

Signals

Represented by a function of one or more independent variables

$$f(t) \quad 0 \leq t \leq \infty$$

↑ independent variable

$$x(k) \quad k = 0, 1, \dots, \infty \quad \text{also written } x_k$$

Continuous time ^{signal} - continuously valued independent variable

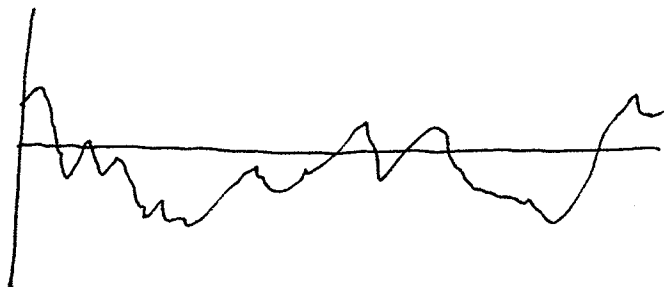
Discrete time ^{signal} - Discrete (integer) valued independent variable

Digital signal - Discrete time signal with discrete values.

warning: continuous time ~~(*)~~ continuous function

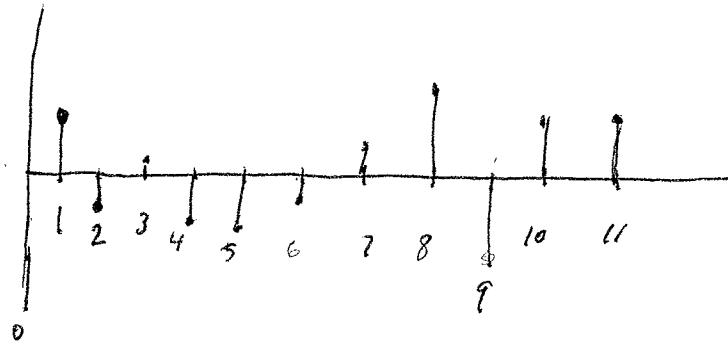
Examples:

1. Audio signal (e.g. speech, music) - continuous time, single indep. variable

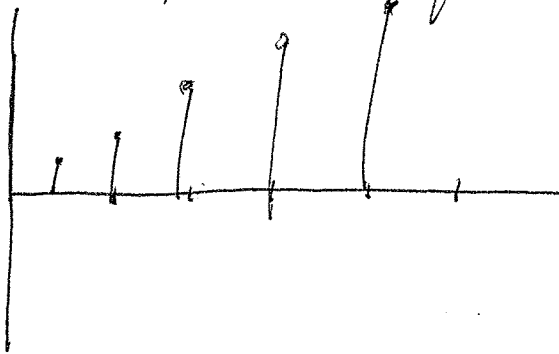


audio

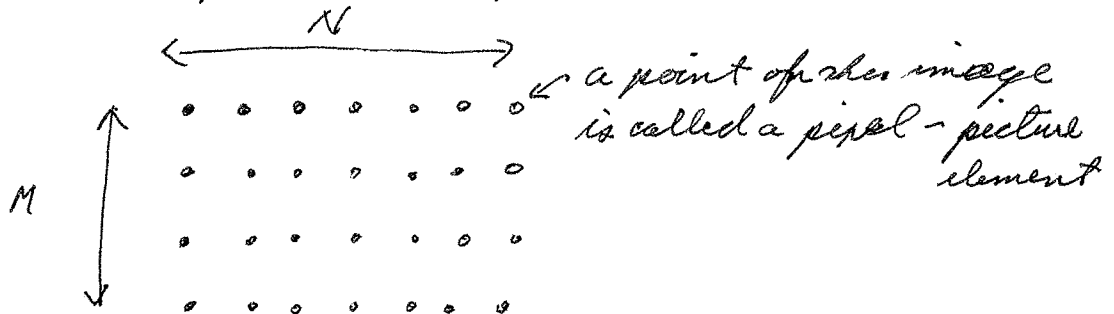
2. Digital Recording (e.g. CD, DAT)
- discrete time, single indep variable,
discrete value



