

2-D Finite Impulse Response (FIR) Filters

- Difference equation

$$y(m, n) = \sum_{k=-N}^N \sum_{l=-N}^N h(k, l) x(m - k, n - l)$$

- For $N = 2$ - $\circ$ input points; $\times$ output point

$$\begin{array}{ccccc} \circ & \circ & \circ & \circ & \circ \\ \circ & \circ & \circ & \circ & \circ \\ \circ & \circ & \times & \circ & \circ \\ \circ & \circ & \circ & \circ & \circ \\ \circ & \circ & \circ & \circ & \circ \end{array}$$

- Number of multiplies per output point $= (2N + 1)^2$
- Transfer function

$$\begin{aligned} H(z_1, z_2) &= \sum_{k=-N}^N \sum_{l=-N}^N h(k, l) z_1^{-k} z_2^{-l} \\ H(e^{j\mu}, e^{j\nu}) &= \sum_{k=-N}^N \sum_{l=-N}^N h(k, l) e^{-j(k\mu + l\nu)} \end{aligned}$$

2-D FIR Filter Example 1

- Consider the impulse response $h(m, n) = h_1(m)h_2(n)$ where

$$4h_1(n) = (\cdots, 1, 2, 1, \cdots)$$

$$h_1(n) = (\delta(n+1) + 2\delta(n) + \delta(n-1))/4$$

Then $h(m, n)$ is a separable function with

$$16h(m, n) = \begin{array}{cccccc} \vdots & \vdots & \vdots & \vdots & \vdots & \\ & 0 & 0 & 0 & 0 & 0 \\ \cdots & 0 & 1 & 2 & 1 & 0 & \cdots \\ \cdots & 0 & 2 & 4 & 2 & 0 & \cdots \\ \cdots & 0 & 1 & 2 & 1 & 0 & \cdots \\ & 0 & 0 & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \end{array}$$

- The DTFT of $h_1(n)$ is

$$\begin{aligned} H_1(e^{j\omega}) &= e^{j\omega} + 2 + e^{-j\omega} \\ &= \frac{1}{2}(1 + \cos(\omega)) \end{aligned}$$

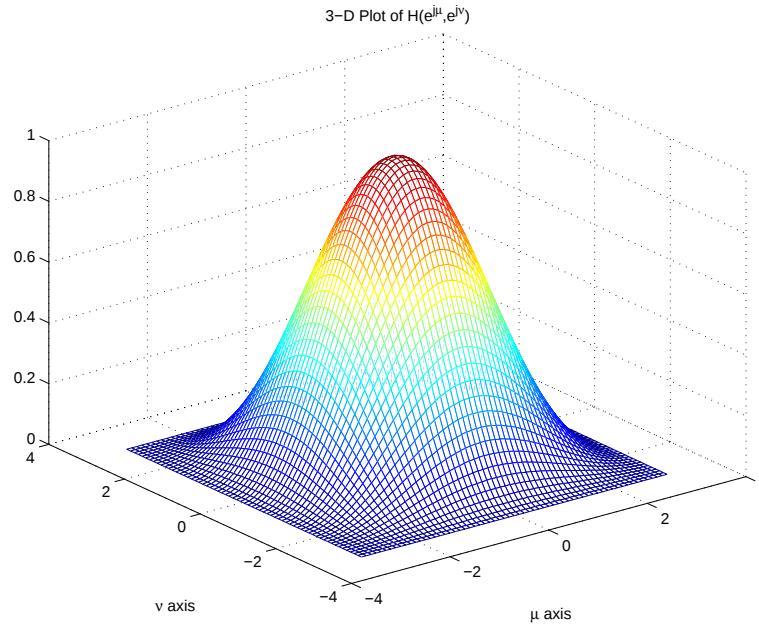
- The DSFT of $h(m, n)$ is

$$\begin{aligned} H(e^{j\mu}, e^{j\nu}) &= H_1(e^{j\mu})H_1(e^{j\nu}) \\ &= \frac{1}{4}(1 + \cos(\mu))(1 + \cos(\nu)) \end{aligned}$$

