

## Time Domain Analysis of Linear Systems

### 1. Discrete time systems

Consider the system  $y_n = T[x_n]$  Linear Time Invariant (LTI)

$$y_n = T[x_n]$$

- The output  $y_n$  is a function of the complete input  $x_n$
- We know from property 2 of DT impulses

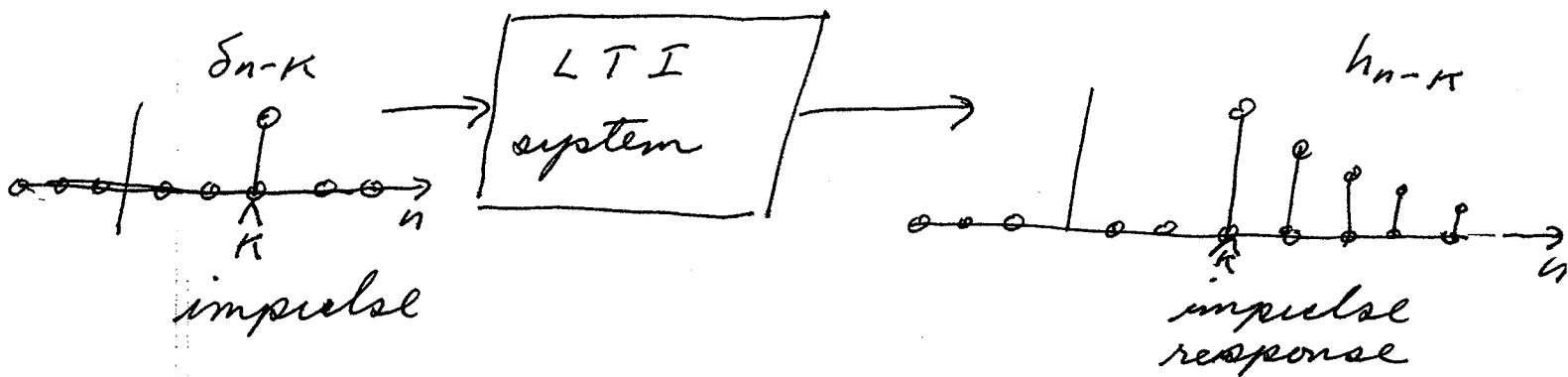
$$x_n = \sum_{k=-\infty}^{\infty} x_k \delta_{n-k}$$

Therefore

$$\begin{aligned}
 y_n &= T[x_n] \\
 &= T \left[ \sum_{k=-\infty}^{\infty} x_k \delta_{n-k} \right] \\
 &= \sum_{k=-\infty}^{\infty} T[x_k \delta_{n-k}] \\
 &= \sum_{k=-\infty}^{\infty} x_k T[\delta_{n-k}]
 \end{aligned}$$

What is  $T[\delta_{n-k}]$ ?

The response of  $S$  to an impulse at time  $k$ .



$$h_n \triangleq T[\delta_n]$$

$h_n$  is called the impulse response of the system  $S$ .

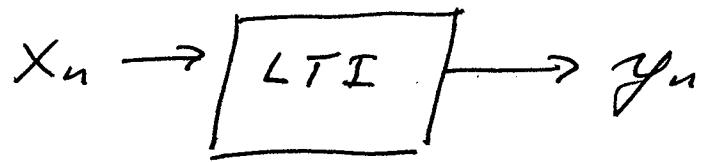
$$\text{Since } S \text{ is LTI} \Rightarrow h_{n-k} = T[\delta_{n-k}]$$

$$y_n = \sum_{k=-\infty}^{\infty} x_k T[\delta_{n-k}]$$

$$y_n = \sum_{k=-\infty}^{\infty} x_k h_{n-k}$$

By measuring  $h_n$ , we know the response of the system to any input!