3. Annual Trade Deficits - discrete time



4. Computer Displayed Image -
discrete space, two parameter

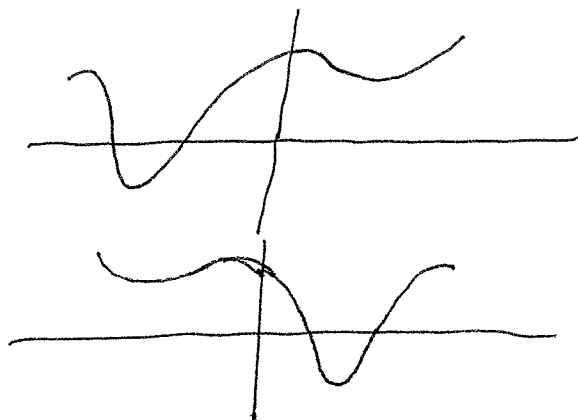


Signal Transformations

Time reversal

$$X(t) \rightarrow X(-t)$$

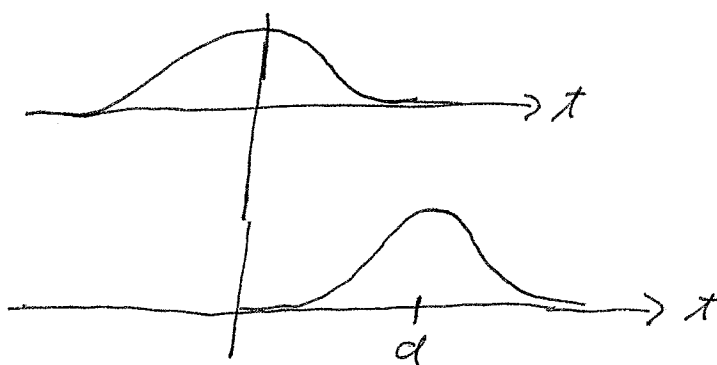
$$X_k \rightarrow X_{-k}$$



Time delay

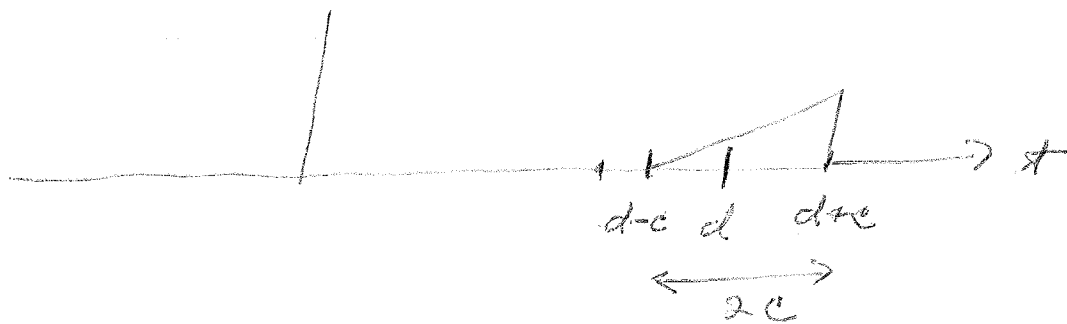
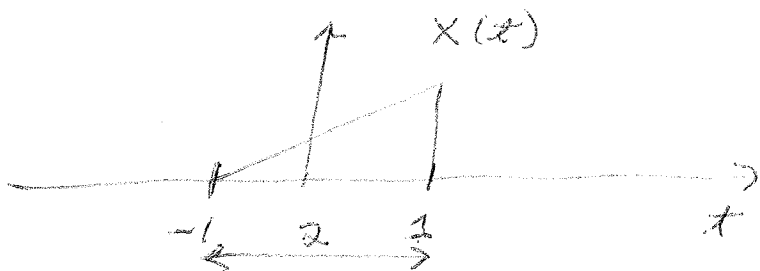
$$X(t) \rightarrow X(t-d)$$

