

Problem 1 (30 points):

Consider a circular waveguide as shown in Fig. 1. The waveguide wall is made of perfect electric conductor (PEC). It is uniformly filled with permittivity ϵ_0 and permeability μ_0 .

- (a) (10 points) Can this waveguide support TE_x (i.e. $E_x=0$) and TE_y (i.e. $E_y=0$) modes? Please refer to Fig. 1 for the coordinate system.
(b) (20 points) Justify your answer to question (a).

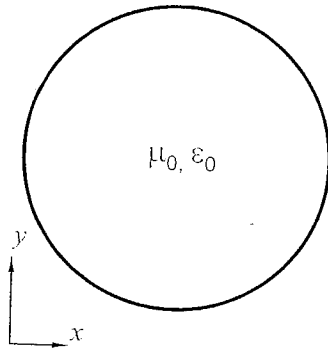


Fig. 1. A circular waveguide made of PEC.

Problem 2 (30 points):

Consider a rectangular waveguide loaded with two dielectric media as shown in Fig. 2. The permeability and permittivity of the two media are μ_0, ϵ_0 , and μ_0, ϵ respectively. The waveguide wall is made of perfect electric conductor.

- (a) (10 points) Can this waveguide support TM_z modes?
(b) (20 points) Justify your answer to question (a).

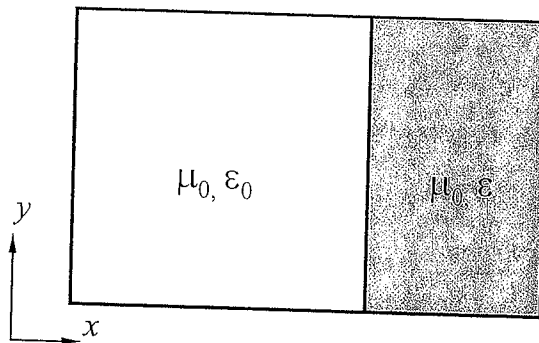


Fig. 2. A partially filled rectangular waveguide.

Problem 3 (40 points):

An infinitely long, uniform rectangular waveguide is filled with air having permittivity ϵ_0 and permeability μ_0 (see the figure below for its cross section). The waveguide wall is made of perfect magnetic conductor (PMC). Note that tangential magnetic field vanishes on a PMC surface. Derive the general expression of H_z for TE_z modes. (Hint: H_z satisfies $\nabla^2 H_z + k^2 H_z = 0$, $k^2 = \omega^2 \mu_0 \epsilon_0$).

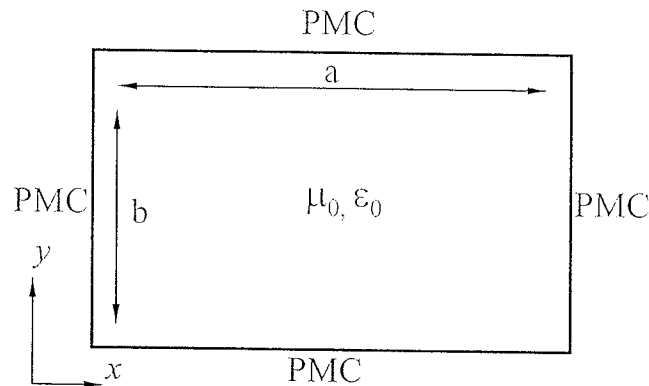


Fig. 3. A rectangular waveguide made of PMC.

Potentially Useful Formula:

Assuming $e^{j(\omega t - k_z z)}$ field dependence,

E_x, E_y, H_x, H_y can be written in terms of E_z and H_z as:

$$E_x = -\frac{1}{k^2 - k_z^2} \left(-jk_z \frac{\partial E_z}{\partial x} + j\omega\mu \frac{\partial H_z}{\partial y} \right)$$

$$E_y = \frac{1}{k^2 - k_z^2} \left(jk_z \frac{\partial E_z}{\partial y} + j\omega\mu \frac{\partial H_z}{\partial x} \right)$$

$$H_x = \frac{1}{k^2 - k_z^2} \left(j\omega\epsilon \frac{\partial E_z}{\partial y} + jk_z \frac{\partial H_z}{\partial x} \right)$$

$$H_y = -\frac{1}{k^2 - k_z^2} \left(j\omega\epsilon \frac{\partial E_z}{\partial x} - jk_z \frac{\partial H_z}{\partial y} \right)$$

