# ECE 538 Digital Signal Processing I Final Exam 2014 Test Date: Tuesday, 16 December 2014

#### **Cover Sheet**

Test Duration: 120 minutes.

Open Book but Closed Notes.

One 8.5 x 11 handwritten crib sheet allowed.

Calculators NOT allowed.

This test contains **FOUR** problems.

All work should be done in the proper space provided.

Good luck in this final exam!
Great having you in class this semester!
Have a joyous holiday season!

**Problem 1.** Consider the upsampler system below in Figure 1.

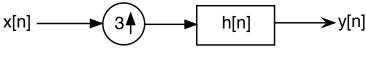
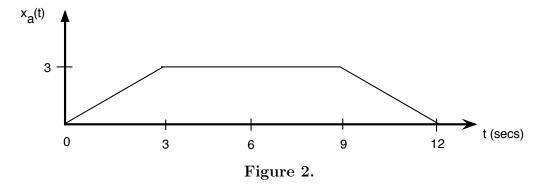


Figure 1.

- (a) Draw block diagram of efficient implementation of the upsampler system in Fig. 1.
- (b) Your answer to part (a) should involve the polyphase components of h[n]:  $h_0[n] = h[3n]$ ,  $h_1[n] = h[3n+1]$ , and  $h_2[n] = h[3n+2]$  and the DTFT of h[n], denoted  $H(\omega)$ .
  - (i) For the ideal case where  $h[n] = 3 \frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , plot the magnitude of the DTFT of  $h_0[n] = h[3n], H_0(\omega)$ , over  $-\pi < \omega < \pi$ .
  - (ii) For the general case where h[n] is an arbitrary impulse response, express the DTFT of  $h_1[n] = h[3n+1]$ , denoted  $H_1(\omega)$ , in terms of  $H(\omega)$ .
  - (iii) For the ideal case where  $h[n] = 3\frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , plot both the magnitude AND phase (two separate plots) of the DTFT  $h_1[n] = h[3n+1]$ ,  $H_1(\omega)$ , over  $-\pi < \omega < \pi$ .
  - (iv) Express the DTFT of  $h_2[n] = h[3n+2]$ , denoted  $H_2(\omega)$ , in terms of  $H(\omega)$ .
  - (v) For the ideal case where  $h[n] = 3\frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , plot both the magnitude AND phase (two separate plots) of the DTFT  $h_2[n] = h[3n+2]$ ,  $H_2(\omega)$ , over  $-\pi < \omega < \pi$ .
- (c) Consider that the input to the system in Figure 1 is a sampled version of the analog signal in Figure 2. For the remaining parts of this problem, the input signal is as defined below where  $x_a(t)$  is the analog signal in Figure 2. Assume that  $1/T_s = 1/3$  is above the Nyquist rate for this signal. That is, even though this signal is not strictly bandlimited, assume that aliasing effects are negligible.

$$x[n] = x_a(nT_s), \quad T_s = 3$$



(i) For the ideal case where  $h[n] = 3\frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , determine the output y[n] of the system in Figure 1, when x[n] is input to the system. Write output in sequence form (indicating where n = 0 is) OR do stem plot.

- (ii) For the ideal case where  $h[n] = 3 \frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , determine the output  $y_0[n] = x[n] * h_0[n]$ , when x[n] is input to the filter  $h_0[n] = h[3n]$ . Write output in sequence form (indicating where is n = 0 OR do stem plot.
- (iii) For the ideal case where  $h[n] = 3 \frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , determine the output  $y_1[n] = x[n] * h_1[n]$ , when x[n] is input to the filter  $h_1[n] = h[3n+1]$ . Write output in sequence form (indicating where n = 0 is) OR do stem plot.
- (iv) For the ideal case where  $h[n] = 3 \frac{\sin(\frac{\pi}{3}n)}{\pi n}$ , determine the output  $y_2[n] = x[n] * h_2[n]$ , when x[n] is input to the filter  $h_2[n] = h[3n+2]$ . Write output in sequence form (indicating where n=0 is) OR do stem plot.

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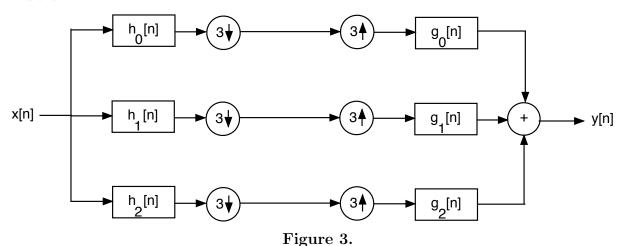
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## Digital Signal Processing I

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Problem 2.



