

EE 301 Midterm Exam #2

October 23, Fall 2009

Name: Key

Instructions:

- Follow all instructions carefully!
- This is a 50 minute exam containing **four** problems totaling 100 points.
- You **may not** use a calculator.
- You **may not** use any notes, books or other references.
- You may only keep pencils, pens and erasers at your desk during the exam.

Good Luck.

Fact Sheet

- Function definitions

$$\text{rect}(t) \triangleq \begin{cases} 1 & \text{for } |t| < 1/2 \\ 0 & \text{otherwise} \end{cases}$$

$$\Lambda(t) \triangleq \begin{cases} 1 - |t| & \text{for } |t| < 1 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{sinc}(t) \triangleq \frac{\sin(\pi t)}{\pi t}$$

- Continuous Time Fourier Series (CTFS)

$$a_k = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\frac{2\pi}{T}t} dt$$

$$x(t) = \sum_{k=-\infty}^{\infty} a_k e^{jk\frac{2\pi}{T}t}$$

- CTFS Properties

$$x(t - t_o) \stackrel{\text{CTFS}}{\Leftrightarrow} a_k e^{-jk\frac{2\pi}{T}t_o}$$

For $x(t)$ real valued $a_k = a_{-k}^*$

$$\frac{dx(t)}{dt} \stackrel{\text{CTFS}}{\Leftrightarrow} jk \frac{2\pi}{T} a_k$$

$$x(t) = \int_{-T/2}^{T/2} x(\tau) y(t - \tau) d\tau \stackrel{\text{CTFS}}{\Leftrightarrow} T a_k b_k$$

$$x(t)y(t) \stackrel{\text{CTFS}}{\Leftrightarrow} \sum_{l=-\infty}^{\infty} a_l b_{k-l}$$

$$\frac{1}{T} \int_{-T/2}^{T/2} |x(t)|^2 dt = \sum_{k=-\infty}^{\infty} |a_k|^2$$

- CTFT

$$X(\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(\omega) e^{j\omega t} d\omega$$

- CTFT Properties

$$x(-t) \stackrel{\text{CTFT}}{\Leftrightarrow} X(-\omega)$$

$$x(t - t_0) \stackrel{\text{CTFT}}{\Leftrightarrow} X(\omega) e^{-j\omega t_0}$$

$$x(at) \stackrel{\text{CTFT}}{\Leftrightarrow} \frac{1}{|a|} X(\omega/a)$$

$$X(t) \stackrel{\text{CTFT}}{\Leftrightarrow} 2\pi x(-\omega)$$

$$x(t) e^{j\omega_0 t} \stackrel{\text{CTFT}}{\Leftrightarrow} X(\omega - \omega_0)$$

$$x(t)y(t) \stackrel{\text{CTFT}}{\Leftrightarrow} \frac{1}{2\pi} X(\omega) * Y(\omega)$$

$$x(t) * y(t) \stackrel{\text{CTFT}}{\Leftrightarrow} X(\omega)Y(\omega)$$

$$\frac{dx(t)}{dt} \stackrel{\text{CTFT}}{\Leftrightarrow} j\omega X(\omega)$$

$$\int_{-\infty}^{\infty} |x(t)|^2 dt = \frac{1}{2\pi} \int_{-\infty}^{\infty} |X(\omega)|^2 d\omega$$

If $x(t) \stackrel{\text{CTFS}}{\Leftrightarrow} a_k$ then

$$x(t) \stackrel{\text{CTFT}}{\Leftrightarrow} \sum_{k=-\infty}^{\infty} 2\pi a_k \delta(\omega - 2\pi k/T)$$

- CTFT pairs

$$\text{sinc}(t) \stackrel{\text{CTFT}}{\Leftrightarrow} \text{rect}(\omega/(2\pi))$$

$$\text{rect}(t) \stackrel{\text{CTFT}}{\Leftrightarrow} \text{sinc}(\omega/(2\pi))$$