$$X_k \rightarrow X_{k-m}$$



Time Scaling

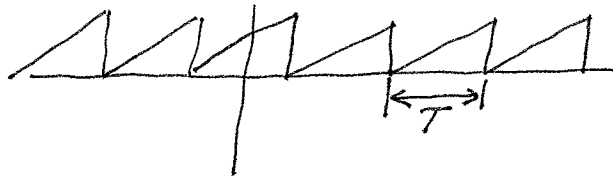
$$x(t) \rightarrow x\left(\frac{t-d}{c}\right)$$



periodic signals

$X(t)$ is periodic with period T if

$$X(t) = X(t+T)$$



$$\begin{aligned} \text{Since } X(t) &= X(t+T) & u &= t+T \\ &= X(u) \\ &= X(u+T) \\ &= X(t+2T) \end{aligned}$$

$$\Rightarrow X(t) = X(t+KT) \quad K \text{ an integer}$$

Therefore $X(t)$ is periodic with period KT for all positive integers K .

Definition: The fundamental period of $X(t)$ is the smallest positive value of T .

Discrete case

$$X_k = X_{k+N} \Rightarrow \text{periodic with period } N$$

Fundamental period is the smallest value of N .

Even/Odd Signals

Even signal

$$X(t) = X(-t) \quad \text{Continuous time}$$

$$X(n) = X(-n) \quad \text{Discrete time}$$

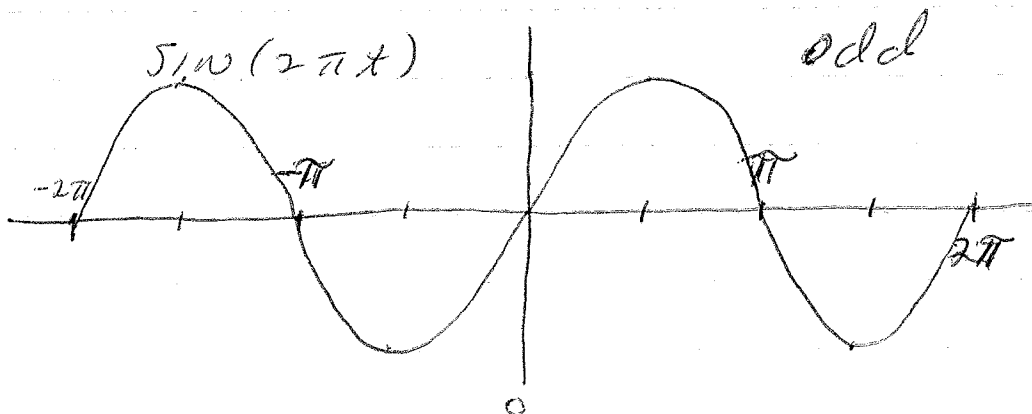
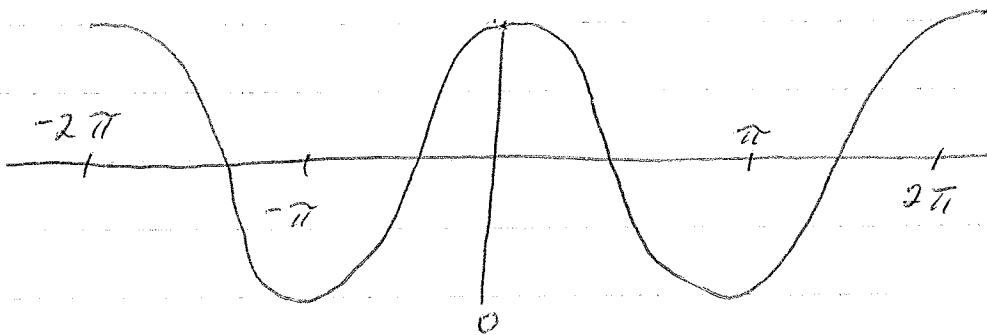
Odd Signal