Vector Potentials \mathbf{A} and \mathbf{F} satisfy:

$$\nabla^2 \mathbf{A} + k^2 \mathbf{A} = -\mu \mathbf{J}$$

$$\nabla^2 \mathbf{F} + k^2 \mathbf{F} = -\epsilon \mathbf{M}$$

Fields in terms of \mathbf{A} and \mathbf{F} :

$$\mathbf{E} = -j\omega \mathbf{A} - \frac{j}{\omega\mu\epsilon} \nabla \nabla \cdot \mathbf{A} - \frac{1}{\epsilon} \nabla \times \mathbf{F}$$

$$\mathbf{H} = \frac{1}{\mu} \nabla \times \mathbf{A} - j\omega \mathbf{F} - \frac{j}{\omega\mu\epsilon} \nabla \nabla \cdot \mathbf{F}$$

Maxwell's Equations, in differential and integral form are: Other potentially useful relationships are:

$$\begin{aligned} \vec{\nabla} \times \vec{E} &= -\frac{\partial \vec{B}}{\partial t} & \oint_c \vec{E} \cdot d\vec{\ell} &= -\frac{\partial}{\partial t} \int_s \vec{B} \cdot d\vec{s} & \vec{D} &= \epsilon_0 \vec{E} + \vec{P} = \epsilon_0 \vec{E} \\ \vec{\nabla} \times \vec{H} &= \vec{J} + \frac{\partial \vec{D}}{\partial t} & \oint_c \vec{H} \cdot d\vec{\ell} &= \int_s \left(\vec{J} + \frac{\partial \vec{D}}{\partial t} \right) \cdot d\vec{s} & \vec{B} &= \mu_0 (\vec{H} + \vec{M}) = \mu \vec{H} \\ \vec{\nabla} \cdot \vec{D} &= \rho & \oint_s \vec{D} \cdot d\vec{s} &= Q & \lambda &= \frac{2\pi}{\beta} = \frac{u}{f} = \frac{1}{f\sqrt{\mu\epsilon}} \\ \vec{\nabla} \cdot \vec{B} &= 0 & \oint_s \vec{B} \cdot d\vec{s} &= 0 & \epsilon_0 &= 8.854 \times 10^{-12} \text{ F/m} \\ & & & & \mu_0 &= 4\pi \times 10^{-7} \text{ H/m} \end{aligned}$$

Vector Differential Relationships in Cylindrical Coordinates (r, ϕ , z)

$$\vec{\nabla} V = \hat{a}_r \frac{\partial V}{\partial r} + \hat{a}_\phi \frac{\partial V}{r \partial \phi} + \hat{a}_z \frac{\partial V}{\partial z}$$

$$\vec{\nabla} \cdot \vec{A} = \frac{1}{r} \frac{\partial}{\partial r} (r A_r) + \frac{\partial A_\phi}{r \partial \phi} + \frac{\partial A_z}{\partial z}$$

$$\vec{\nabla} \times \vec{A} = \hat{a}_r \left(\frac{\partial A_z}{r \partial \phi} - \frac{\partial A_\phi}{\partial z} \right) + \hat{a}_\phi \left(\frac{\partial A_r}{\partial z} - \frac{\partial A_z}{\partial r} \right) + \hat{a}_z \frac{1}{r} \left[\frac{\partial}{\partial r} (r A_\phi) - \frac{\partial A_r}{\partial \phi} \right]$$

$$\nabla^2 V = \frac{1}{r} \frac{\partial}{\partial r} \left(r \frac{\partial V}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2 V}{\partial \phi^2} + \frac{\partial^2 V}{\partial z^2}$$

Divergence Thm: $\int_v \vec{\nabla} \cdot \vec{A} dv = \oint_s \vec{A} \cdot d\vec{s}$

For lossless transmission lines:

Stokes Thm: $\int_v \vec{\nabla} \times \vec{A} \cdot d\vec{s} = \oint_c \vec{A} \cdot d\vec{\ell}$

$$Z_{in} = R_o \frac{Z_L + jR_o \tan \beta \ell}{R_o + jZ_L \tan \beta \ell}$$