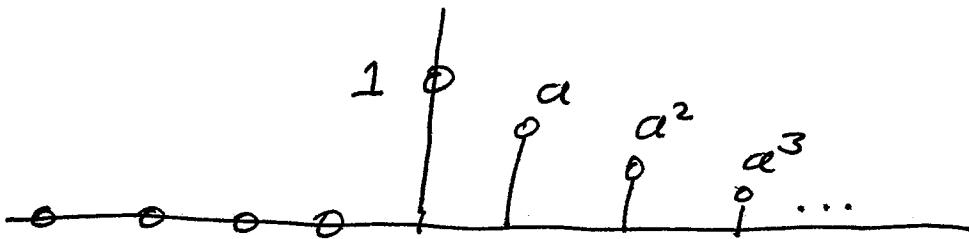
## Example



$$y_n = T[x_n]$$

$$h_n = T[\delta_n] = \alpha^n u_n \cancel{u_n}$$

where  $|\alpha| < 1$



What is the output when  $x_n = u_n$ ?  
(step response)

$$y_n = \sum_{k=-\infty}^{\infty} x_k h_{n-k}$$

$$= \sum_{k=-\infty}^{\infty} u_k (\alpha^{n-k} u_{n-k})$$

$$= \begin{cases} \sum_{k=0}^n \alpha^{n-k} & \text{for } n \geq 0 \\ 0 & \text{for } n < 0 \end{cases}$$

$$= a_n \left\{ \sum_{k=0}^n a^{n-k} \right\}$$

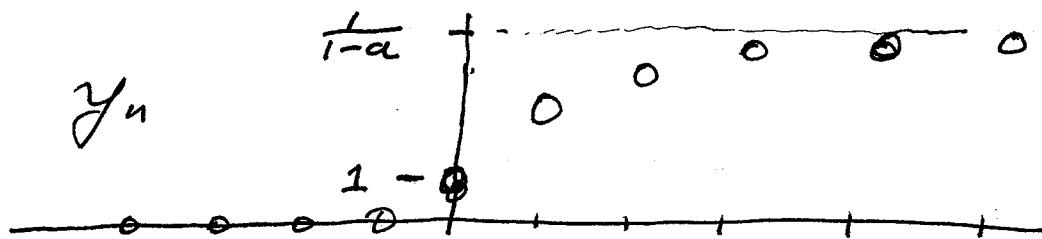
$\ell = n-k$

$$= a_n \sum_{k=0}^n a^\ell$$

$$= a_n \left( \sum_{\ell=0}^{\infty} a^\ell - \sum_{\ell=n+1}^{\infty} a^\ell \right)$$

$$= a_n \left( \frac{1}{1-a} - a^{n+1} \sum_{\ell=0}^{\infty} a^\ell \right)$$

$$= a_n \left( \frac{1-a^{n+1}}{1-a} \right)$$



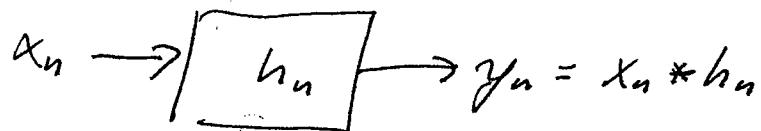
Remember

$$\sum_{\ell=0}^{n-1} a^\ell = \frac{1-a^n}{1-a}$$

We will use this a lot!

The Convolution of any two functions,  $x_n$  &  $h_n$  is given by

$$x_n * h_n = \sum_{k=-\infty}^{\infty} x_k h_{n-k}$$



property 1)  $x_n * h_n = h_n * x_n$

$$\begin{aligned} x_n * h_n &= \sum_{k=-\infty}^{\infty} x_k h_{n-k} \\ &= \sum_{u=-\infty}^{\infty} x_{n-u} h_u & u = n - k \\ &= \sum_{u=-\infty}^{\infty} h_u x_{n-u} & k = n - u \\ &= h_n * x_n \end{aligned}$$

property 2)  $(x_n * z_n) * h_n = x_n * (z_n * h_n)$

$$\left( \sum_{k=-\infty}^{\infty} x_k z_{n-k} \right) * h_n$$

$$= \sum_{u=-\infty}^{\infty} \sum_{k=-\infty}^{\infty} x_k z_{n-k} h_{n-u}$$

$$\begin{aligned} &= \sum_{k=-\infty}^{\infty} x_k \sum_{u=-\infty}^{\infty} z_{n-k} h_{n-u} & l = u - k \\ &= \sum_{k=-\infty}^{\infty} x_k \sum_{l=-\infty}^{\infty} z_l h_{(n-k)-l} & u = l + k \end{aligned}$$

