k) ,$$

where

$$h(n) = T \operatorname{sinc}(nT) .$$

- a) Calculate a closed form expression for the frequency response, $H(e^{j\omega})$, of the LTI system. (Hint: You can use the sampling formula.)
- b) Sketch the function, $H(e^{j\omega})$, on the interval $[-2\pi, 2\pi]$ for $T = 1/2$. Also sketch it for $T = 1$ and $T = 3/2$.
- c) Sketch the impulse response, $h(n)$, on the interval $n = -5, \dots, +5$ for $T = 1/2$. Also sketch it for $T = 1$ and $T = 3/2$.
- d) If you would like to design a discrete-time filter with a cut-off frequency of ω_c , how should T be chosen?
- e) Calculate the $\sum_{n=-\infty}^{\infty} h(n)$ when $|T| < 1$.

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