

Short-Time Synthesis

(Show subband coder structure) (Filter Bank Summation Method)

Recall

STDTFT simply breaks signal into narrowband blocks, and shifts each block down to baseband. Note that we are in time domain!! Can we put signal back together again by simply shifting each block back to where it was, and then summing?

$$\begin{aligned}
 S_r(n) &= \sum_k s(k) h(n-k) e^{-j \frac{2\pi k r}{N}} \\
 &= \sum_k s(n-k) h(k) e^{-j \frac{2\pi (n-k) r}{N}} \\
 &= \underbrace{\left\{ \sum_k s(n-k) h_r(k) \right\}}_{\text{NB BPF}} \underbrace{e^{j \frac{2\pi n r}{N}}}_{\text{downshift}}
 \end{aligned}$$

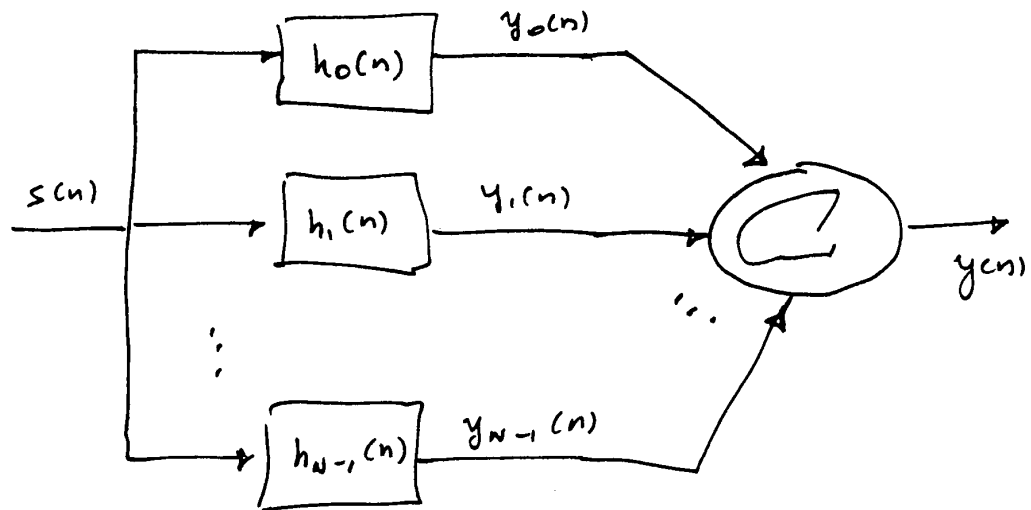
where

$$h_r(k) = h(k) e^{j \frac{2\pi k r}{N}}$$

This is just unit sample response of a bandpass filter.

$$\begin{aligned}
 \text{Define } y_r(n) &= \underbrace{e^{j \frac{2\pi n r}{N}}}_{\text{upshift}} S_r(n) \\
 &= \sum_n s(n-k) h_r(k)
 \end{aligned}$$

Now consider following system



Let's find relation between $y(n)$ and $s(n)$

$$y(n) = \sum_{l=0}^{N-1} y_l(n)$$

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$$\begin{aligned} \therefore Y(e^{j\omega}) &= \sum_{l=0}^{N-1} Y_l(e^{j\omega}) \\ &= \sum_{l=0}^{N-1} S(e^{j\omega}) H_l(e^{j\omega}) \end{aligned}$$

\therefore overall frequency response is

$$H(e^{j\omega}) = \sum_{l=0}^{N-1} H_l(e^{j\omega})$$

Now

$$H_l(e^{j\omega}) = H(e^{j(\omega - \frac{2\pi l}{N})})$$

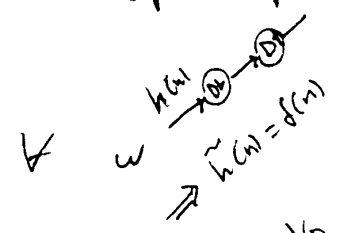
look at down-sampler followed

∴ in order for $y(n) = x(n)$

by upsampler

We must have

$$\sum_{l=0}^{N-1} H(e^{j(\omega - \frac{2\pi l}{N})}) \equiv 1$$



What does this imply about h(n)?

Consider

$$p(n) = N \sum_{k=-\infty}^{\infty} \delta(n - kN)$$

downsampling $Y(e^{j\omega}) = \frac{1}{D} \sum_{k=0}^{D-1} X(e^{j(\omega - \frac{2\pi k}{D})})$
 upsampling $X(e^{j\omega}) = Y(e^{j\omega D})$

$$P(e^{j\omega}) = N \sum_{k=-\infty}^{\infty} e^{-j\omega kN} = \frac{2\pi}{N} \sum_{k=-\infty}^{\infty} \delta(\omega - \frac{2\pi k}{N})$$

(derive from rep & comb (see following page))

Let $\tilde{h}(n) = p(n) h(n)$

$$\text{rep}_1 [s(t)] \xleftrightarrow{\text{CTFT}} \frac{1}{T} \text{comb}_T [1]$$

$$\therefore \sum_k \delta(t-k) \xleftrightarrow{\text{CTFT}} \sum_k \delta(f-k)$$

Now $\mathcal{F}\left\{ \sum_k \delta(t-k) \right\} = \sum_k e^{-j2\pi f k}$

$$\therefore \sum_k e^{-j2\pi f k} = \sum_k \delta(f-k)$$

Let $\omega N = 2\pi f$

$$\sum_k e^{-j\omega N k} = \sum_k \delta\left(\frac{\omega N}{2\pi} - k\right)$$

but $\delta(a\omega - b) = \frac{1}{|a|} \delta\left(\omega - \frac{b}{a}\right)$

$$\therefore \sum_k e^{-j\omega N k} = \frac{2\pi}{N} \sum_k \delta\left(\omega - \frac{2\pi k}{N}\right)$$

or $N \sum_k e^{-j\omega N k} = 2\pi \sum_k \delta\left(\omega - \frac{2\pi k}{N}\right)$

$$\Rightarrow \tilde{H}(e^{j\omega}) = \frac{1}{2\pi} \int_{-\pi}^{\pi} P(e^{j(\omega-\mu)}) H(e^{j\mu}) d\mu$$

$$= \sum_{k=0}^{N-1} H(e^{j\omega - \frac{2\pi k}{N}})$$

Note change in limits

So condition becomes

$$\tilde{h}(n) \equiv \delta(n)$$

i.e. $h(n) = \begin{cases} 1, & n=0 \\ 0, & n=kN \quad k \neq 0 \end{cases}$

example:

$$H(e^{j\omega}) = \text{rect}\left(\frac{\omega}{2\pi/N}\right)$$

$$h(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \text{rect}\left(\frac{\omega}{2\pi/N}\right) e^{+j n \omega} d\omega$$

$$= \frac{1}{2\pi} \int_{-\frac{\pi}{N}}^{\frac{\pi}{N}} e^{j n \omega} d\omega$$

$$= \frac{1}{2\pi} \left[\frac{e^{j n \omega}}{j n} - \frac{e^{-j n \omega}}{-j n} \right]$$

$$= \frac{1}{N} \left(\frac{1}{\pi n} \right) \sin(\pi n / N)$$

$$= \frac{1}{N} \text{sinc}\left(\frac{n}{N}\right)$$