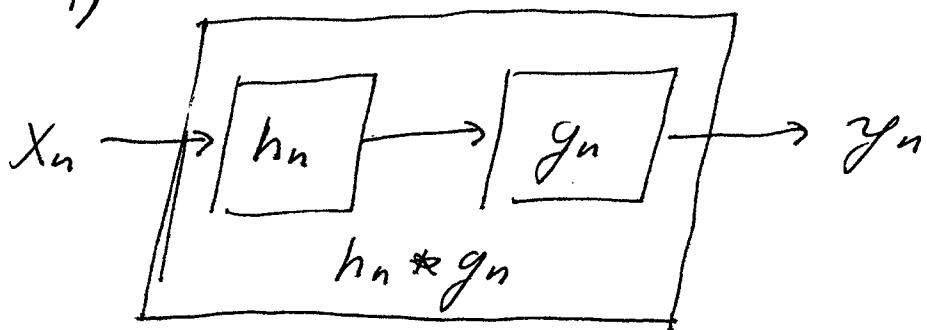
$$\begin{aligned} &= \sum_{k=-\infty}^{\infty} x_k (z_n * h_n)_{n-k} \\ &= x_n * (z_n * h_n) \end{aligned}$$

property 3)

$$x_n * h_n + x_n * g_n = x_n * (h_n + g_n)$$

Examples

1)

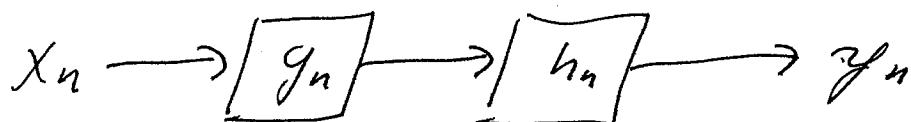
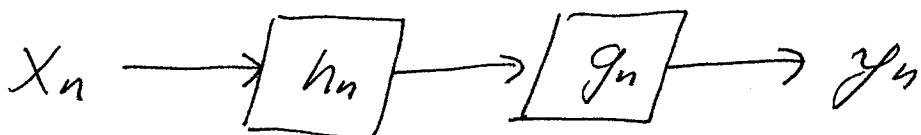


$$y_n = (x_n * h_n) * g_n$$

$$= x_n * (h_n * g_n)$$

The series connection of two LTI systems is an LTI system.

2)



$$y_n = (x_n * h_n) * g_n$$

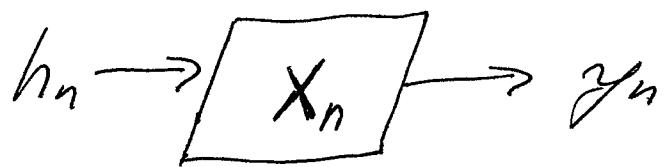
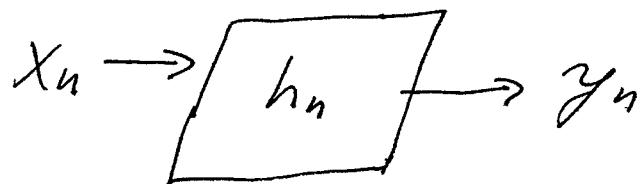
$$= x_n * (h_n * g_n)$$

$$= x_n * (g_n * h_n)$$

$$= (x_n * g_n) * h_n$$

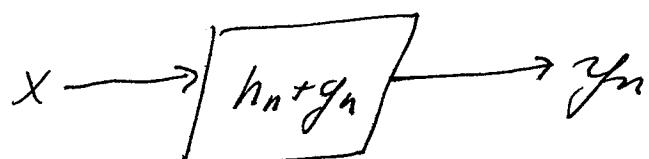
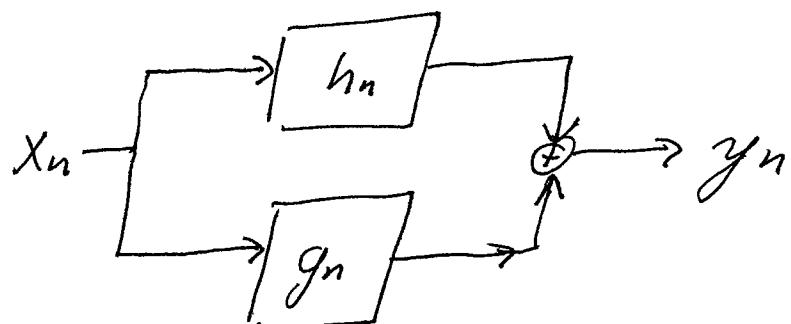
The order of LTI systems can be exchanged