For $a > 0$

$$\frac{1}{(n-1)!} t^{n-1} e^{-at} u(t) \stackrel{\text{CTFT}}{\Leftrightarrow} \frac{1}{(j\omega + a)^n}$$

$$\sum_{k=-\infty}^{\infty} \delta(t - kT) \stackrel{\text{CTFT}}{\Leftrightarrow} \frac{1}{T} \sum_{k=-\infty}^{\infty} 2\pi \delta(\omega - 2\pi k/T)$$

- DFT

$$X_k = \frac{1}{N} \sum_{n=0}^{N-1} x(n) e^{-j2\pi kn/N}$$

$$x(n) = \sum_{k=0}^{N-1} X_k e^{j2\pi kn/N}$$

- DTFT

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n}$$

$$x(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega) e^{j\omega n} d\omega$$

- DTFT pairs

$$a^n u(n) \stackrel{\text{DTFT}}{\Leftrightarrow} \frac{1}{1 - ae^{-j\omega}}$$

- Sampling and Reconstruction

$$y(n) = x(nT)$$

$$Y(\omega) = \frac{1}{T} \sum_{k=-\infty}^{\infty} X\left(\frac{\omega - 2\pi k}{T}\right)$$

$$s(t) = \sum_{k=-\infty}^{\infty} y(k) \delta(t - kT)$$

$$S(\omega) = Y(\omega T)$$

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Problem 1. (25pt) Sine wave inputs to LTI systems

Let $y(t) = S[x(t)]$ be an LTI system with input $x(t)$, output $y(t)$, and real valued impulse response $h(t)$. Determine the output, $y(t)$, in the following cases.

- a) $x(t) = e^{j\omega_0 t}$.
- b) $x(t) = e^{-j\omega_0 t}$.
- c) $x(t) = \cos(\omega_0 t)$.
- d) $x(t) = B \cos(\omega_0 t + \phi)$, where B and ϕ are real values.

$$\begin{aligned}
 a) \quad y(t) &= \int_{-\infty}^{\infty} h(z) x(t-z) dz \\
 &= \int_{-\infty}^{\infty} h(z) e^{j\omega_0(t-z)} dz \\
 &= e^{j\omega_0 t} \underbrace{\int_{-\infty}^{\infty} h(z) e^{-j\omega_0 z} dz}_c
 \end{aligned}$$

$$y(t) = C e^{j\omega_0 t}$$

$$\begin{aligned}
 b) \quad y(t) &= C^* e^{-j\omega_0 t} \\
 &= \int_{-\infty}^{\infty} h(z) e^{-j(\omega_0)z} dz = \left[\int_{-\infty}^{\infty} h(z) e^{j\omega_0 z} dz \right]^*
 \end{aligned}$$

$$c) \quad x(t) = \frac{1}{2} [e^{j\omega_0 t} + e^{-j\omega_0 t}] \Rightarrow C^*$$

$$y(t) = \frac{1}{2} [C e^{j\omega_0 t} + C^* e^{-j\omega_0 t}] \quad \text{Let } C \stackrel{*}{=} A e^{j\theta}$$

$$= \frac{1}{2} A [e^{j(\omega_0 t + \theta)} + e^{-j(\omega_0 t - \theta)}]$$

$$= A \cos(\omega_0 t + \theta)$$

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d) By Linearity

$$\begin{aligned} S[B \cos(\omega_0 t)] &= B S[\cos(\omega_0 t)] \\ &= BA \cos(\omega_0 t + \theta) \end{aligned}$$

By time invariance

$$\begin{aligned} S[B \cos(\omega_0 t + \phi)] &= S[B \cos(\omega_0(t + \frac{\phi}{\omega_0}))] \\ &= AB \cos(\omega_0(t + \frac{\phi}{\omega_0}) + \theta) \\ &= AB \cos(\omega_0 t + \theta + \phi) \end{aligned}$$

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Problem 2. (25pt) *The continuous time Fourier series*

For each example, $x(t)$ is a period signal with period T and associated frequency $\omega_0 = \frac{2\pi}{T}$.

a) Calculate the Fourier series coefficients, a_k , for the signal, $x(t) = e^{j2\omega_0 t} + e^{j5\omega_0 t}$.