$$X(t) = -X(-t) \quad \text{Continuous time}$$

$$X(n) = -X(-n) \quad \text{Discrete Time}$$

examples:

$$\cos(2\pi x) \quad \text{even}$$



Energy/Power

Energy

$$E = \int_{-\infty}^{\infty} (x(t))^2 dt \quad \text{CT}$$

$$E = \sum_{n=-\infty}^{\infty} (x(n))^2 \quad \text{DT}$$

Power

$$P = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T (x(t))^2 dt \quad \text{CT}$$

$$P = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N (x(n))^2 \quad \text{DT}$$

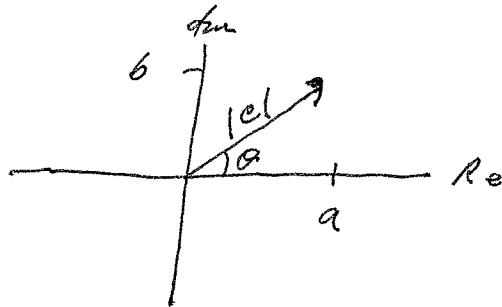
Average

$$\text{Ave} = \lim_{T \rightarrow \infty} \frac{1}{2T} \int_{-T}^T x(t) dt \quad \text{CT}$$

$$\text{Ave} = \lim_{N \rightarrow \infty} \frac{1}{2N+1} \sum_{n=-N}^N x(n)$$

Exponential Signals: Continuous time:

example 1) $x(t) = c e^{\delta t}$ $\delta = \alpha + j\omega$
 $c = a + ib$



$$c = a + ib$$

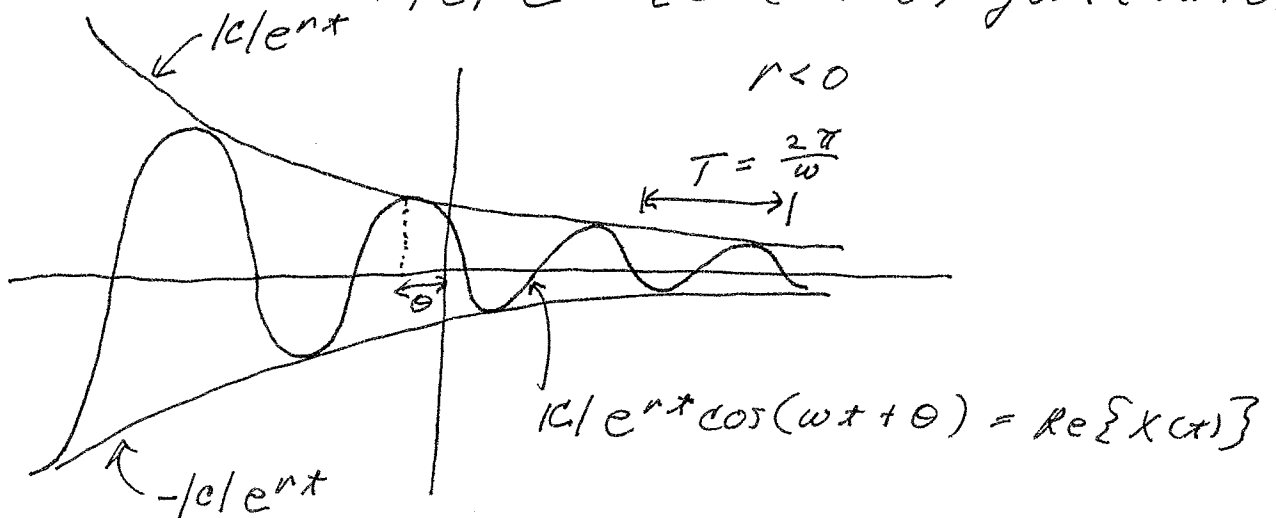
$$= |c| \cos \theta + j |c| \sin \theta$$