3)



$$y_n = x_n * h_n = h_n * x_n !$$

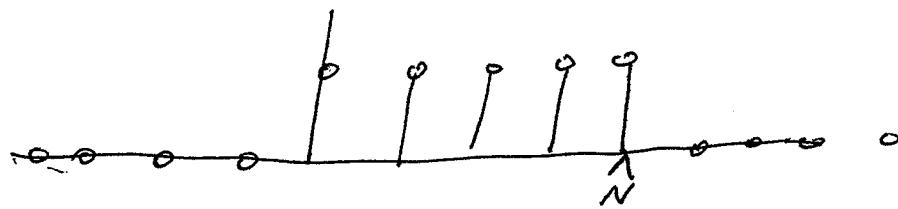
property 3)  $x_n * h_n + x_n * g_n = x_n * (h_n + g_n)$



Example :

$$y_n = x_n * h_n$$

$$h_n = \begin{cases} 1 & 0 \leq n \leq N \\ 0 & \text{otherwise} \end{cases}$$



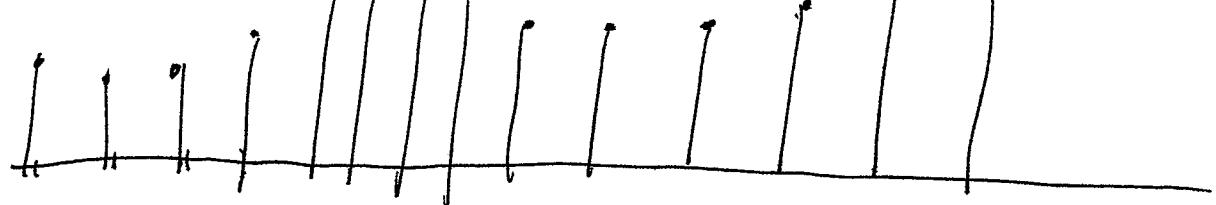
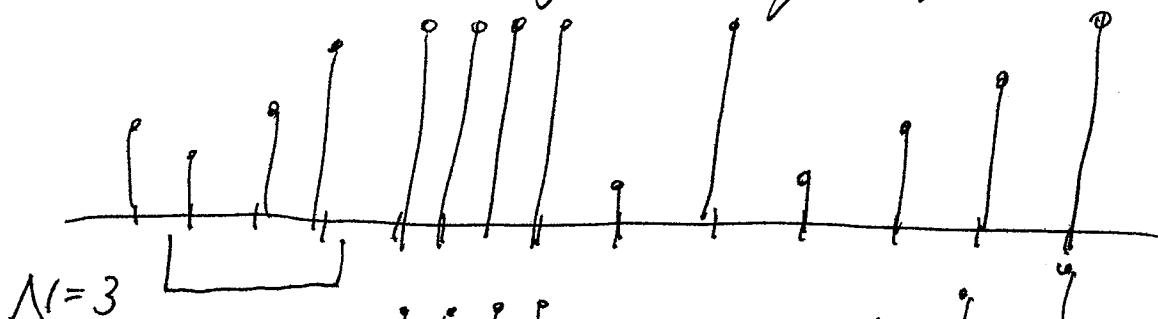
$$y_n = \sum_{k=-\infty}^{\infty} x_k h_{n-k} = x * h$$

$$= \sum_{k=-\infty}^{\infty} h_k x_{n-k} = \dots h * x$$

$$= \sum_{k=0}^N h_k x_{n-k}$$

$$y_n = \sum_{k=0}^N x_{n-k} = \sum_{k=n-N}^n x_k$$

moving average filter



$$h_n = u_n - u_{n-N-1}$$

$$y_n = x_n * h_n = x_n * u_n + x_n * (-u_{n-N-1})$$

$$\begin{aligned} x_n * u_n &= \sum_{K=-\infty}^{\infty} x_K u_{n-K} \\ &= \sum_{K=-\infty}^n x_K \quad \leftarrow \begin{array}{l} \text{integrator} \\ \text{summer} \end{array} \end{aligned}$$

$$\begin{aligned} x_n * (-u_{n-N-1}) &= -(x_n * u_n)_{n-N-1} \\ &= - \sum_{K=-\infty}^{n-N-1} x_K \end{aligned}$$

$$\begin{aligned} y_n &= \sum_{K=-\infty}^n x_K - \sum_{K=-\infty}^{n-N-1} x_K \\ &= \sum_{K=N-N}^n x_K \end{aligned}$$