In the frequency domain, the DTFT of the output is related to the DTFT at the input according to the expression below.

$$Y(\omega) = A(\omega)X(\omega) + B(\omega)X\left(\omega - \frac{2\pi}{3}\right) + C(\omega)X\left(\omega - \frac{4\pi}{3}\right)$$

- (a) Determine expressions for  $A(\omega)$ ,  $B(\omega)$ , and  $C(\omega)$  in terms of  $H_0(\omega)$ ,  $H_1(\omega)$ , and  $H_2(\omega)$  and  $G_0(\omega)$ ,  $G_1(\omega)$ , and  $G_2(\omega)$ .
- (b) Consider the case where  $g_k[n] = h_k^*[-n]$ , k = 0, 1, 2. Determine expressions for  $A(\omega)$ ,  $B(\omega)$ , and  $C(\omega)$  in terms of  $H_0(\omega)$ ,  $H_1(\omega)$ , and  $H_2(\omega)$ .
- (c) Consider the case where  $g_k[n] = h_k^*[-n]$ , k = 0, 1, 2 AND  $h_k[n] = e^{jk\frac{2\pi}{3}n}h[n]$ , k = 0, 1, 2. Determine expressions for  $A(\omega)$ ,  $B(\omega)$ , and  $C(\omega)$  all in terms of  $H(\omega)$ .
- (d) Consider the case where  $g_k[n] = h_k^*[-n]$ , k = 0, 1, 2 AND  $h_k[n] = e^{jk\frac{2\pi}{3}n}h[n]$ , k = 0, 1, 2 AND h[n] = u[n] u[n-3]. Show that  $A(\omega) = 9$ ,  $B(\omega) = 0$ , and  $C(\omega) = 0$  for all  $\omega$ .
- (e) Under the conditions stated in part (d) directly above, develop and draw the efficient implementation of the entire Filter Bank in Figure 3. Show as much detail as possible. Using the efficient implementation, explain how the system is a Perfect Reconstruction Filter Bank.

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**Problem 3.** Let x[n] be of length L=7, i.e., x[n]=0 for n<0 and  $n\geq 7$ , and h[n] also be of length M=7. Let  $X_8(k)$  and  $H_8(k)$  denote 8-point DFT's of x[n] and h[n], respectively. The 8-point inverse DFT of the product  $Y_8(k)=X_8(k)H_8(k)$ , denoted  $y_8[n]$ , produces the following values:

n	0	1	2	3	4	5	6	7
$y_8[n]$	-2	0	-2	0	-2	0	7	0

Let  $X_{10}(k)$  and  $H_{10}(k)$  denote the 10-point DFT's of the aforementioned sequences x[n] and h[n]. The 10-point inverse DFT of the product  $Y_{10}(k) = X_{10}(k)H_{10}(k)$ , denoted  $y_{10}[n]$ , produces the following values:

n	0	1	2	3	4	5	6	7	8	9
$y_{10}[n]$	-2	0	-2	0	-1	0	7	0	-1	0

Given  $y_8[n]$  and  $y_{10}[n]$ , find the **linear** convolution of x[n] and h[n], i. e., list all of the numerical values of y[n] = x[n] \* h[n]. You must briefly explain and show all work.

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## Digital Signal Processing I

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Problem 4.

(a) Let  $X_8(k) = X(2\pi k/8)$ , where  $X(\omega)$  is the DTFT of the sequence

$$x[n] = (0.9)^n u[n] \stackrel{DTFT}{\longleftrightarrow} X(\omega) = \frac{1}{1 - 0.9e^{-j\omega}}$$

That is,  $X_8(k)$  is what we obtain by sampling  $X(\omega)$  at N=8 equi-spaced points in the interval  $0 \le \omega < 2\pi$ . Theory derived in class and in the textbook dictates that the 8-pt inverse DFT of  $X_8(k)$  may be expressed as

$$x_8[n] = \sum_{\ell=-\infty}^{\infty} x[n-\ell 8] \{u[n] - u[n-8]\} \xrightarrow{OFT} X_8(k) = \frac{1}{1 - 0.9e^{-j2\pi k/8}}; k = 0, 1, ..., 7$$

Determine a simple, closed-form expression for  $x_8[n]$ . A *closed-form* expression contains NO summations and it is NOT a listing of numbers. *Hint:* 

$$\frac{1}{1 - (.9)^8} = 1.7558$$

- (b) Consider normalizing  $x_8[n]$  so that it's first value is one,  $\tilde{x}_8[n] = x_8[n]/x_8[0]$ , n = 0, 1, ..., 7. Compare  $\tilde{x}_8[n]$  and x[n] over n = 0, 1, ..., 7. Are they the same or different? Briefly explain your answer as to why or why not they are the same.
- (c) Let  $Y_8(k) = Y(2\pi k/8)$ , where  $Y(\omega)$  is the DTFT of the sequence

$$y[n] = (0.9)^n u[n] + (0.8)^n u[n]$$

That is,  $Y_8(k)$  is what we obtain by sampling  $Y(\omega)$  at N=8 equi-spaced points in the interval  $0 \le \omega < 2\pi$ . Theory derived in class and in the textbook dictates that the 8-pt inverse DFT of  $Y_8(k)$  may be expressed as

$$y_8[n] = \sum_{\ell = -\infty}^{\infty} y[n - \ell 8] \{u[n] - u[n - 8]\} \overset{DFT}{\longleftrightarrow} Y_8(k) = \frac{1}{1 - 0.9e^{-j2\pi k/8}} + \frac{1}{1 - 0.8e^{-j2\pi k/8}}$$

Determine a simple, closed-form expression for  $y_8[n]$ . A *closed-form* expression contains NO summations and it is NOT a listing of numbers. *Hint:* 

$$\frac{1}{1 - (.8)^8} = 1.2016$$

(d) Consider normalizing  $y_8[n]$  so that it's first value is one,  $\tilde{y}_8[n] = y_8[n]/y_8[0]$ , n = 0, 1, ..., 7. Compare  $\tilde{y}_8[n]$  and y[n] over n = 0, 1, ..., 7. Are they the same or different? Briefly explain your answer as to why or why not they are the same.

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