b) Calculate the Fourier series coefficients, a_k , for the signal, $x(t) = \cos(2\omega_0 t)$.

c) Calculate the Fourier series coefficients, a_k , for the signal, $x(t) = B \cos(2\omega_0 t + \phi)$.

d) Calculate a closed form expression for the signal, $x(t)$, when $x(t) = \sum_{k=-\infty}^{\infty} e^{jk\omega_0 t}$.

$$a) \quad a_k = \begin{cases} 1 & \text{if } k=2 \\ 1 & \text{if } k=5 \\ 0 & \text{otherwise} \end{cases} \quad a_k = \delta(k-2) + \delta(k-5)$$

$$b) \quad a_k = \begin{cases} \frac{1}{2} & \text{if } k=2 \\ \frac{1}{2} & \text{if } k=-2 \\ 0 & \text{otherwise} \end{cases} \quad a_k = \frac{1}{2}(\delta(k-2) + \delta(k+2))$$

$$c) \quad x(t) = \frac{B}{2} (e^{j(2\omega_0 t + \phi)} + e^{-j(2\omega_0 t + \phi)})$$

$$= \frac{B}{2} e^{j\phi} e^{j2\omega_0 t} + \frac{B}{2} e^{-j\phi} e^{-j2\omega_0 t}$$

$$a_k = \begin{cases} \frac{B}{2} e^{j\phi} & \text{if } k=2 \\ \frac{B}{2} e^{-j\phi} & \text{if } k=-2 \\ 0 & \text{otherwise} \end{cases}$$

$$a_k = \frac{B}{2} e^{j\phi} \delta(k-2) + \frac{B}{2} e^{-j\phi} \delta(k+2)$$

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d) We know that $x(t) = \delta(t)$ $|t| < \frac{T}{2}$

$$\xrightarrow{\text{CTFS}} dk = \frac{1}{T} \int_{-T/2}^{T/2} \delta(t) e^{-j k \omega_0 t} dt = \frac{1}{T}$$

So

$$x(t) = T \delta(t) \xrightarrow{\text{CTFS}} dk = 1$$

$$x(t) = T \delta(t) = \sum_{K=-\infty}^{\infty} dk e^{j k \omega_0 t}$$

$$T \delta(t) = \sum_{k=-\infty}^{\infty} e^{j k \omega_0 t}$$

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Problem 3.(25pt) Orthonormal Basis Functions

Let $\phi_k(t) = e^{j2\pi kt}$ for k an integer be functions on the interval $[0, 1]$ with associated inner product given by

$$\langle \phi_k, \phi_l \rangle = \int_0^1 \phi_k(t) \phi_l^*(t) dt .$$

- a) Prove that the functions $\{\phi_k\}_{k=-\infty}^{\infty}$ are normal.
- b) Prove that the functions $\{\phi_k\}_{k=-\infty}^{\infty}$ are orthogonal.
- c) Assume that $x(t)$ is a real valued signal that has the form

$$x(t) = \sum_{k=-\infty}^{\infty} a_k \phi_k(t) .$$

Then show that Parseval's theorem applies, i.e. that

$$\int_0^1 |x(t)|^2 dt = \sum_{k=-\infty}^{\infty} |a_k|^2$$

- d) Let $\tilde{x}(t)$ be formed by filtering the signal $x(t)$ so that only the first M harmonics are retained. Then the RMS error between these two signals is defined as

$$\text{RMS} = \sqrt{\int_0^1 |x(t) - \tilde{x}(t)|^2 dt}$$

Then calculate an expression for the RMS error between $x(t)$ and $\tilde{x}(t)$ in terms of the coefficients a_k .

$$\begin{aligned} a) \quad \langle \phi_n, \phi_n \rangle &= \int_0^1 e^{jk\omega_0 t} e^{-jn\omega_0 t} dt \\ &= \int_0^1 1 dt = 1 \end{aligned}$$

$$\begin{aligned} b) \quad \langle \phi_K, \phi_L \rangle &= \int_0^1 e^{jk\omega_0 t} e^{-jl\omega_0 t} dt \\ &= \int_0^1 e^{j\omega_0(K-l)t} dt \\ &= \frac{1}{j\omega_0(K-l)} \left[e^{j\frac{\omega_0}{2\pi}(K-l)t} \right]_0^1 = \frac{1}{j2\pi(K-l)} (1-1) \\ &= 0 \end{aligned}$$