Let  $T$  be a LTI continuous time system

$$y(t) = T[x(t)]$$

$$= T\left[ \underbrace{\int_{-\infty}^{\infty} x(\lambda) \delta(t-\lambda) d\lambda}_{\text{this is a sum of infinite signals}} \right]$$

$$\text{superposition} = \int_{-\infty}^{\infty} T[x(\lambda) \delta(t-\lambda)] d\lambda$$

$$\text{Homogeneity} = \int_{-\infty}^{\infty} x(\lambda) [ \delta(t-\lambda) ] d\lambda$$

$T[\delta(t-\lambda)]$  - continuous time impulse response

$$= h(t-\lambda)$$

$$y(t) = \boxed{\int_{-\infty}^{\infty} x(\lambda) h(t-\lambda) d\lambda}$$

continuous time convolution integral

$$x(t) * h(t) = \int_{-\infty}^{\infty} x(\lambda) h(t-\lambda) d\lambda$$

## Properties of CT Convolution

property 1)

$$x(t) * h(t) = h(t) * x(t)$$

property 2)

$$(x(t) * h(t)) * g(t) = x(t) * (h(t) * g(t))$$

property 3)

$$x(t) * (h(t) + g(t)) = x(t) * h(t) + x(t) * g(t)$$

## Flip and Shift

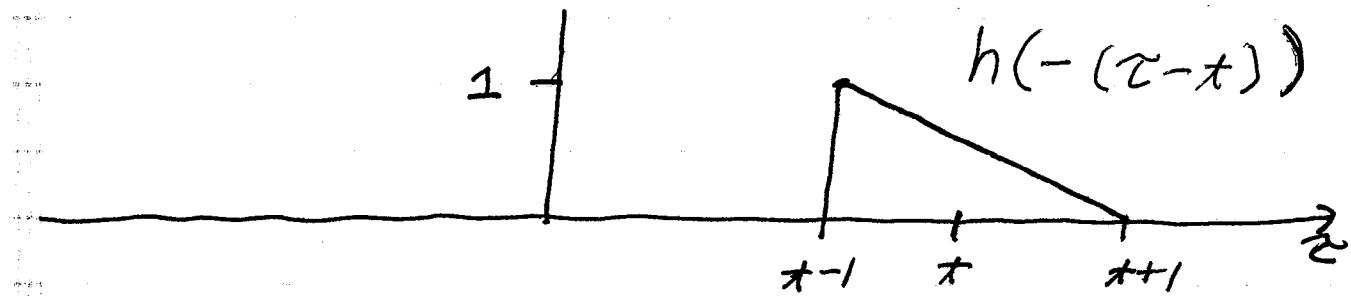
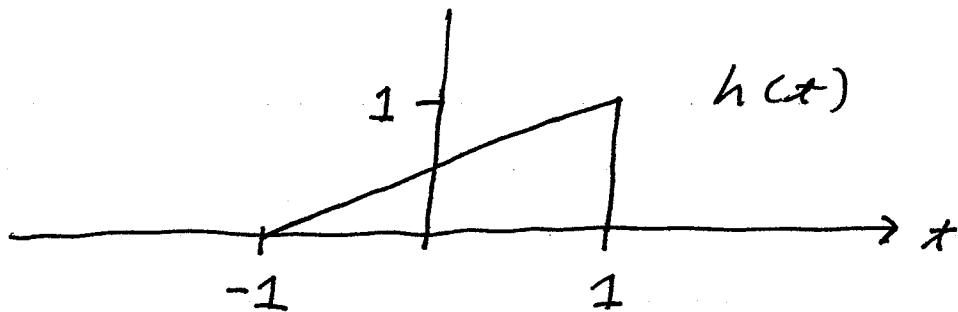
$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau$$

Think of  $x(\tau)$  and  $h(t-\tau)$  as functions of  $\underline{\tau}$

$$h(t-\tau) = h(-( \tau - t))$$

↑ flip      ↑ shift

"flip and shift" view of convolution.