$$= |c| e^{j\theta}$$

$$x(t) = |c| e^{j\theta} e^{\delta t}$$

$$= |c| e^{\alpha t} e^{j(\omega t + \theta)}$$

$$= |c| e^{\alpha t} \{ \cos(\omega t + \theta) + j \sin(\omega t + \theta) \}$$



$$\cos(\omega(x+T) + \theta) = \cos(\omega x + \theta)$$

$$\Rightarrow \omega T = 2\pi$$

$$T = \frac{2\pi}{\omega}$$

example 2) $X_k = C \alpha^k$

$$C = a + jb = |C| e^{j\theta}$$

$$\alpha = e^s = e^n e^{j\omega}$$

$$\begin{aligned} X_k &= |C| e^{j\theta} (e^n e^{j\omega})^k \\ &= |C| e^{nk} e^{j(\omega k + \theta)} \end{aligned}$$

Discrete Time periodic exponentials

$$X_k = e^{j\omega k}$$

Is X_k periodic?

$$e^{j\omega(k+N)} \stackrel{?}{=} e^{j\omega k}$$

$\Leftrightarrow \omega N = m 2\pi$ for m an integer

periodic $\Leftrightarrow \omega = \frac{m}{N} 2\pi$
where N is the period

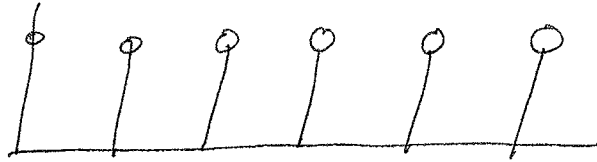
Let $\hat{\omega} = \omega + 2\pi$

$$\begin{aligned} e^{j\hat{\omega}k} &= e^{j\omega k + j2\pi k} \\ &= e^{j\omega k} e^{j2\pi k} \\ &= e^{j\omega k} \end{aligned}$$

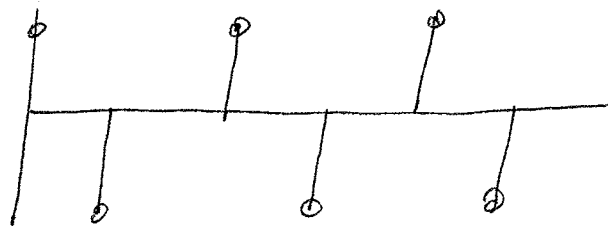
Frequencies of ω and $\omega + 2\pi$ are equivalent

examples:

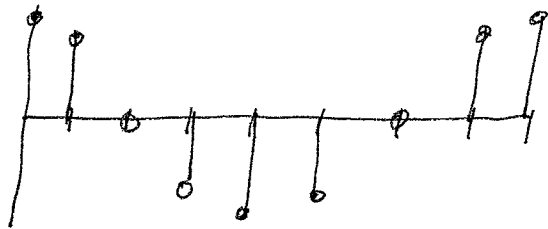
$$\left. \begin{array}{l} \omega = 0 \\ \omega = 2\pi \end{array} \right\} \Rightarrow e^{j\omega K} = 1$$



$$\omega = \pi \Rightarrow e^{j\omega K} = (-1)^K$$



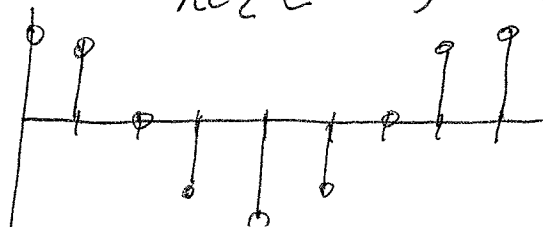
$$\omega = \frac{1}{8}2\pi \quad \text{Re}\{e^{j\omega K}\} = \cos(\omega K)$$