Frequency Response of 2-D FIR Filter Example 1

- Plot of frequency response

$$H(e^{j\mu}, e^{j\nu}) = \frac{1}{4} (1 + \cos(\mu)) (1 + \cos(\nu))$$



- This is a low pass filter with $H(e^{j0}, e^{j0}) = 1$

2-D FIR Filter Example 2

- Consider the impulse response $h(m, n) = h_1(m)h_2(n)$ where

$$4h_1(n) = (\cdots, 1, 2, 1, \cdots)$$

$$h_1(n) = (\delta(n+1) - 2\delta(n) + \delta(n-1))/4$$

Then $h(m, n)$ is a separable function with

$$16h(m, n) = \begin{array}{cccccc} & \vdots & \vdots & \vdots & \vdots & \vdots \\ & 0 & 0 & 0 & 0 & 0 \\ \cdots & 0 & 1 & -2 & 1 & 0 & \cdots \\ \cdots & 0 & -2 & 4 & -2 & 0 & \cdots \\ \cdots & 0 & 1 & -2 & 1 & 0 & \cdots \\ & 0 & 0 & 0 & 0 & 0 \\ & \vdots & \vdots & \vdots & \vdots & \vdots \end{array}$$

- The DTFT of $h_1(n)$ is

$$\begin{aligned} H_1(e^{j\omega}) &= e^{j\omega} + 2 + e^{-j\omega} \\ &= \frac{1}{2}(1 - \cos(\omega)) \end{aligned}$$

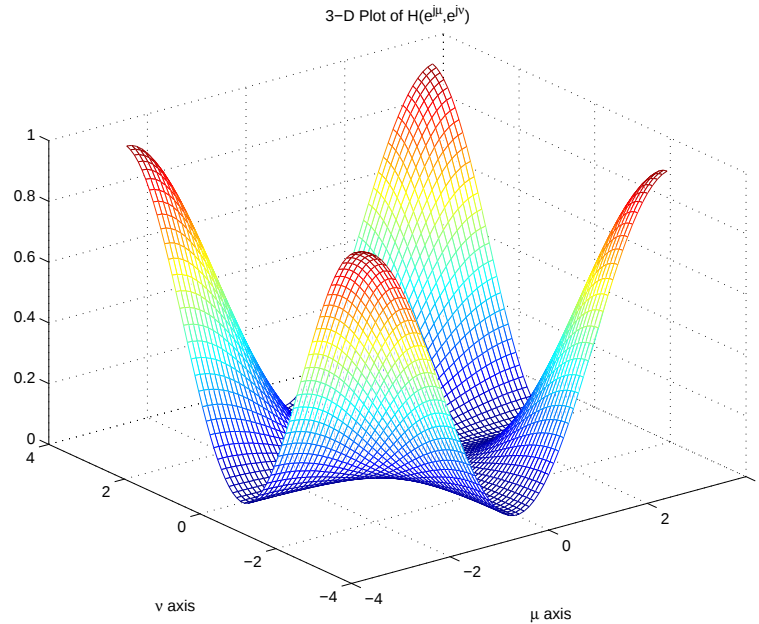
- The DSFT of $h(m, n)$ is

$$\begin{aligned} H(e^{j\mu}, e^{j\nu}) &= H_1(e^{j\mu})H_1(e^{j\nu}) \\ &= \frac{1}{4}(1 - \cos(\mu))(1 - \cos(\nu)) \end{aligned}$$

Frequency Response of 2-D FIR Filter Example 2

- Plot of frequency response

$$H(e^{j\mu}, e^{j\nu}) = \frac{1}{4} (1 - \cos(\mu)) (1 - \cos(\nu))$$



- This is a **high pass** filter with $H(e^{j0}, e^{j0}) = 0$

Ordering of Points in a Plane

- Recursive filter implementations require the ordering of points in the plane $s = (s_1, s_2)$ and $r = (r_1, r_2)$.

- Quarter plane

– For $s, r \in \mathcal{Z}^2$, then $s < r$ means $(s_2 \leq r_2)$ and $(s_1 \leq r_1)$ and $s \neq r$.