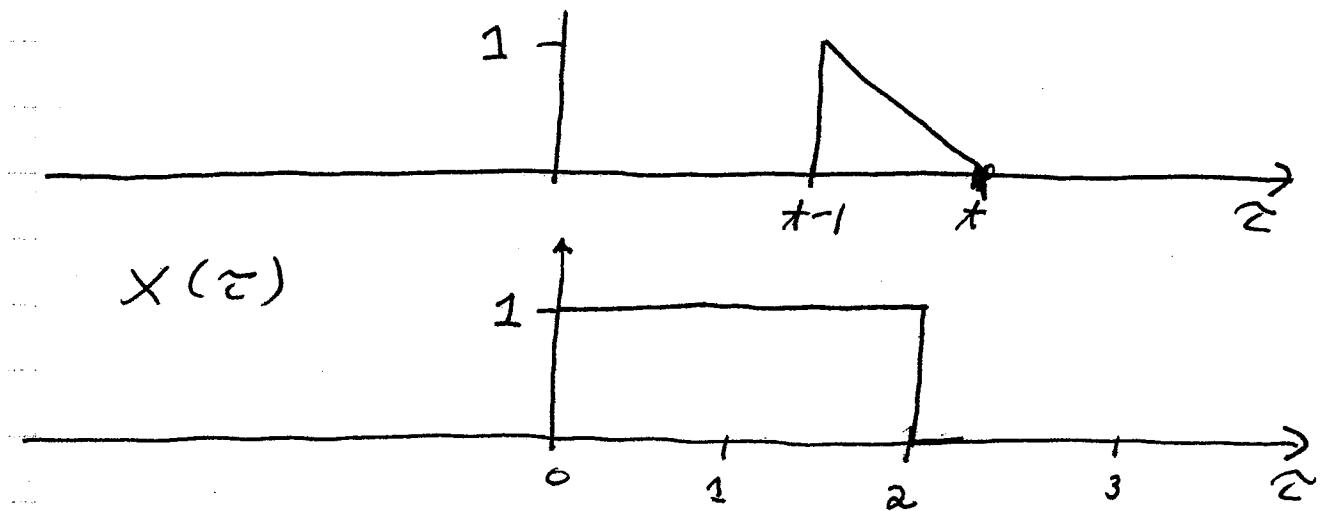


Example

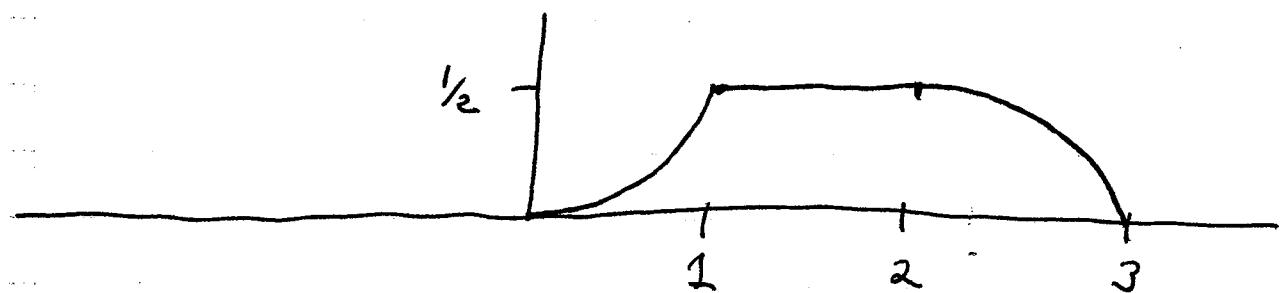
$$x(t) = u(t) - u(t-2)$$

$$h(t) = (u(t) - u(t-1)) t$$

$$h(t-\tau)$$



$$y(t) = \int x(z) h(t-z) dz$$



## Memoryless LTI System

memoryless  $\Leftrightarrow h_n = 0$  for  $n \neq 0$   
 $h(t) = 0$  for  $t \neq 0$

$$\Leftrightarrow h_n = G \delta_n$$
$$h(t) = G \delta(t)$$

for some gain factor  $G$

## Causal LTI Systems



Causal  $\Leftrightarrow h_n = 0$  for  $n < 0$   
 $h(t) = 0$  for  $t < 0$

## Stable LTI Systems

An LTI system is BIBO stable

$$\Leftrightarrow \sum_{n=-\infty}^{\infty} |h_n| < \infty$$

$$\int_{-\infty}^{\infty} |h(t)| dt < \infty$$

Proof

$$y(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$$

$$|y(t)| \leq \left| \int_{-\infty}^{\infty} x(z) h(t-z) dz \right|$$

$$\leq \int_{-\infty}^{\infty} |x(z) h(t-z)| dz$$

$$\leq M \int_{-\infty}^{\infty} |h(t-z)| dz$$

$$\leq M \int_{-\infty}^{\infty} |h(z)| dz$$

$$= M \int_{-\infty}^{\infty} |h(z)| dz = N$$

$$\int_{-\infty}^{\infty} |h(z)| dz < \infty \Rightarrow \text{BIBO stability}$$

what about  $\Leftarrow$ ? (proof by contradiction)