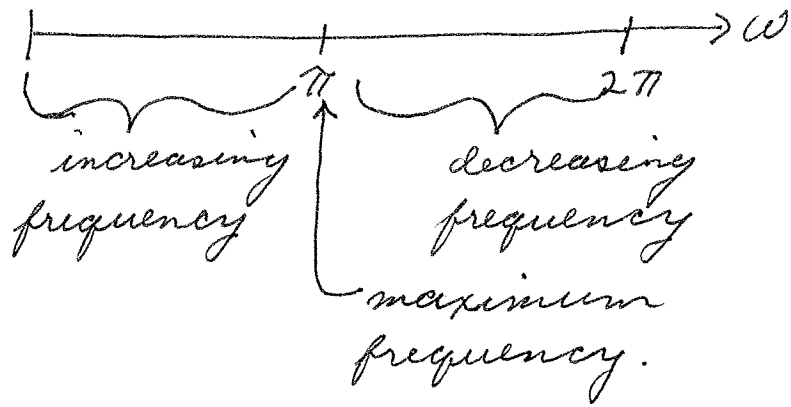


$$\begin{aligned} \omega = \frac{7}{8}2\pi &\Leftrightarrow \hat{\omega} = \frac{7}{8}2\pi - 2\pi \\ &= -\frac{1}{8}2\pi \end{aligned}$$

$$e^{j\omega K} = e^{j\hat{\omega} K} = e^{-\frac{1}{8}2\pi K}$$

$$\text{Re}\{e^{j\omega K}\} = \cos\left(\frac{1}{8}2\pi K\right)$$

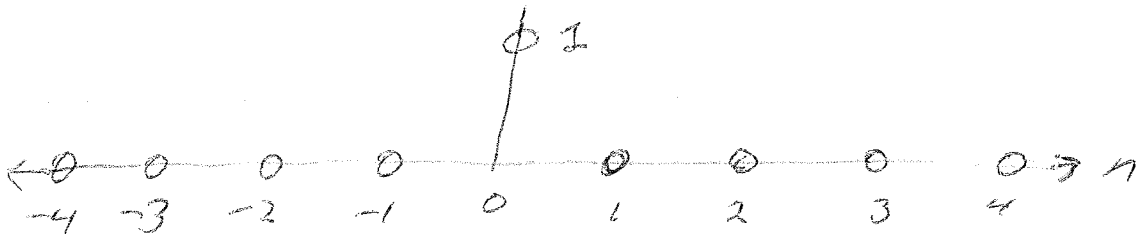




Impulse and Step Functions

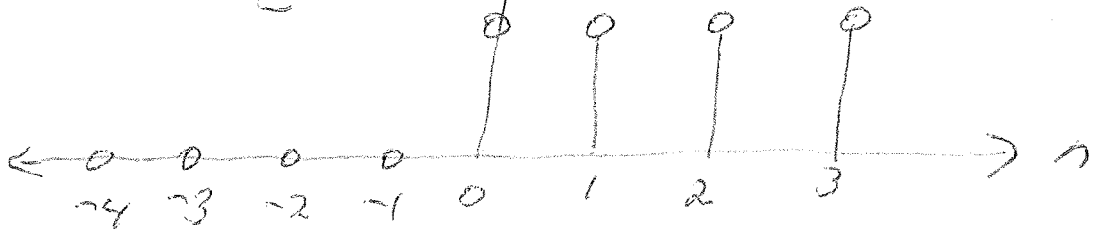
1. Discrete time Impulse

$$\delta_n = \begin{cases} 1 & \text{if } n=0 \\ 0 & \text{otherwise} \end{cases}$$



2. Discrete time step functions

$$u_n = \begin{cases} 1 & \text{if } n \geq 0 \\ 0 & \text{otherwise} \end{cases}$$