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- Symmetric half plane

– For $s, r \in \mathcal{Z}^2$, then $s < r$ means $(s_2 < r_2)$.

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- Nonsymmetric half plane

– For $s, r \in \mathcal{Z}^2$, then $s < r$ means $(s_2 < r_2)$ or $((s_2 = r_2) \text{ and } (s_1 < r_1))$.

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2-D Infinite Impulse Response (IIR) Filters

- Difference equation

$$y(m, n) = \sum_{k=-N}^N \sum_{l=-N}^N b(k, l)x(m - k, n - l) + \sum_{k=-P}^P \sum_{l=1}^P a(k, l)y(m - k, n - l) + \sum_{k=1}^P a(k, 0)y(m - k, n)$$

Simplified notation

$$y_s = \sum_r b_r x_{s-r} + \sum_{r > (0,0)} a_r y_{s-r}$$

- For nonsymmetric half plane with $N = 0$ and $P = 2$

$$\begin{array}{ccccc} \circ & \circ & \circ & \circ & \circ \\ \circ & \circ & \circ & \circ & \circ \\ \circ & \circ & \times & & \end{array}$$

- Number of multiplies per output point = $(2N + 1)^2 + (P + 2)P$
- Transfer function

$$H(z_1, z_2) = \frac{\sum_{k=-N}^N \sum_{l=-N}^N b(k, l) z_1^{-k} z_2^{-l}}{1 - \sum_{k=-P}^P \sum_{l=1}^P a(k, l) z_1^{-k} z_2^{-l} - \sum_{k=1}^P a(k, 0) z_1^{-k}}$$

$$H(e^{j\mu}, e^{j\nu}) = \frac{\sum_{k=-N}^N \sum_{l=-N}^N b(k, l) e^{-j(k\mu + l\nu)}}{1 - \sum_{k=-P}^P \sum_{l=1}^P a(k, l) e^{-j(k\mu + l\nu)} - \sum_{k=1}^P a(k, 0) e^{-j(k\mu)}}$$

2-D IIR Filter Example 1

- Consider the difference equation

$$y(m, n) = x(m, n) + ay(m-1, n) + ay(m, n-1)$$

- Spatial dependencies - $\circ$ previous value; $\times$ current value

$$\begin{array}{c} \circ \\ \circ \times \end{array}$$

- The transfer functions is then

$$\begin{aligned} H(z_1, z_2) &= \frac{1}{1 - az_1^{-1} - az_2^{-1}} \\ H(e^{j\mu}, e^{j\nu}) &= \frac{1}{1 - ae^{-j\mu} - ae^{-j\nu}} \end{aligned}$$

2-D IIR Filter Example 2

- Consider the difference equation

$$y(m, n) = x(m, n) + ay(m-1, n) + ay(m, n-1) + 2ay(m+1, n-1)$$

- Spatial dependencies - $\circ$ previous value; $\times$ current value

$$\begin{array}{cc} \circ & \circ \\ \circ & \times \end{array}$$

- The transfer functions is then

$$H(z_1, z_2) = \frac{1}{1 - az_1^{-1} - az_2^{-1} - 2az_1^{+1}z_2^{-1}}$$

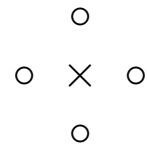
$$H(e^{j\mu}, e^{j\nu}) = \frac{1}{1 - ae^{-j\mu} - ae^{-j\mu} - 2ae^{+j\mu-j\nu}}$$

2-D IIR Filter Example 3

- Consider the difference equation

$$y(m, n) = x(m, n) + ay(m-1, n) + ay(m, n-1) + ay(m+1, n) + ay(m, n+1)$$

- Spatial dependencies - $\circ$ previous value; $\times$ current value



- Theoretically, the transfer functions is then

$$H(z_1, z_2) = \frac{1}{1 - az_1^{-1} - az_2^{-1} - az_1 - az_2}$$

$$H(e^{j\mu}, e^{j\nu}) = \frac{1}{1 - ae^{-j\mu} - ae^{-j\nu} - ae^{j\mu} - ae^{j\nu}}$$

- THIS DOESN'T WORK**