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c) $\int_0^T |x(t)|^2 dt = \langle x, x \rangle$

$$= \left\langle \sum_{k=-\infty}^{\infty} a_k \phi_k, \sum_{l=-\infty}^{\infty} a_l \phi_l \right\rangle$$

$$= \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} a_k a_l^* \langle \phi_k, \phi_l \rangle$$

$$= \sum_{k=-\infty}^{\infty} \sum_{l=-\infty}^{\infty} a_k a_l^* \delta_{k-l} = \sum_{k=-\infty}^{\infty} |a_k|^2$$

d) $\int_0^T |x(t) - \tilde{x}(t)|^2 dt = \sum_{k=-\infty}^{M-1} |a_k|^2 + \sum_{k=M+1}^{\infty} |a_k|^2$

$$= 2 \sum_{k=M+1}^{\infty} |a_k|^2$$

$$\text{RMSE} = \sqrt{2 \sum_{k=M+1}^{\infty} |a_k|^2}$$

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Problem 4.(25pt) Periodic inputs to LTI systems

Let $x(t)$ be a period signal with period T and associated frequency $\omega_o = \frac{2\pi}{T}$; and let a_k be the associated Fourier series expansion coefficients for $x(t)$.

The signal $x(t)$ is used as the input to an LTI system with output $y(t)$ and real valued impulse response $h(t)$.

- Show that $y(t)$ is periodic with period T .
- Derive an expression for, b_k , the Fourier series coefficients of $y(t)$ in terms of the a_k 's and coefficients computed from the impulse response, $h(t)$.
- Calculate an expression for b_k when $h(t) = \text{rect}(t/T)$.
- Use the result of c) above to compute an explicit expression for $y(t)$.

$$a) \quad y(t) = \int_{-\infty}^{\infty} x(t-z) h(z) dz$$

$$\begin{aligned} y(t+T) &= \int_{-\infty}^{\infty} x(t+T-z) h(z) dz \\ &= \int_{-\infty}^{\infty} x(t-z) h(z) dz \\ &= y(t) \end{aligned}$$

$$b) \quad x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j k \omega_o t}$$

$$y(t) = S[x(t)] = S\left[\sum_{k=-\infty}^{\infty} a_k e^{j k \omega_o t}\right]$$

$$= \sum_{k=-\infty}^{\infty} a_k S[e^{j k \omega_o t}] = \sum_{k=-\infty}^{\infty} a_k c_k e^{j k \omega_o t}$$

$$\text{where } c_k = \int_{-\infty}^{\infty} h(t) e^{-j k \omega_o t} dt$$

So

$$b_k = a_k c_k$$

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$$\begin{aligned}
 \textcircled{c}) \quad C_n &= \int_{-\infty}^{\infty} \text{rect}(t/T) e^{-j k \omega_0 t} dt \\
 &= \int_{-T/2}^{T/2} e^{-j k \omega_0 t} dt = \frac{1}{-jk\omega_0} e^{-j k \omega_0 t} \Big|_{-T/2}^{T/2} \\
 &= \frac{1}{-jk\omega_0} [e^{j\pi k} - e^{-j\pi k}] \\
 &= T \frac{1}{\pi k} \frac{e^{j\pi k} - e^{-j\pi k}}{2j} \\
 &= T S_{\text{SWC}}(k) = T \delta(k)
 \end{aligned}$$

$$b_k = a_k \delta(k) = \begin{cases} a_0 & \text{for } k=0 \\ 0 & \text{otherwise} \end{cases}$$

$$\text{d)} \quad Y(t) = \sum_{k=-\infty}^{\infty} b_k e^{j k \omega_0 t} = b_0 \mathbf{1} = a_0$$

$$Y(t) = a_0$$