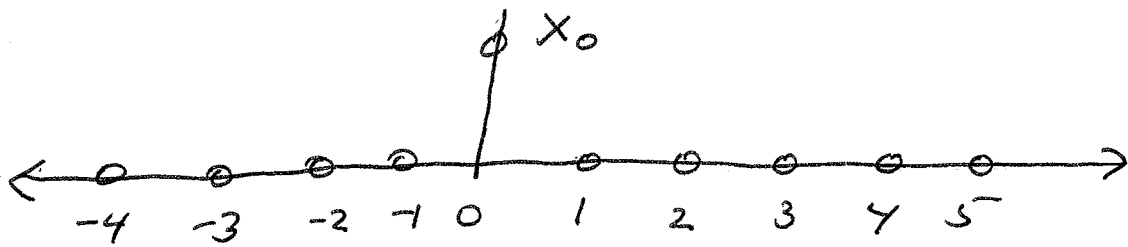
• Relationship between δ_n and u_n

$$\delta_n = u_n - u_{n-1}$$

n	$u_n - u_{n-1}$
-1	$0 - 0 = 0$
0	$1 - 0 = 1$
1	$1 - 1 = 0$
2	$1 - 1 = 0$

• Multiplication by DT impulses

$$X_n \delta_n = \begin{cases} X_0 & n=0 \\ 0 & \text{otherwise} \end{cases}$$



$$\sum_{n=-\infty}^{\infty} X_n \delta_n = X_0$$

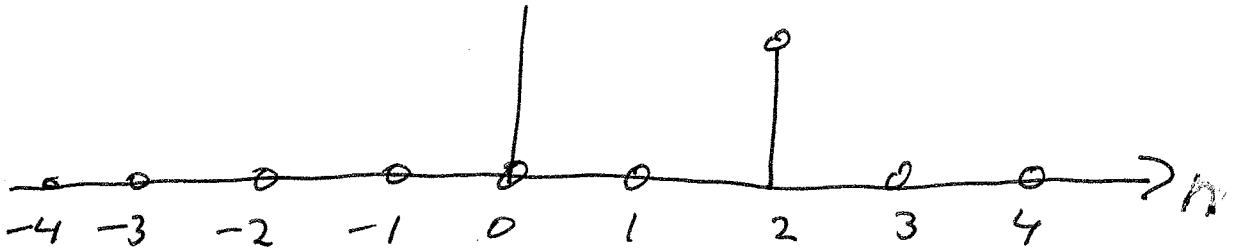
property 1

$$\sum_{n=-\infty}^{\infty} X_n \delta_n = X_0$$

Representation of DT signals with DT impulses

$$\delta_{n-k} = \begin{cases} 1 & \text{for } n=k \\ 0 & \text{otherwise} \end{cases}$$

$$\delta_{n-2}$$



$$\text{So } X_n \delta_{n-k} = \begin{cases} X_k & \text{for } n=k \\ 0 & \text{otherwise} \end{cases}$$

Therefore

$$\begin{aligned} \sum_{n=-\infty}^{\infty} X_n \delta_{n-k} &= \dots + 0 + 0 + X_k + 0 + 0 + \dots \\ &= X_k \end{aligned}$$

So we know that

$$\sum_{n=-\infty}^{\infty} X_n \delta_{n-k} = X_k$$

Since δ_n is a symmetric function,
we know that $\delta_{n-k} = \delta_{k-n}$.
Therefore,

$$\sum_{n=-\infty}^{\infty} X_n \delta_{k-n} = X_k$$

Substituting variables $\left. \begin{array}{l} \kappa \rightarrow n \\ n \rightarrow \kappa \end{array} \right\}$,

we have

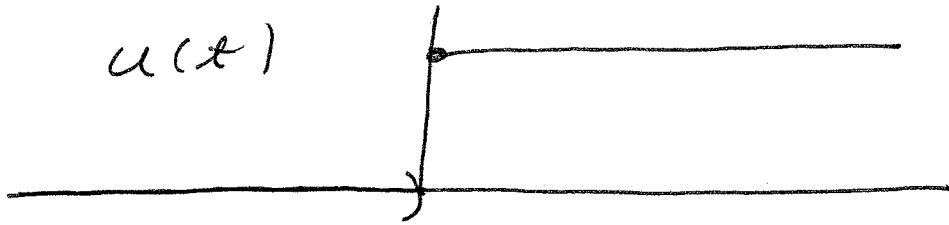
$$\sum_{\kappa=-\infty}^{\infty} X_{\kappa} \delta_{n-\kappa} = X_n$$