assume  $\int_{-\infty}^{\infty} |h(t)| dt = \infty$

choose  $x(t) = \text{sign}\{h(-t)\}$

$$= \begin{cases} 1 & h(-t) \geq 0 \\ -1 & h(-t) < 0 \end{cases}$$

$$y(t) = \int_{-\infty}^{\infty} x(z) h(t-z) dz$$

$$= \int_{-\infty}^{\infty} \text{sign}\{h(-z)\} h(t-z) dz$$

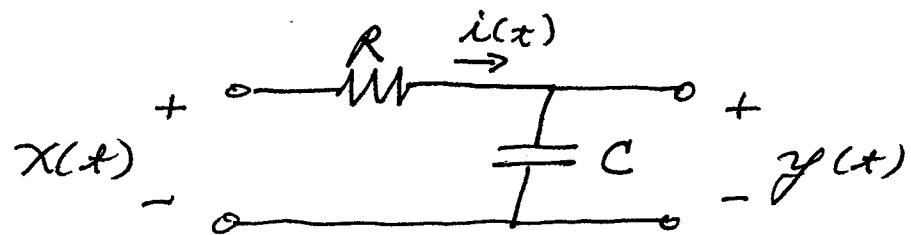
$$y(0) = \int_{-\infty}^{\infty} |h(-z)| dz = \infty$$

But since  $|x(t)| \leq 1 \Rightarrow B \not\supseteq B_0$

$$BIBO \Leftrightarrow \int_{-\infty}^{\infty} |h(t)| dt < \infty$$

## LTI Analysis of Linear differential Equations

Consider the following system



We know

$$x(t) = R i(t) + y(t)$$

and

$$i(t) = C \frac{dy(t)}{dt}$$

So we have

$$x(t) = RC \frac{dy(t)}{dt} + y(t)$$

Define  $\alpha \triangleq (RC)^{-1}$

$$\frac{dy(t)}{dt} + \alpha y(t) = \alpha x(t)$$

$$e^{\alpha t} \frac{dy(t)}{dt} + \alpha e^{\alpha t} y(t) = \alpha e^{\alpha t} x(t)$$

$$\frac{d}{dt} [e^{\alpha t} y(t)] = \alpha e^{\alpha t} x(t)$$

$$e^{\alpha t} y(t) = \int_{-\infty}^t \alpha e^{\alpha \tau} x(\tau) d\tau + K$$

$K$  is a constant

$$\begin{aligned} y(t) &= \int_{-\infty}^t \alpha e^{\alpha(\tau-t)} x(\tau) d\tau + K \\ &= \int_{-\infty}^t \alpha e^{-\alpha(t-\tau)} x(\tau) d\tau + K \end{aligned}$$

If the circuit starts at "rest",  
then

$$\lim_{t \rightarrow -\infty} y(t) = 0$$

$$\lim_{t \rightarrow -\infty} y(t) = \lim_{t \rightarrow -\infty} \int_{-\infty}^t \alpha e^{-\alpha(t-\tau)} x(\tau) d\tau + K$$

$$0 = 0 + k$$

$$\Rightarrow k = 0$$

So we have

$$y(t) = \int_{-\infty}^t \alpha e^{-\alpha(t-z)} x(z) dz$$

$$y(t) = \int_{-\infty}^{\infty} \alpha e^{-\alpha(t-z)} u(t-z) x(z) dz$$

$$y(t) = \int_{-\infty}^{\infty} x(z) \underbrace{\alpha e^{-\alpha(t-z)} u(t-z) dz}_{h(t-z)}$$

Comments

1) This is an LTI system

2)

$$y(t) = x(t) * h(t)$$

Impulse response

3)

$$h(t) = \alpha e^{-\alpha t} u(t)$$

$$= \alpha e^{-\frac{1}{k_0} t} u(t)$$