property 2

$$\sum_{\kappa=-\infty}^{\infty} X_{\kappa} \delta_{n-\kappa} = X_n$$

Continuous time step function

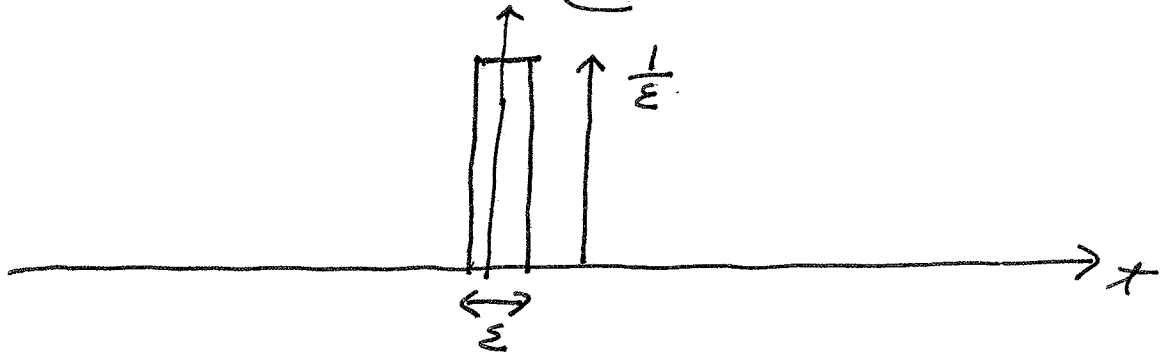
$$u(t) = \begin{cases} 1 & \text{for } t \geq 0 \\ 0 & \text{otherwise} \end{cases}$$



• Continuous time Impulse

- consider the function

$$\delta_{\epsilon}(t) = \begin{cases} \frac{1}{\epsilon} & \text{for } |t| < \frac{\epsilon}{2} \\ 0 & \text{otherwise} \end{cases}$$



Comments

- As $\epsilon \rightarrow 0$ spike becomes large

- ... But area remains constant

$$\int_{-\infty}^{\infty} \delta_{\epsilon}(t) dt = \int_{-\epsilon/2}^{\epsilon/2} \frac{1}{\epsilon} dt = 1$$

Consider

$$\lim_{\epsilon \rightarrow 0} \int_{-\infty}^{\infty} \delta_{\epsilon}(t) X(t) dt$$

$$= \lim_{\epsilon \rightarrow 0} \int_{-\epsilon/2}^{\epsilon/2} X(t) \frac{1}{\epsilon} dt = X(0)$$

(assumes that $X(t)$ is continuous)

Definition: Delta function (Not rigorous!)

$$\delta(x) \stackrel{\Delta}{=} \lim_{\epsilon \rightarrow 0} \delta_{\epsilon}(x)$$

property 1

$$\int_{-\infty}^{\infty} x(t) \delta(t) dt = x(0)$$

In order to derive property 2, let

$$x(t) \stackrel{\Delta}{=} y(d-t)$$

then we have (by prop 1)

$$\int_{-\infty}^{\infty} y(d-t) \delta(t) dt = y(d-0)$$

Make the change of variables

$$z \stackrel{\Delta}{=} d-t \Rightarrow t = d-z$$

then

$$\int_{-\infty}^{\infty} y(z) \delta(d-z) dz = y(d)$$

Substituting t for τ yields

property 2

$$\int_{-\infty}^{\infty} y(\tau) \delta(t-\tau) d\tau = y(t)$$

This is the decomposition of signal into impulses.